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CANDOFD (Computer Analysis of Networks With
Design Orientation in the Frequency Domain)

by

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Memorandum No. ERL-M253

9 September 1968

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Abstract:

This report describes the operation and use of CANDOFD (Computer Analysis of Networks with Design Orientation in the Frequency Domain), a frequency domain analysis program, for linear time invariant networks. The networks may contain dependent and independent sources of all types, capacitances, resistances and inductances.

The network analysis problem is to obtain the complex branch currents and voltages, by solving a set of simultaneous complex algebraic equations derived from the complex branch relations and Kirchhoff's voltage and current laws.

In the program CANDOFD, nonindependent source tree voltages and non-independent source link currents form a basis set of variables, denoted by \underline{x} . This formulation yields the automatic satisfaction of Kirchhoff's laws.

The problem reduces to solving a set of n coupled complex equations of the form

$$E_i(\underline{x}) = 0 \quad (1)$$

The set of equations (1) is solved by minimizing a performance criterion ϵ , defined as

$$\epsilon = \sum_n E_i \cdot E_i^* \quad (2)$$

where $*$ denotes the complex conjugate. Fletcher-Powell⁽²⁾ and Rohrer⁽³⁾ search minimization and step size selection algorithms are utilized, and the iterative procedure is terminated when ϵ is reduced to a desired value or when the desired number of minimization steps have taken place.

The tree picking and internal current scaling algorithms are such that large value spread and large impedance spread problems can be handled

reasonably effectively and efficiently. The DC response of a network may be obtained by simply solving the zero frequency problem.

The program CANDOFD is written in FORTRAN IV for the CDC 6400 computer operating under the (CAL) Scope 3.1 system.

Acknowledgments:

I wish to acknowledge the invaluable assistance and encouragement of Professor R. A. Rohrer, whose guidance was instrumental in the production of CANDOFD.

I also wish to acknowledge the financial support of the Air Force Office of Scientific Research, under Grant AF-AFOSR-1219-67, and the Computer Center of the University of California at Berkeley, which has been most gracious in its granting of subsidized computer time.

Introduction:

CANDOFD is the frequency domain version of CANDO⁽¹⁾, and, as in CANDO, a trade-off exists between speed and accuracy of analysis. CANDOFD obtains a network analysis, at a specified frequency, by minimizing a performance criterion related to the complex branch errors, with Kirchhoff's laws being automatically (implicitly) satisfied. The iterative minimization scheme is terminated when either the squared branch error sum (performance criterion) falls below a desired value, or when a specified number of iterative minimization steps have been executed.

CANDOFD, with its accuracy vs speed trade off, is a Design Oriented analysis program in the sense that, in the initial phases of design, a quick (but not very accurate) analysis is desirable, whereas, in the final phase of design, an accurate analysis is necessary.

Section I describes the theory related to the formulation and minimization of the desired performance criterion, and indicates the solution to both the DC and general frequency point problems. Section II describes the MAIN program and subroutines associated with CANDOFD. The Appendices describe the use of CANDOFD in solving network problems, the way dependent sources must be modeled, and include a selection of sample problems.

in an open circuit, and no DC solution exists for such a network. If the independent current sources, in such a cutset, are zero valued, CANDOFD yields the correct network DC solution.

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SECTION I: NETWORK ANALYSIS

Theory:

The solution of a network, in the frequency domain, consists of finding the complex vectors \hat{i}_b and \hat{v}_b , representing all the branch currents and voltages, respectively. The solution is obtained by solving the following simultaneous set of linear equations.

$$f_b(\hat{v}_b, \hat{i}_b) = 0 \quad \text{complex branch relations}$$

$$\hat{Q}\hat{i}_b = 0 \quad \text{Kirchhoff's current law}$$

$$\hat{B}\hat{v}_b = 0 \quad \text{Kirchhoff's voltage law}$$

where \hat{Q} is the fundamental cutset matrix and \hat{B} is the fundamental loop matrix. Upon the selection of an optimal tree (see subroutine PT), we may renumber the NB branches of our network in such a way that the 1st NN-1 branches (NN is the number of nodes in our network), form the tree.

With the above numbering scheme, we may partition \hat{Q} and \hat{B} as follows:

$$\hat{Q} = [\hat{I} \mid \hat{F}] \quad \text{and} \quad \hat{B} = [-\hat{F}' \mid \hat{I}]$$

where \hat{I} is the identity matrix. Hence we have, from Kirchhoff's laws;

$$\hat{i}_t = -\hat{F}\hat{i}_l$$

and

$$\hat{v}_l = \hat{F}'\hat{v}_t,$$

where the subscripts t and l refer to tree branches and links respectively.

Our optimal tree picking algorithm requires that all independent voltage sources be tree branches and that all independent current sources

The first part of the report deals with the general principles of the theory of the structure of the atom. It is shown that the structure of the atom is determined by the laws of quantum mechanics, and that the energy levels of the atom are given by the solutions of the Schrödinger equation. The second part of the report deals with the application of these principles to the structure of the hydrogen atom. It is shown that the energy levels of the hydrogen atom are given by the solutions of the Schrödinger equation, and that the wave functions of the hydrogen atom are given by the solutions of the Schrödinger equation.

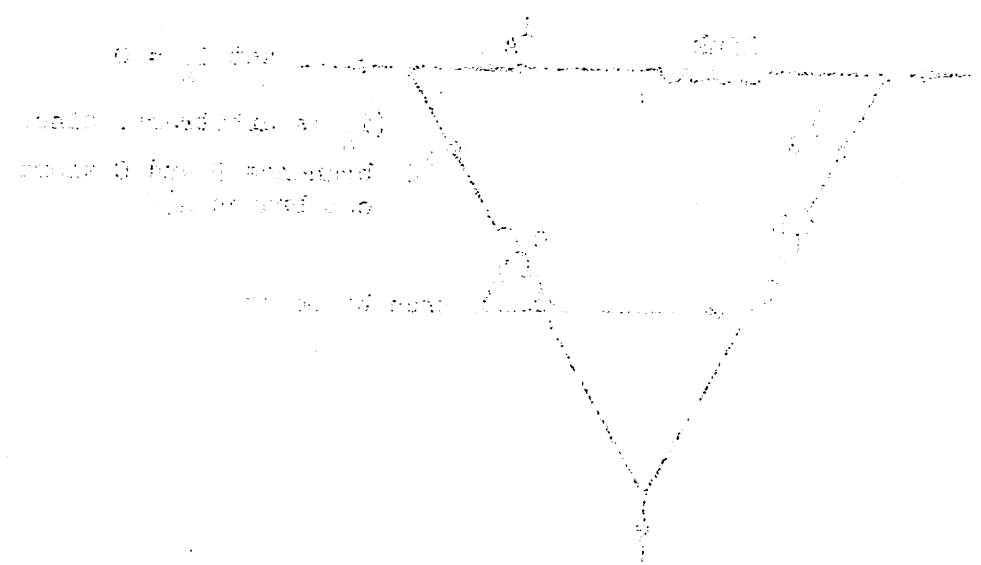


Fig. 1

The third part of the report deals with the application of these principles to the structure of the hydrogen atom. It is shown that the energy levels of the hydrogen atom are given by the solutions of the Schrödinger equation, and that the wave functions of the hydrogen atom are given by the solutions of the Schrödinger equation. The fourth part of the report deals with the application of these principles to the structure of the hydrogen atom. It is shown that the energy levels of the hydrogen atom are given by the solutions of the Schrödinger equation, and that the wave functions of the hydrogen atom are given by the solutions of the Schrödinger equation.

The fifth part of the report deals with the application of these principles to the structure of the hydrogen atom. It is shown that the energy levels of the hydrogen atom are given by the solutions of the Schrödinger equation, and that the wave functions of the hydrogen atom are given by the solutions of the Schrödinger equation.

Kirchhoff's current and voltage laws become automatically satisfied by selecting the tree voltages and link currents to be a basis set of variables, with the following being true;

$$\hat{i}_t = - \hat{F} \hat{i}_l$$

and

$$\hat{v}_l = \hat{F}' \hat{v}_t$$

where, as before, the subscripts t and l refer to tree branches and links respectively, and where $\hat{Q} = [\hat{I} \mid \hat{F}]$ is the fundamental cutset matrix based on our optimal tree.

To further reduce the order of the system, independent sources and reactive elements are not considered to be variables. (In the DC problem, all reactive elements act as zero valued independent sources.)

Thus the system of equations we are solving is one corresponding to a network consisting solely of independent sources (with real values), dependent sources and resistances.

The network solution is obtained in a way analogous to the solution of the equations associated with the "general frequency problem" (see next section), and will not be detailed here.

Solving the general frequency problem:

We wish to find \hat{i}_b and \hat{v}_b , the complex branch current and voltage vectors respectively, at a general frequency ω . The problem reduces to solving a coupled system of complex linear equations, consisting of the branch relations, and Kirchhoff's current and voltage laws.

A tree is selected according to our optimal tree algorithm (see subroutine PT), allowing us to reduce the order of the system, and hence to

improve the convergence of our iterative minimization scheme.

The network is renumbered so that the 1st $NN-1$ branches form the tree, and the subsequent $(NB-NN+1)$ branches are links. Kirchhoff's current and voltage laws become automatically satisfied by selecting the tree branch voltages and link currents as a basis set of variables.

If $\underline{Q} = [\underline{I} \mid \underline{F}]$ is the fundamental cutset matrix associated with our optimal tree, we have

$$\underline{i}_{\underline{t}} = - \underline{F} \underline{i}_{\underline{l}}$$

and

$$\underline{v}_{\underline{l}} = \underline{F}' \underline{v}_{\underline{t}}$$

we thus wish to solve the complex set of coupled linear equations

$$\underline{E}(\underline{v}_{\underline{t}}, \underline{i}_{\underline{l}}) = \underline{0} \quad (1)$$

which are our complex branch errors (See Subroutine ERROR).

To further reduce the order of the system, independent source values are not treated as variables. Notice that this requires that independent voltage sources be tree branches, and that independent current sources be links.

The solution to (1) is obtained by minimizing a performance criterion ϵ , defined by

$$\epsilon = \sum_i \underline{E}_i^* \cdot \underline{E}_i = \underline{E}_{\underline{i}}^* \cdot \underline{E}_{\underline{i}} \quad (2)$$

Note that we have an exact solution when $\epsilon = 0$

The Fletcher-Powell⁽²⁾ minimization algorithm (see subroutine FLPOW) yields the directions along which ϵ is minimized, and Rührer search⁽³⁾ (see

subroutine FNDALPH) is used to find the corresponding optimal step size.

The variables are selected to be the nonindependent source tree voltages (real and imaginary parts are distinct variables) and the non-independent source link currents. The Fletcher-Powell and Rohrer search algorithms require the knowledge of the gradient vectors,

$$\frac{\partial \epsilon}{\partial \underline{v}_{tr}}, \quad \frac{\partial \epsilon}{\partial \underline{v}_{ti}}, \quad \frac{\partial \epsilon}{\partial \underline{i}_{lr}}, \quad \frac{\partial \epsilon}{\partial \underline{i}_{li}}$$

where the subscripts r and i indicate the real and imaginary parts respectively. These gradient vectors are computed in subroutine FNDGRAD. Once ϵ is "sufficiently" minimized, \underline{v}_t and \underline{i}_ℓ are "known" (to a desired accuracy). We then compute \underline{i}_b and \underline{v}_b as follows:

$$\underline{i}_b = \begin{bmatrix} -\underline{F} \\ \underline{I} \end{bmatrix} \underline{i}_\ell$$

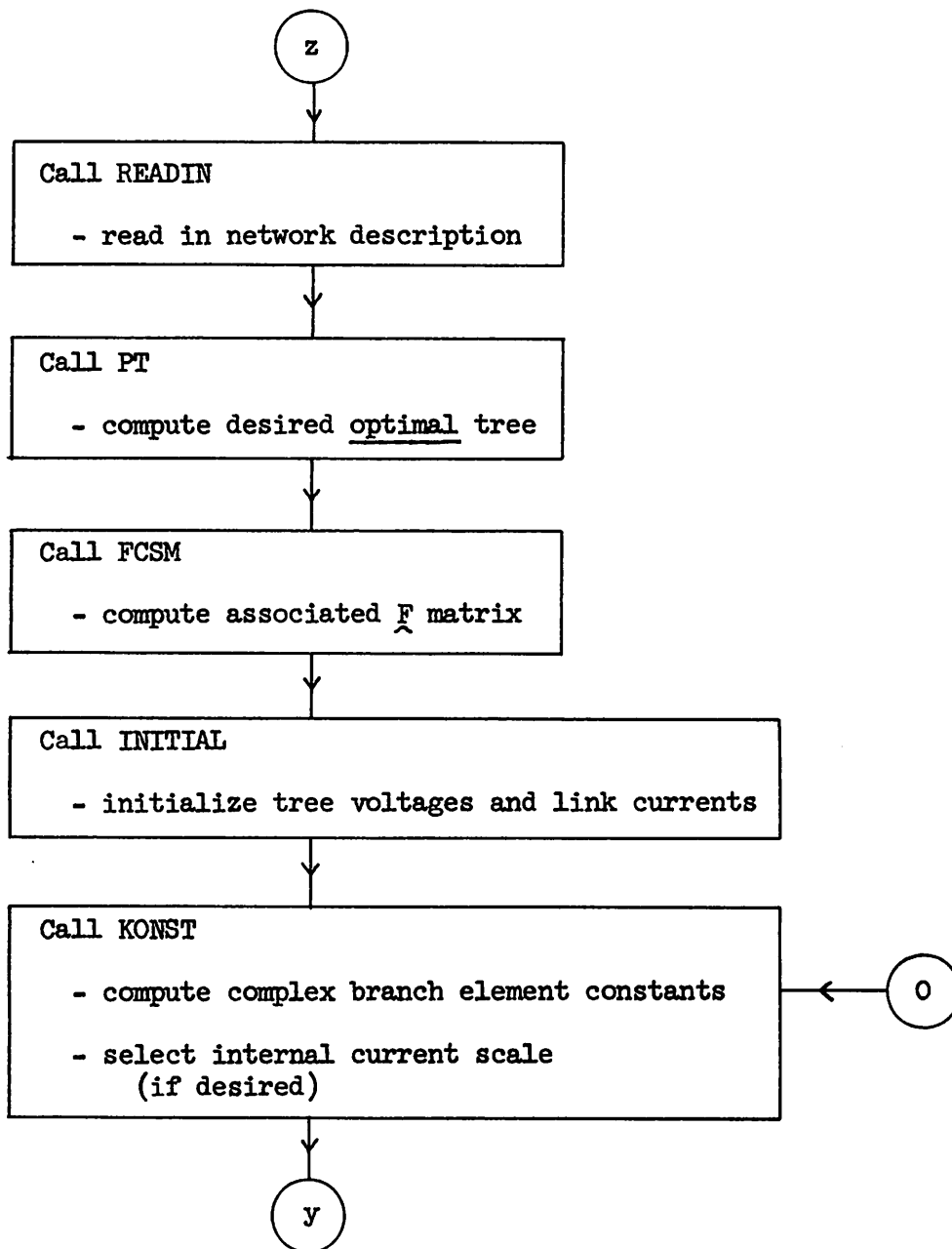
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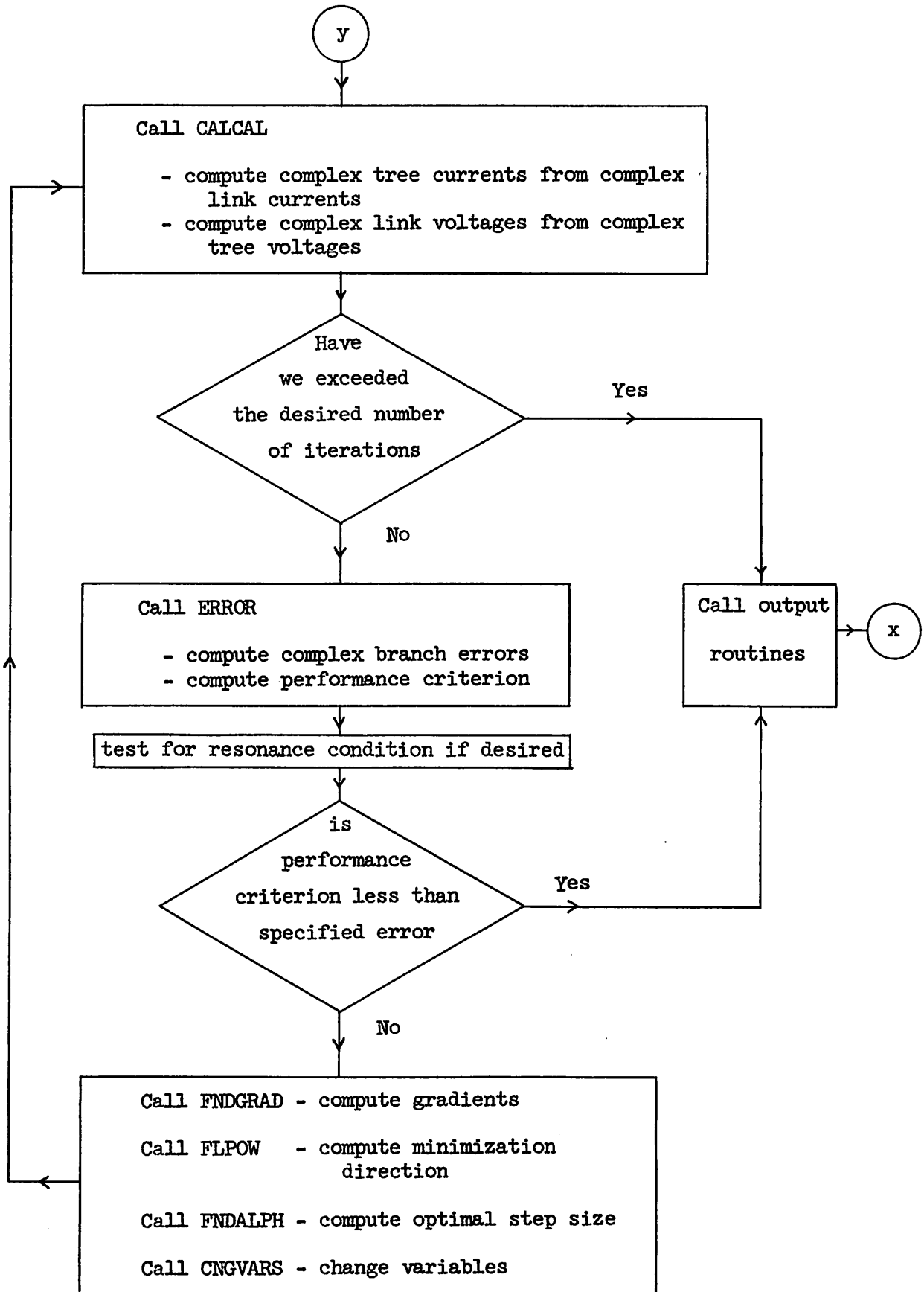
$$\underline{v}_b = \begin{bmatrix} \underline{I} \\ \underline{F}' \end{bmatrix} \underline{v}_t,$$

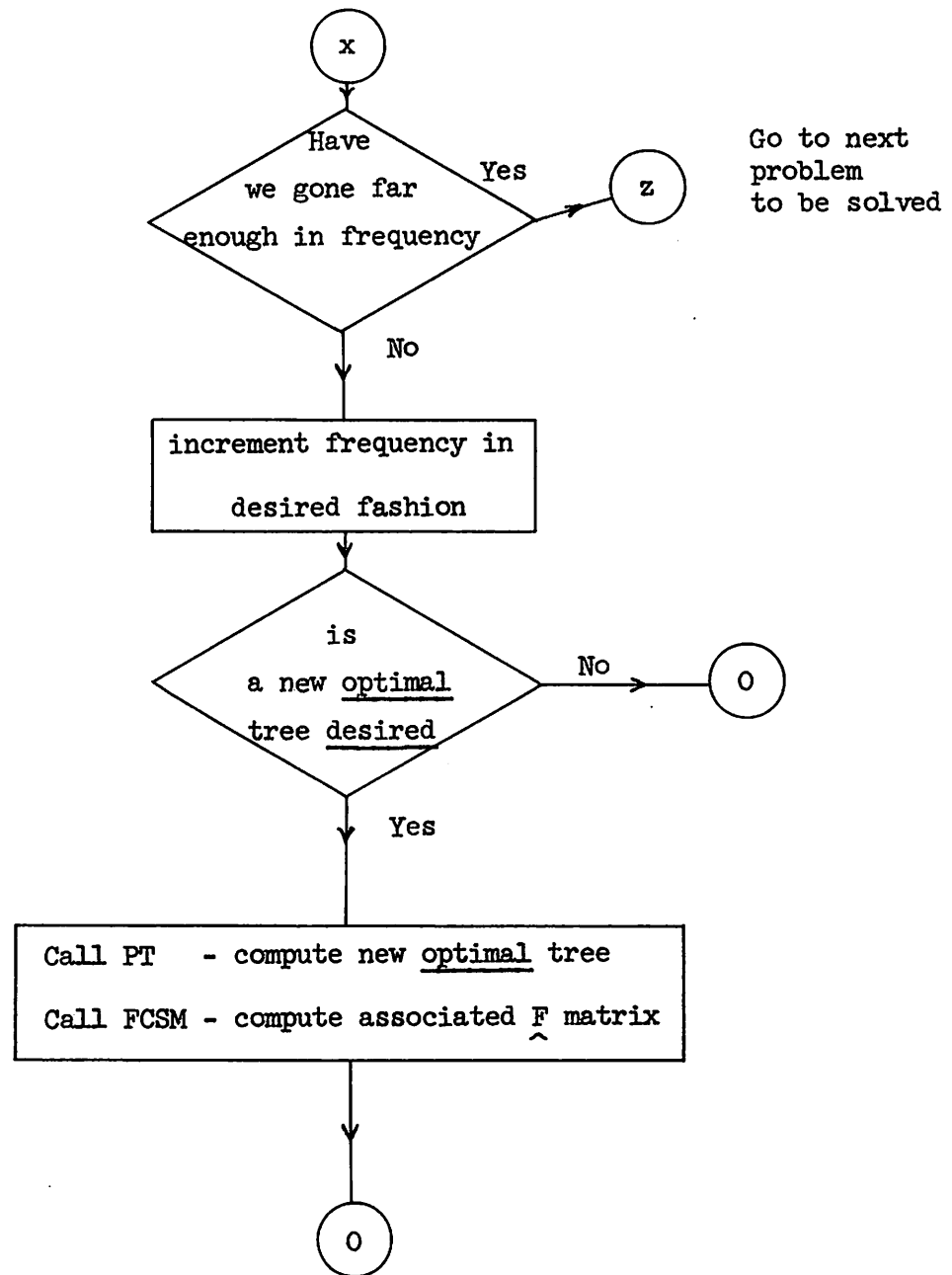
and hence, we proceed (if desired) to the next frequency point in an analogous way. The special case, for lossless networks, where the desired frequency corresponds to a resonance frequency, is discussed in the following section.

SECTION II. PROGRAM CANDOFDMain Program:

The main section of program CANDOFD ensures that the desired subroutines are called in the correct sequence. It solves the DC and general frequency point problems, in the way described in Section I. The main program also controls the calls to the output subroutines.

Block Diagram:





Subroutine READIN:

This subroutine reads in the complete network description along with all pertinent output and control specifications. As a data card check, the given network description and the pertinent control specifications are printed out.

For a detailed description of the variables read in, see the section entitled "Data Cards for CANDOFD."

Subroutine PT:

At each call, subroutine PT computes an optimal tree. The optimal tree picking scheme will be enunciated below.

If the given connected network has NN nodes and NB branches, it will have NN-1 tree branches and NB-NN+1 links.

An optimal tree consists of the first NN-1 branches which do not form loops, with the following selection priority scheme (in descending order):

- 1 - independent voltage sources
- 2 - equivalent impedance magnitudes in ascending order of value
- 3 - independent current sources

The equivalent impedance magnitude values are given by the following prescription:

Controlled voltage source $\rightarrow 10^{-50}$

Resistance (R) $\rightarrow R$

Capacitance (C) $\rightarrow \frac{1}{\omega * C}$

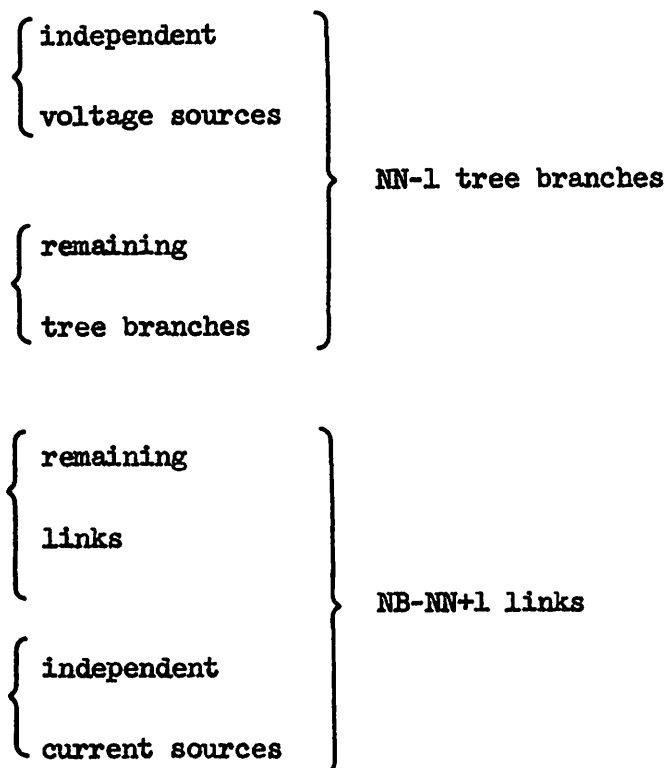
Inductance (L) $\rightarrow \omega * L$

Controlled current source $\rightarrow 10^{50}$

where ω is the frequency at which the optimal tree is being computed. When $\omega * C = 0$, the corresponding impedance magnitude is set to 10^{51} .

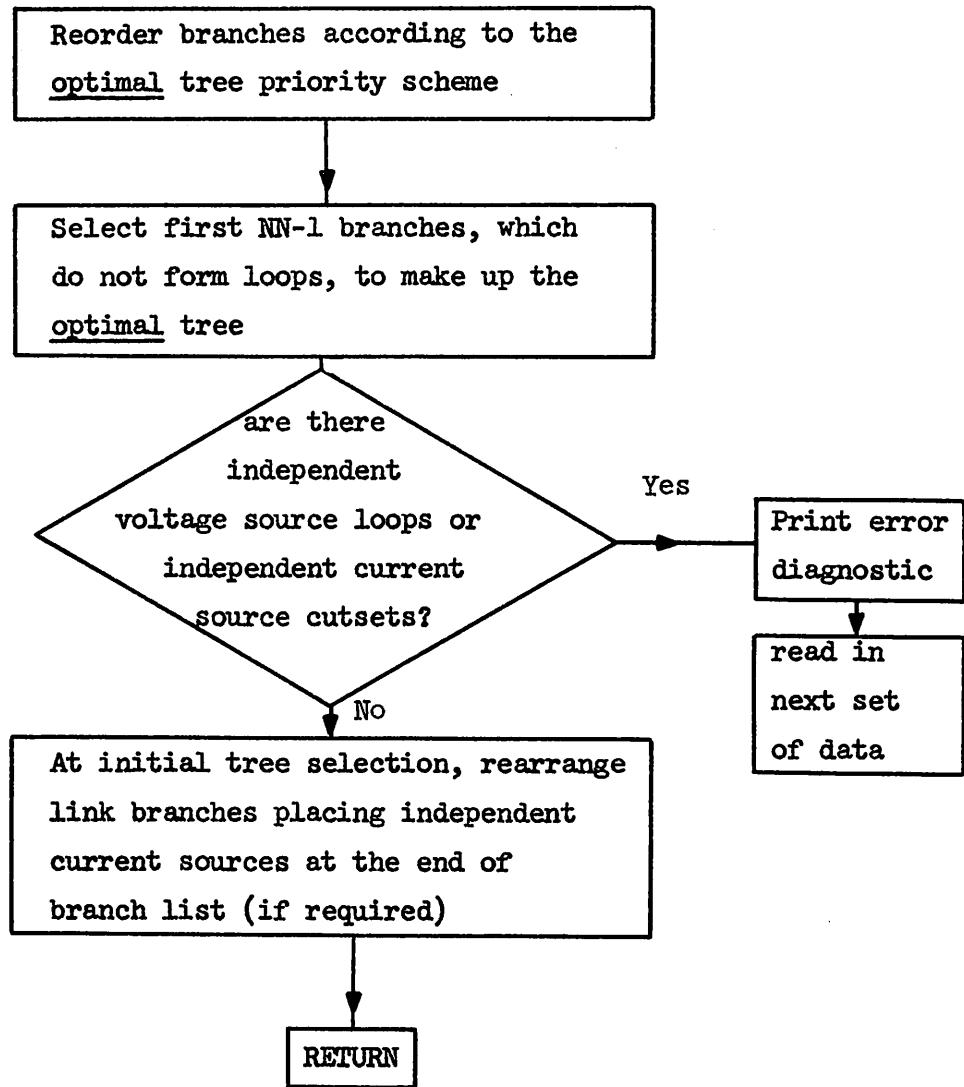
An independent voltage source appearing in a link indicates the presence of a loop of independent voltage sources. An independent current source appearing in a tree branch indicates the presence of an independent current source cutset. In both of these cases, an error diagnostic will be printed out, and the program will proceed to the next set of data.

The internal numbering scheme is as follows:



The above numbering scheme allows us to ignore the independent source branches during subsequent optimal tree selections.

Subroutine PT, when desired, prints out the computed optimal tree and the rearranged network.

Block diagram:Subroutine FT

Subroutine FCSM: (Fundamental Cutset Matrix)

Given a connected network, and a specified tree, there exists a unique (real) $(NN-1 \times NB)$ fundamental cutset matrix \underline{Q} , where NN and NB are the number of nodes and the number of branches in the network, respectively.

Kirchhoff's current law is then given by

$$\underline{Q} \underline{i}_b = \underline{0} \quad ,$$

where \underline{i}_b is the complex branch current vector.

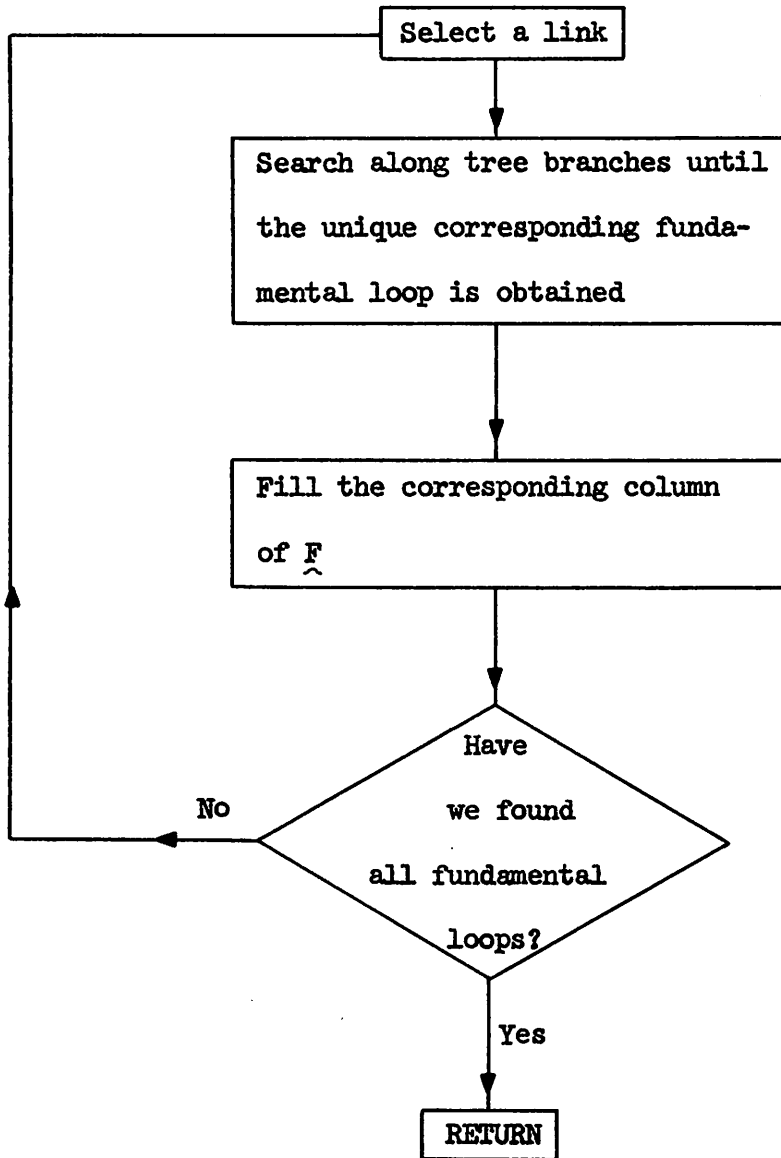
If the branch numbering is selected such that the first $NN-1$ branches are tree branches (see subroutine PT), then \underline{Q} can be partitioned as follows:

$$\underline{Q} = [\underline{I} \quad \underline{F}] \quad ,$$

and only the $(NN-1 \times NB-NN+1)$ \underline{F} matrix need be computed and stored, resulting in a substantial saving of computer memory.

To each link, of our connected network, there corresponds a unique fundamental loop, consisting of the link itself and sufficient tree branches required to close the loop. Each column of \underline{F} corresponds to such a fundamental loop. This intermediate scheme is used to compute \underline{F} .

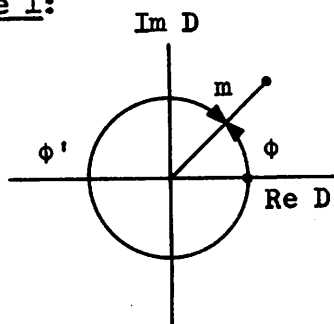
If desired, FCSM can output the \underline{F} portion of the fundamental cutset matrix.

Block Diagram:Subroutine FCSM

Subroutine VOLCUR:

This subroutine computes the real and imaginary components of independent source signals, from their magnitudes and phases (-360° to 360°).

Let D designate the complex independent source signal. Let m designate the given signal magnitude and let ϕ designate the given signal phase (-360° to 360°). ϕ' designates an equivalent, allowed, signal phase.

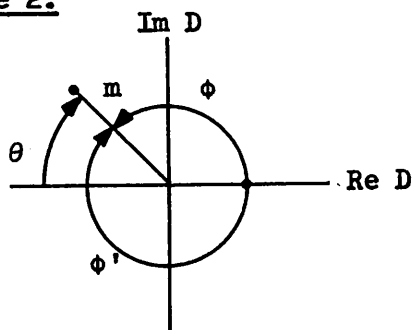
Case 1:

$$\text{Re } D = m \cdot \cos (\phi)$$

$$\text{Im } D = m \cdot \sin (\phi)$$

$$0 \leq \phi \leq 90^\circ$$

$$\text{or } -360^\circ \leq \phi' \leq -270^\circ$$

Case 2:

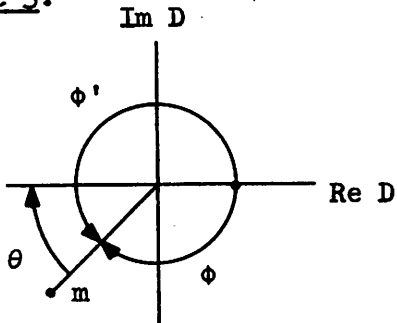
$$\text{Re } D = -m \cdot \cos (\theta)$$

$$\text{Im } D = m \cdot \sin (\theta)$$

$$\theta \geq 0$$

$$90^\circ \leq \phi \leq 180$$

$$\text{or } -270^\circ \leq \phi' \leq -180^\circ$$

Case 3:

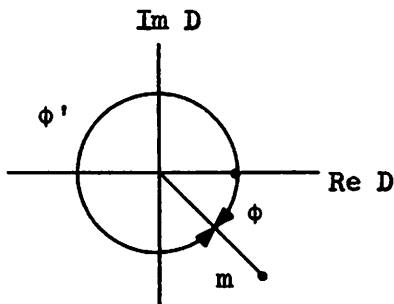
$$\text{Re } D = -m \cdot \cos(\theta)$$

$$\text{Im } D = -m \cdot \sin(\theta)$$

$$\theta \cong 0$$

$$180^\circ \cong \phi' \cong 270^\circ$$

$$\text{or } -180^\circ \cong \phi \cong -90^\circ$$

Case 4:

$$\text{Re } D = m \cdot \cos(\phi)$$

$$\text{Im } D = -m \cdot \sin(\phi)$$

$$270^\circ \cong \phi' \cong 360^\circ$$

$$\text{or } -90^\circ \cong \phi \cong 0^\circ$$

Independent source signals are read in in terms of magnitude and phase, for the sake of convenience. Subroutine VOLCUR transforms these signals into regular complex signals, which CANDOFD requires for execution.

Subroutine INITIAL:

This subroutine initializes the tree branch voltages and link currents of a given network. The independent source signals are set to their appropriate complex values, and the remaining signals are set equal to complex zeros. When $\omega = 0$, the independent source phases are ignored, and the real parts of the independence source signals are set equal to the signal magnitudes.

Subroutine CALCAL:

This subroutine computes the complex tree branch currents \underline{i}_b and the complex link voltages \underline{v}_l from the complex link currents \underline{i}_l and the complex tree voltages \underline{v}_t respectively, via

$$\underline{i}_t = - \underline{F} \underline{i}_l$$

and

$$\underline{v}_l = \underline{F}' \underline{v}_t$$

where \underline{F} is the nontrivial part of the fundamental (and real) cutset matrix (see subroutine FCSM), and where the prime denotes the transpose.

Thus we see that Kirchhoff's voltage and current laws are automatically satisfied, for any given connected network, by forcing \underline{i}_t and \underline{v}_l to satisfy the above relations.

SUBROUTINE KONST:

This subroutine computes the (complex) branch element constants required in the branch relations (see Subroutine ERROR). When desired (NSCALE \neq 1), KONST selects a new current scale factor each time a new topological tree is computed. There are no constants associated with independent sources.

The complex constants are computed according to the prescription given below, where ω is the source frequency, S denotes the current scale factor (e.g., S = 1000 \Rightarrow currents are in milliamps) and λ_i denotes the constant associated with the ith branch.

Dependent voltage source:

a) Voltage controlled case

$$\operatorname{Re}(\lambda_i) = K_i$$

$$\operatorname{Im}(\lambda_i) = 0$$

where K_i is the ith coupling constant and is unitless.

b) Current controlled case

$$\operatorname{Re}(\lambda_i) = K_i/S$$

$$\operatorname{Im}(\lambda_i) = 0$$

where K_i is the ith coupling constant, in ohms.

Capacitances:

a) Tree branch case

$$\operatorname{Re}(\lambda_i) = 0$$

$$\text{Im}(\lambda_i) = \begin{cases} -\frac{1}{\omega * S * C_i} & \omega \neq 0 \\ \text{not defined} & \omega = 0 \end{cases}$$

b) Link case

$$\text{Re}(\lambda_i) = 0$$

$$\text{Im}(\lambda_i) = \omega * C_i * S$$

where C_i is the i th capacitance value, in farads,

Resistances:

a) Tree branch case

$$\text{Re}(\lambda_i) = R_i / S$$

$$\text{Im}(\lambda_i) = 0$$

b) Link case

$$\text{Re}(\lambda_i) = S / R_i$$

$$\text{Im}(\lambda_i) = 0$$

where R_i is the i th resistance value in ohms.

Inductances:

a) Tree branch case

$$\text{Re}(\lambda_i) = 0$$

$$\text{Im}(\lambda_i) = \frac{\omega * L_i}{S}$$

b) Link case

$$\text{Re}(\lambda_i) = 0$$

$$\text{Im}(\lambda_i) = \begin{cases} -\frac{S}{\omega * L_i} & \omega \neq 0 \\ \text{not defined} & \omega = 0 \end{cases}$$

where L_i is the i th inductance value in henrys.

Dependent current sources:

a) Voltage controlled case

$$\operatorname{Re}(\lambda_i) = K_i * S$$

$$\operatorname{Im}(\lambda_i) = 0$$

where K_i is the i th coupling constant, in ohms.

b) Current controlled case

$$\operatorname{Re}(\lambda_i) = K_i$$

$$\operatorname{Im}(\lambda_i) = 0$$

where K_i is the i th coupling constant, and is unitless.

It should be noted that no zero valued inductances can be links and no zero valued capacitances can be tree branches (see comment 8, Appendix E).

Automatic computation of scale factor:

When desired, subroutine KONST will compute a new current scale factor each time a new topological tree is selected.

The algorithm, for selecting the optimum scale factor, is as follows:

1. Select smallest link impedance magnitude (say $|Z_i|$).
2. Set internal current unit to be $\frac{\text{amps}}{|Z_i|}$, where $|Z_i|$ is in ohms, e.g., if $|Z_i| = 14800$ ohms, the internal current unit becomes $\frac{1}{14.8}$ milliamps.

If our optimal tree algorithm yields a topology which contains only link current sources, the scale factor is automatically set equal to 1.0.

SUBROUTINE ERROR:

This subroutine computes the complex branch errors E_i , and the performance criterion ϵ , defined by

$$\epsilon = \frac{1}{2} \sum_i E_i \cdot E_i^*$$

where i ranges over all non-independent source branches and $*$ designates the complex conjugate.

ϵ is to be subsequently minimized and, when both the branch relations and Kirchhoff's laws are simultaneously satisfied, ϵ will be identically zero.

Complex branch errors:Notation:

The λ_i denote the complex branch element constants computed in subroutine KONST. V_i and i_i denote the complex branch voltage and current respectively. The E_i , of course, denote the complex branch error associated with the i th branch.

Dependent voltage sources:

a) Voltage controlled case

$$E_i = -V_i + \lambda_i * V_m$$

b) Current controlled case

$$E_i = -V_i + \lambda_i * i_m$$

where m denotes the controlling branch.

Capacitances:

a) Tree branch case.

$$E_i = -V_i + \lambda_i * i_i \quad \omega \neq 0$$

$$E_i = \text{complex zero} \quad \omega = 0$$

b) Link case

$$E_i = -i_i + \lambda_i * V_i$$

Resistances:

a) Tree branch case

$$E_i = -V_i + \lambda_i * i_i$$

b) Link case

$$E_i = -i_i + \lambda_i * V_i$$

Inductances:

a) Tree branch case

$$E_i = -V_i + \lambda_i * i_i$$

b) Link case

$$E_i = -i_i + \lambda_i * V_i \quad \omega \neq 0$$

$$E_i = \text{complex zero} \quad \omega = 0$$

Dependent current sources:

a) Voltage controlled case

$$E_i = -i_i + \lambda_i * V_m$$

b) Current controlled case

$$E_i = -i_i + \lambda_i * i_m$$

where m denotes the controlling branch.

SUBROUTINE CNGVARS:

Given the complex direction along which we minimized, \tilde{S}^i , and the (real) optimal step size α^i , the complex non independent source tree voltages and link currents are modified as follows:

$$\begin{bmatrix} \hat{v}_t \\ \hat{i}_\ell \end{bmatrix}_{\text{new}} = \begin{bmatrix} \hat{v}_t \\ \hat{i}_\ell \end{bmatrix}_{\text{old}} - \alpha^i * \tilde{S}^i$$

It should be emphasized that only the basis set of variables, i.e., the non independent source tree voltages and link currents, are changed. \hat{v}_t and \hat{i}_ℓ designate, in this context, non-independent source signals. Tree currents and link voltages follow directly from the implicit formulation of Kirchhoff's laws (see subroutine CALCAL).

SUBROUTINE FNDGRAD:

This subroutine computes the complex gradient vector of the performance criterion, with respect to non-independent source tree voltages and non-independent source link currents.

Theory:a) Tree voltage gradient vector

$$\text{define } \operatorname{Re}(\underline{g}_t) = \frac{d\varepsilon}{d[\operatorname{Re}\underline{v}_t]} \quad (1)$$

$$\text{and } \operatorname{Im}(\underline{g}_t) = \frac{d\varepsilon}{d[\operatorname{Im}\underline{v}_t]}$$

where \underline{g}_t is the tree voltage complex gradient vector.

Recalling that

$$\underline{v}_b = \begin{bmatrix} \underline{z}_1 \\ \vdots \\ \underline{z}_N \end{bmatrix} \underline{v}_t \quad (2)$$

the tree voltage gradient vector \underline{g}_t can be expressed, via the chain rule, as

$$\underline{g}_t = \begin{bmatrix} \frac{\partial \underline{v}_b}{\partial \underline{v}_t} \end{bmatrix} \cdot \frac{\partial \varepsilon}{\partial \underline{v}_b} \quad (3)$$

where

$$\frac{\partial \underline{v}_b}{\partial \underline{v}_t} = \begin{bmatrix} \frac{\partial v_{b_1}}{\partial v_{t_1}} & \frac{\partial v_{b_2}}{\partial v_{t_1}} & \dots & \frac{\partial v_{b_{NB}}}{\partial v_{t_1}} \\ \dots & \dots & \dots & \dots \\ \frac{\partial v_{b_1}}{\partial v_{t_{NN-1}}} & \dots & \dots & \frac{\partial v_{b_{NB}}}{\partial v_{t_{NN-1}}} \end{bmatrix} \quad (4)$$

also

$$\left. \begin{aligned} \text{Re} \left[\frac{\partial \epsilon}{\partial \tilde{v}_b} \right] &= \frac{\partial \epsilon}{\partial [\text{Re} \tilde{v}_b]} \\ \text{and} \quad \text{Im} \left[\frac{\partial \epsilon}{\partial \tilde{v}_b} \right] &= \frac{\partial \epsilon}{\partial [\text{Im} \tilde{v}_b]} \end{aligned} \right\} \quad (5)$$

from (2), it is immediately apparent that

$$\frac{\partial \tilde{v}_b}{\partial \tilde{v}_t} = \left[\begin{array}{c} \tilde{I} \\ \tilde{F}' \end{array} \right] \quad (6)$$

Now, $\frac{\partial \epsilon}{\partial \tilde{v}_b}$ may be partitioned as follows:

$$\frac{\partial \epsilon}{\partial \tilde{v}_b} = \left[\begin{array}{c} \frac{\partial \epsilon}{\partial \tilde{v}_t} \\ \text{---} \\ \frac{\partial \epsilon}{\partial \tilde{v}_\ell} \end{array} \right] \quad (7)$$

Combining (3), (6) and (7) we obtain

$$\tilde{g}_t = [\tilde{I} : \tilde{F}'] \cdot \left[\begin{array}{c} \frac{\partial \epsilon}{\partial \tilde{v}_t} \\ \text{---} \\ \frac{\partial \epsilon}{\partial \tilde{v}_\ell} \end{array} \right] \quad (8)$$

Equation (8) immediately reduces to the desired form

$$\tilde{g}_t = \frac{d\epsilon}{d\tilde{v}_t} = \frac{\partial \epsilon}{\partial \tilde{v}_t} + \tilde{F}' \cdot \frac{\partial \epsilon}{\partial \tilde{v}_\ell} \quad (9)$$

where $\operatorname{Re} \left[\frac{\partial \epsilon}{\partial \tilde{v}_t} \right] = \frac{\partial \epsilon}{\partial [\operatorname{Re} \tilde{v}_t]}$,

$$\operatorname{Im} \left[\frac{\partial \epsilon}{\partial \tilde{v}_t} \right] = \frac{\partial \epsilon}{\partial [\operatorname{Im} \tilde{v}_t]} ,$$

$$\operatorname{Re} \left[\frac{\partial \epsilon}{\partial \tilde{v}_\ell} \right] = \frac{\partial \epsilon}{\partial [\operatorname{Re} \tilde{v}_\ell]} ,$$

and $\operatorname{Im} \left[\frac{\partial \epsilon}{\partial \tilde{v}_\ell} \right] = \frac{\partial \epsilon}{\partial [\operatorname{Im} \tilde{v}_\ell]}$.

b) Link current gradient vector.

Since

$$\tilde{i}_b = \begin{bmatrix} -\tilde{z}_F \\ \tilde{z}_I \end{bmatrix} \cdot \tilde{i}_\ell \quad (10)$$

we may proceed exactly as in the tree voltage gradient case, and arrive at the desired form;

$$\tilde{g}_\ell = \frac{d\epsilon}{d\tilde{i}_\ell} = \frac{\partial \epsilon}{\partial \tilde{i}_\ell} - \tilde{F}' \cdot \frac{\partial \epsilon}{\partial \tilde{i}_t} \quad (11)$$

where \tilde{g}_ℓ , $\frac{d\epsilon}{d\tilde{i}_\ell}$, $\frac{\partial \epsilon}{\partial \tilde{i}_\ell}$ and $\frac{\partial \epsilon}{\partial \tilde{i}_t}$ have analogous meanings to their counterparts in the tree voltage gradient case.

c) Branch gradient vectors.

The branch gradients $\frac{\partial \epsilon}{\partial \tilde{i}_b}$ and $\frac{\partial \epsilon}{\partial \tilde{v}_b}$ are computed as follows; (where $\frac{\partial \epsilon}{\partial \tilde{v}_i}$ and $\frac{\partial \epsilon}{\partial \tilde{i}_i}$ are branch gradients associated with the i th branch):

Tree branches:

Independent voltage sources:

a) not associated with controlled sources $\frac{\partial \epsilon}{\partial \tilde{v}_i} = \frac{\partial \epsilon}{\partial \tilde{i}_i} = \text{complex zero}$

b) associated with controlled source m

$$\frac{\partial \epsilon}{\partial v_i} = \text{complex zero}$$

$$\frac{\partial \epsilon}{\partial i_i} = \lambda_m * E_m$$

- dependent voltage sources:

$$\frac{\partial \epsilon}{\partial v_i} = - E_i$$

$$\frac{\partial \epsilon}{\partial i_i} = \text{complex zero}$$

- dependent current sources:

$$\frac{\partial \epsilon}{\partial v_i} = \text{complex zero}$$

$$\frac{\partial \epsilon}{\partial i_i} = - E_i$$

- tree capacitances at $\omega = 0$:

$$\frac{\partial \epsilon}{\partial v_i} = \frac{\partial \epsilon}{\partial i_i} = \text{complex zero}$$

- remaining tree branches:

$$\frac{\partial \epsilon}{\partial v_i} = - E_i$$

$$\text{Re} \left[\frac{\partial \epsilon}{\partial i_i} \right] = \text{Re}(\lambda_i) * \text{Re}(E_i) + \text{Im}(\lambda_i) * \text{Im}(E_i)$$

$$\text{Im} \left[\frac{\partial \epsilon}{\partial i_i} \right] = - \text{Im}(\lambda_i) * \text{Re}(E_i) + \text{Re}(\lambda_i) * \text{Im}(E_i)$$

Link branches:

- Independent current sources:

- a) not associated with controlled sources

$$\frac{\partial \epsilon}{\partial v_i} = \frac{\partial \epsilon}{\partial i_i} = \text{complex zero}$$

- b) associated with controlled source m

$$\frac{\partial \epsilon}{\partial v_i} = \lambda_m * E_m$$

$$\frac{\partial \epsilon}{\partial i_i} = \text{complex zero}$$

- dependent current sources:

$$\frac{\partial \epsilon}{\partial v_i} = \text{complex zero}$$

$$\frac{\partial \epsilon}{\partial i_i} = - E_i$$

- dependent voltage sources:

$$\frac{\partial \epsilon}{\partial v_i} = - E_i$$

$$\frac{\partial \epsilon}{\partial i_i} = \text{complex zero}$$

- link inductance at $\omega = 0$:

$$\frac{\partial \epsilon}{\partial v_i} = \frac{\partial \epsilon}{\partial i_i} = \text{complex zero}$$

- remaining links

$$\text{Re} \left[\frac{\partial \epsilon}{\partial v_i} \right] = \text{Re}(\lambda_i) * \text{Re}(E_i) + \text{Im}(\lambda_i) * \text{Im}(E_i)$$

$$\text{Im} \left[\frac{\partial \epsilon}{\partial v_i} \right] = - \text{Im}(\lambda_i) * \text{Re}(E_i) + \text{Re}(\lambda_i) * \text{Im}(E_i)$$

$$\frac{\partial \epsilon}{\partial i_i} = - E_i$$

Hence, knowing $\frac{\partial \epsilon}{\partial i_b}$ and $\frac{\partial \epsilon}{\partial v_b}$, $\frac{d\epsilon}{dv_t}$ and $\frac{d\epsilon}{di_{\ell}}$ follow directly from equations (9) and (11). Note that, at $\omega = 0$, $\frac{d\epsilon}{dv}$ for tree capacitances and $\frac{d\epsilon}{di}$ for link inductances must be constrained to be complex zero.

SUBROUTINE FNDALPH:

This subroutine computes α^i , the optimal step size at the i th iterative step, which minimizes ϵ along $\underline{\hat{S}}^i$, a (complex) direction computed by subroutine FLPOW.

The constant α^i is such that

$$\min_{\lambda} \epsilon(\underline{\hat{x}}^i - \lambda \cdot \underline{\hat{S}}^i) = (\underline{\hat{x}}^i - \alpha^i \cdot \underline{\hat{S}}^i)$$

where $\underline{\hat{x}}$ is the set of (complex) nonindependent source tree voltages and link currents, at the i th iterative step.

The analytical expression for α^i is:

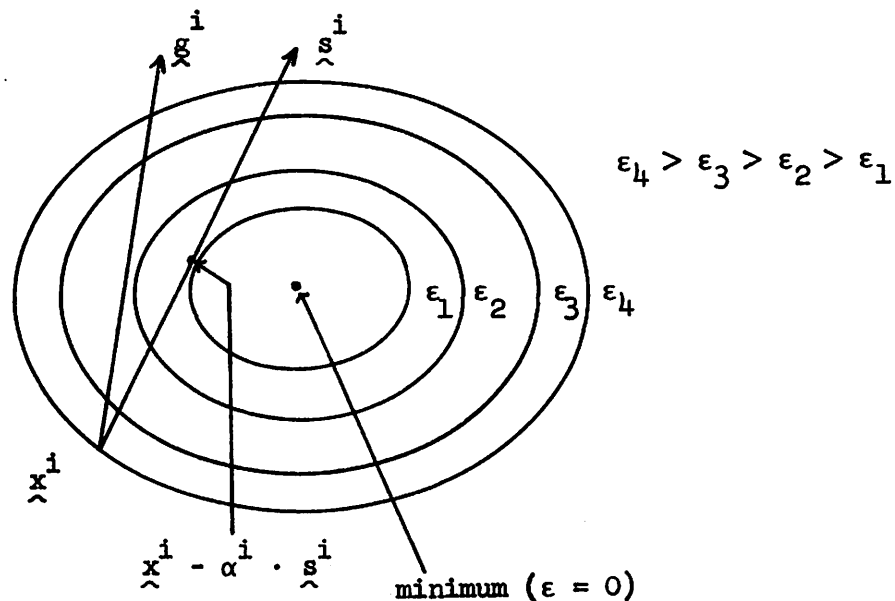
$$\alpha^i = \frac{1}{2} \frac{A}{\epsilon(\underline{\hat{x}} - \underline{\hat{S}}^i) + A - \epsilon(\underline{\hat{x}})},$$

and

$$A = \sum_k [\operatorname{Re}(\underline{g}_k^i) \cdot \operatorname{Re}(\underline{S}_k^i) + \operatorname{Im}(\underline{g}_k^i) \cdot \operatorname{Im}(\underline{S}_k^i)]$$

where \underline{g}^i is the complex gradient at the i th iterative step, and k ranges over all the nonindependent source branches.

The situation may be graphically depicted as follows:



SUBROUTINE FLPOW:

This subroutine computes the complex direction $\underline{\hat{S}}$ along which we wish to minimize the performance criterion ϵ .

The Fletcher-Powell⁽²⁾ algorithm is utilized with the initial direction being that of steepest descent (i.e., along the gradient).

Fletcher-Powell algorithm:

Let $\underline{\hat{H}}$ be a positive definite symmetric matrix. The dimension of $\underline{\hat{H}}$ is $N \times N$ where

$$N = 2 \times (\text{NB} - \text{number of independent sources})$$

The direction along which we perform the minimization is given by

$$\underline{\hat{y}}^i = \underline{\hat{H}}^i \cdot \underline{\hat{\omega}}^i ,$$

where the superscript i indicates that we are at the i th iterative step, and $\underline{\hat{\omega}}^i$ is defined as follows:

$$\underline{\hat{\omega}}^i = \begin{bmatrix} \text{Re}(\underline{\hat{g}}_t) \\ \text{Re}(\underline{\hat{g}}_\ell) \\ \text{Im}(\underline{\hat{g}}_t) \\ \text{Im}(\underline{\hat{g}}_\ell) \end{bmatrix}$$

where $\underline{\hat{g}}_t$ and $\underline{\hat{g}}_\ell$ are defined in subroutine FNDGRAD.

The $\underline{\hat{H}}$ matrix is updated as follows:

$$\underline{\hat{H}}^1 = \underline{\hat{I}} \text{ (Identity matrix)}$$

$$\underline{\hat{H}}^{i+1} = \underline{\hat{H}}^i - \frac{\underline{\hat{y}}^i \cdot (\underline{\hat{y}}^i)'}{(\underline{\hat{y}}^i)' \cdot (\underline{\hat{\omega}}^{i+1} - \underline{\hat{\omega}}^i)} - \frac{[\underline{\hat{H}}^i \cdot (\underline{\hat{\omega}}^{i+1} - \underline{\hat{\omega}}^i)] \cdot [\underline{\hat{H}}^i \cdot (\underline{\hat{\omega}}^{i+1} - \underline{\hat{\omega}}^i)]'}{(\underline{\hat{\omega}}^{i+1} - \underline{\hat{\omega}}^i)' \cdot \underline{\hat{H}}^i \cdot (\underline{\hat{\omega}}^{i+1} - \underline{\hat{\omega}}^i)}$$

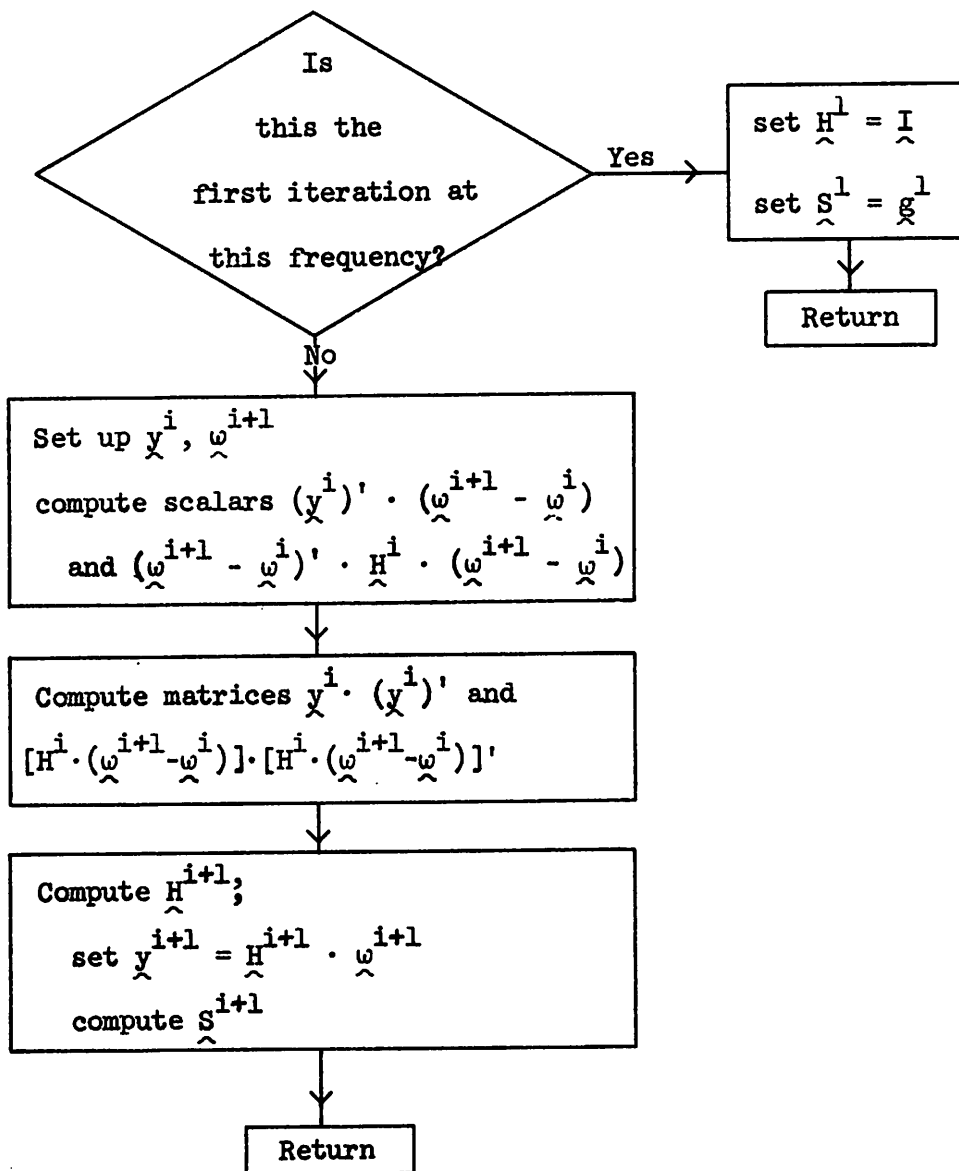
Then,

$$\underline{y}^{i+1} = \underline{H}^{i+1} \cdot \underline{s}^{i+1}$$

and, we identify

$$\begin{bmatrix} \text{Re} (\underline{S}^{i+1}) \\ \text{-----} \\ \text{Im} (\underline{S}^{i+1}) \end{bmatrix} = \underline{y}^{i+1}$$

The \underline{H} matrix converges to the inverse of the second derivative matrix, known as the Hessian matrix.

Block DiagramSUBROUTINE FLPOW

SUBROUTINE READOUT:

This subroutine outputs all the desired branch current and/or voltage magnitudes and phases, in the desired order, and in the original network numbering scheme. Hence for any branch, one may output its current and/or its voltage.

If the frequency at which the network was analyzed had to be modified due to the presence of an "ideal resonance," a diagnostic is provided.

Note that a one to one correspondence, between desired output variables and their internal ordering, needs be calculated each time a new tree is selected. This is done in subroutine PT.

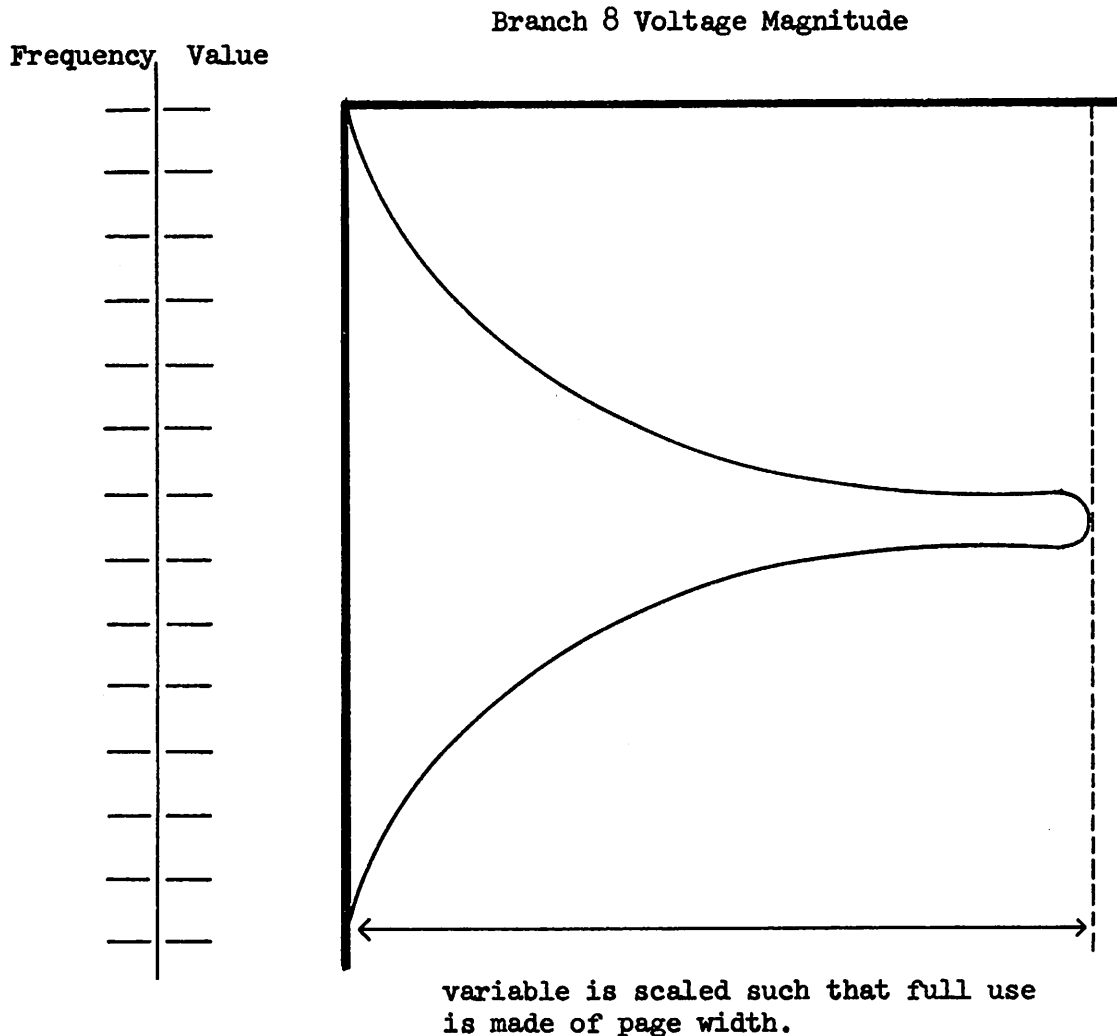
Subroutine READOUT converts the desired output variables from their internal complex representation, to a more desirable (magnitude, phase) representation.

SUBROUTINE GRAPH:

The purpose of this subroutine is to provide graphical outputs of up to five signals. A graphical output of any branch voltage or current, magnitude or phase, can be obtained.

Subroutine GRAPH is called, by the main program, at each frequency point. The desired voltages and currents are stored in an array. When either the final frequency point or the 200th frequency point is reached, the desired magnitudes and phases are computed. The variables are then scaled individually so that maximum use is made of the output page width.

The following figure illustrates a typical graph output.



SUBROUTINE ALLOUT:

This subroutine outputs all the tree voltages and link currents (including those associated with independent sources), in the optimal tree numbering scheme.

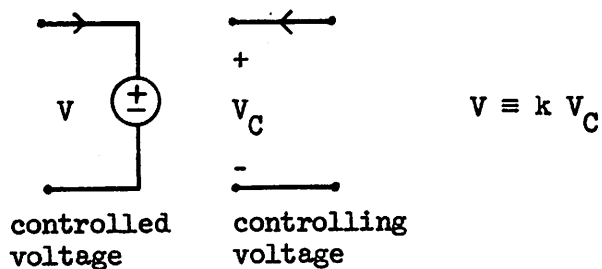
Hence, when ALLOUT is used, one should also output the corresponding tree information, which will immediately yield the isomorphism between the original network and the new topologies.

APPENDIX A:Dependent source modeling:

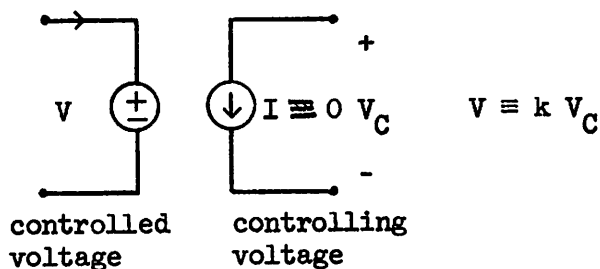
Each dependent source requires two branches for its complete specification, both of which must be a part of the network description, and hence must be read in as data.

Voltage controlled voltage source:

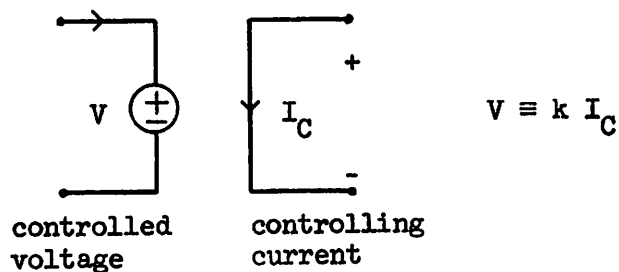
Ideal model:



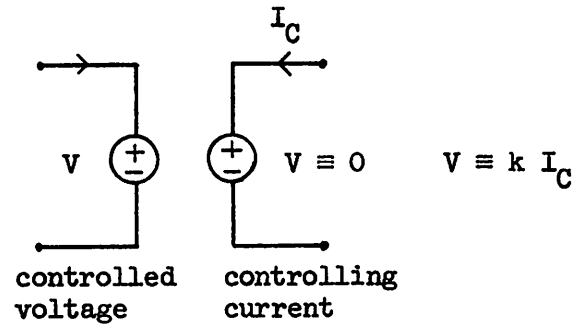
CANDOFD model:

Current controlled voltage source:

Ideal model:

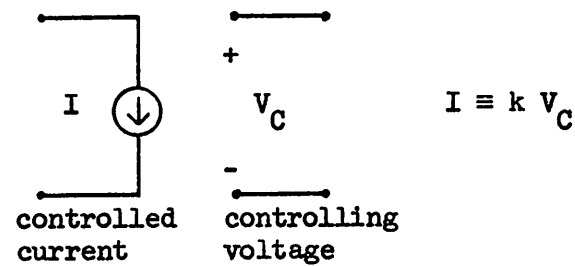


CANDOFD model:

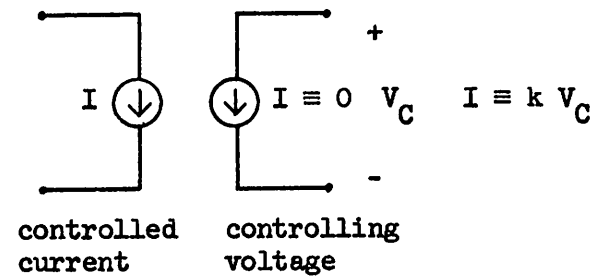


Voltage controlled current source:

Ideal model:

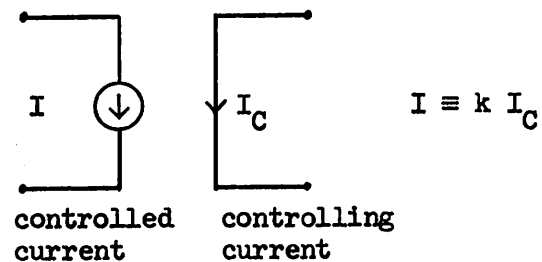


CANDOFD model:

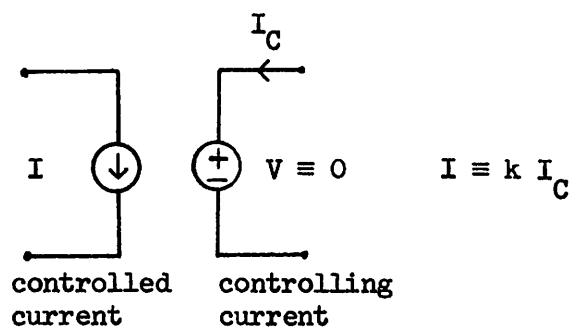


Current controlled current source:

Ideal model:



CANDOFD model:



Thus we see that in all cases the controlling current is taken to be the current through a zero-valued voltage source, and the controlling voltage is taken to be the voltage across a zero-valued current source.

The coupling constants, denoted by k , are those that should be read in as corresponding elements of the VALUE array.

APPENDIX B:Data cards for CANDOFDNotation:

I ⇒ integer format

A ⇒ alphanumeric format

E ⇒ exponential or floating point format

col ⇒ column on data card

Card #1:

Variables read in, in sequential order

NN, number of nodes (I)

NB, number of branches (I)

WSTART, initial frequency (E)

WEND, final frequency (E)

FACTOR, frequency multiplicative factor, if desired, (E)

EPS, error below which performance criterion must fall (E)

NCONT, Tree and \hat{F} output control (I)

NCONT = 1 \Rightarrow outputs desired

NCONT = 0 \Rightarrow outputs not desired

NITREE, Tree picking control (I)

NITREE = 1 \Rightarrow pick a new tree at each frequency point

NITREE = 0 \Rightarrow pick a tree only at the initial frequency point

col 1-5, - NN

col 6-10, - NB

col 11-25, - WSTART

col 26-40, - WEND

col 41-55, - FACTOR

col 56-70, - EPS

col 75, - NCONT

col 80, - NITREE

DATA CARD #1

Card #2:

Variables read in, in sequential order

NGRAPH, number of graphical outputs (I)

NALLOUT, control variable for use of

ALLOUT subroutine (I)

NALLOUT = 1 \Rightarrow use of ALLOUT is desired

NALLOUT = 0 \Rightarrow use of ALLOUT is not desired

JOUT, number of outputs desired (I)

To be used in conjunction with subroutine READOUT

SCALE, scale factor (E)

e.g., scale factor of 10^3 sets current unit to milliamps

NITT, number of iterative minimization steps, per frequency point (I)

If only an error control is desirable, set NITT = 0

FREQADD, frequency additive step, if desired (E)

NSCALE, Control variable for use of internal, automatic current scaling

NSCALE = 1 \Rightarrow use scale factor read in as SCALE

NSCALE \neq 1 \Rightarrow use internal, automatic current scaling algorithm

col 1-5, - NGRAPH
 col 6-10, - NALLOUT
 col 11-15, - JOUT
 col 16-30, - SCALE
 col 31-35, - NITT
 col 36-50, - FREQADD
 col 51-55, - NSCALE
 col 56-60, - IDRES

DATA CARD #2

IDRES, Control variable for use of internal ideal resonance testing.

IDRES = 1 \Rightarrow bypass ideal resonance tests

IDRES \neq 1 \Rightarrow test for ideal resonances

Network description data cards:

For each network branch, the following card is needed. The order in which network branches are read in is arbitrary.

Variables read in, in sequential order

TYPE, Branch type (A)

E - independent voltage source

V - controlled voltage source

C - capacitance

L - inductance

R - resistance

I - controlled current source

J - independent current source

IBRAN, Branch number (I)

CONTYPE, dependent source controlling type (A)

V - voltage controlled

I - current controlled

KONBRAN, dependent source controlling branch (I)

LEAV, node which branch leaves (I)

LENT, node which branch enters (I)

VALUE, value of branch element (E)

- Resistance in ohms

- Inductance in henrys

- Capacitance in farads

- Dependent source coupling constants in ohms, mhos, or unitless

(See Appendix A)

AMAG, independent source signal magnitude (E)

in volts, for independent voltage source

in amperes, for independent current source

APHASE, independent source signal phase, in degrees (E)

- $360^\circ \leq \text{phase} \leq +360$

col 1, - TYPE

col 2-4, - IBRAN

col 6, - CONTYPE

col 7-9, - KONBRAN

col 11-12, - LEAV

col 14-15, - LENT

col 21-35, - VALUE

col 36-50, - AMAG

col 51-65, - APHASE

DATA CARD DESCRIBING

ONE NETWORK BRANCH

OUTPUT specifications:

Only one type of output is allowed for any one network analysis, i.e., only one of NGRAPH, NALLOUT and JOUPT can be nonzero.

The following data cards follow immediately after the network description data cards.

- If NGRAPH \neq 0, we need NGRAPH (\leq 5) data cards with the following information

col 1-5, branch number (I)

col 10, branch signal desired (I)

1 \Rightarrow current desired

0 \Rightarrow voltage desired

col 15, signal information desired (I)

1 \Rightarrow phase plot desired

0 \Rightarrow magnitude plot desired

- If JOUT \neq 0, we need JOUT (\leq 200) data cards with the following information

col 1-5, branch number (I)

col 10, branch signal desired (I)

1 \Rightarrow current desired

0 \Rightarrow voltage desired

- If NALLOUT \neq 0, no other data cards are required.

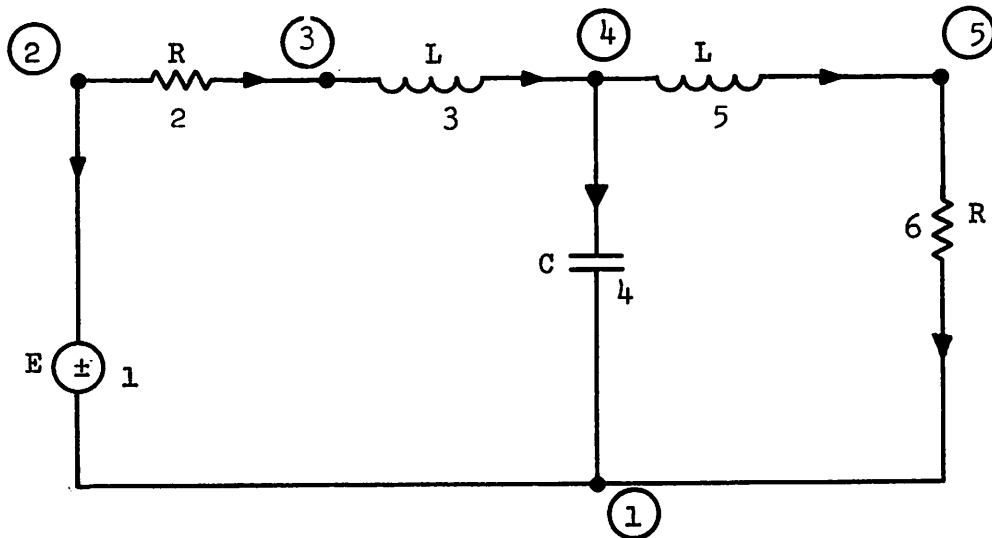
Note that when one or more networks are analyzed in one batch, the last data card of the batch should be a blank card.

APPENDIX C:Sample Problems:Notation:

- circled numbers indicate node numbers
- uncircled numbers indicate branch numbers
- R's indicate resistances, in ohms
- L's indicate inductances, in henrys
- C's indicate capacitances, in farads
- NN is the number of nodes
- NB is the number of branches
- WSTART is the initial frequency in radians/sec, (WSTART = 0
implies a DC solution is desired)
- WEND is the final frequency of concern (in radians/sec)
- FACTOR is the multiplicative frequency factor (when desired)
e.g., if ω_1 is the present frequency point, the next frequency
point will be given by $\omega_1 * \text{FACTOR}$
- FREQADD is the additive frequency increment (when desired)
e.g., if ω_1 is the present frequency point, the next frequency
point is given by $\omega_1 + \text{FREQADD}$
- EPS is the desired performance criterion error
- The central processor time, given for the sample problems, includes
the input-output execution time.
- Only the pertinent computer output is reproduced with the sample
problems .

Sample Problem #1:

Low Pass Butterworth Filter:



$$NN = 5$$

$$NB = 6$$

$$R = 10^3$$

$$L = 0.01$$

$$C = 2 \times 10^{-8}$$

$$|E| = 10 \text{ V}, \quad \angle E = 0.0^\circ$$

Specifications:

- Initial frequency = 10^3 (WSTART = 10^3)
- Final frequency = 5×10^5 (WEND = 5×10^5)
- Frequency additive factor = 10^4 (FREQADD = 10^4)
- Desire error to fall below 10^{-5} or a maximum of 18 iterative minimization steps at each frequency point.

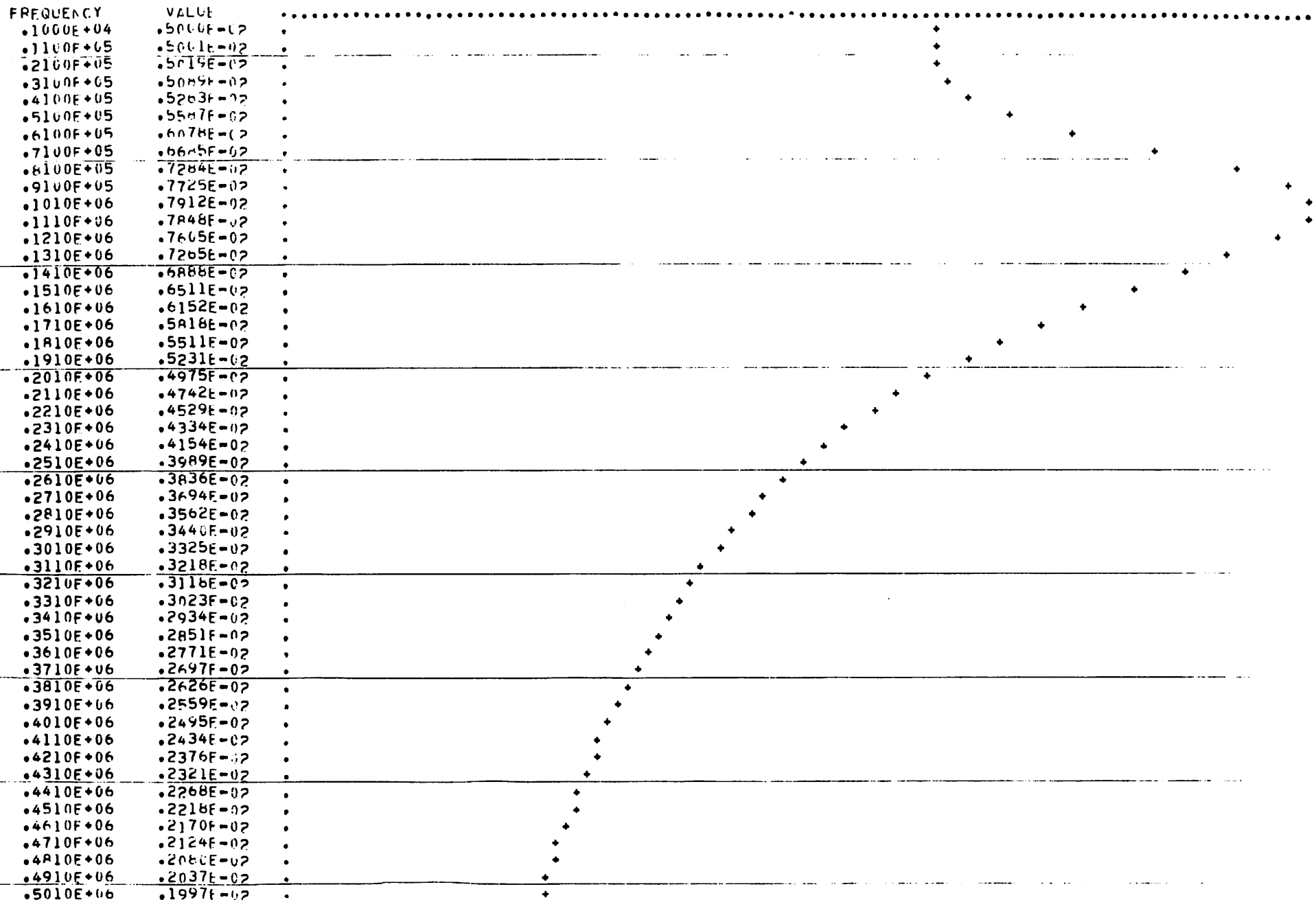
$$(EPS = 10^{-5}, \text{ NITR} = 18)$$

- Desire tree and \underline{F} to be outputted (NCONT = 1)
- Desire internal automatic current scaling. (SCALE is arbitrary,
NSCALE \neq 1)
- Desire the following graphical outputs
 - Branch 4 current magnitude
 - Branch 1 current magnitude
 - Branch 1 current phase
 - Branch 6 voltage magnitude
 - Branch 6 voltage phase
- Desire an optimal tree to be selected at each frequency (NTREE = 1)

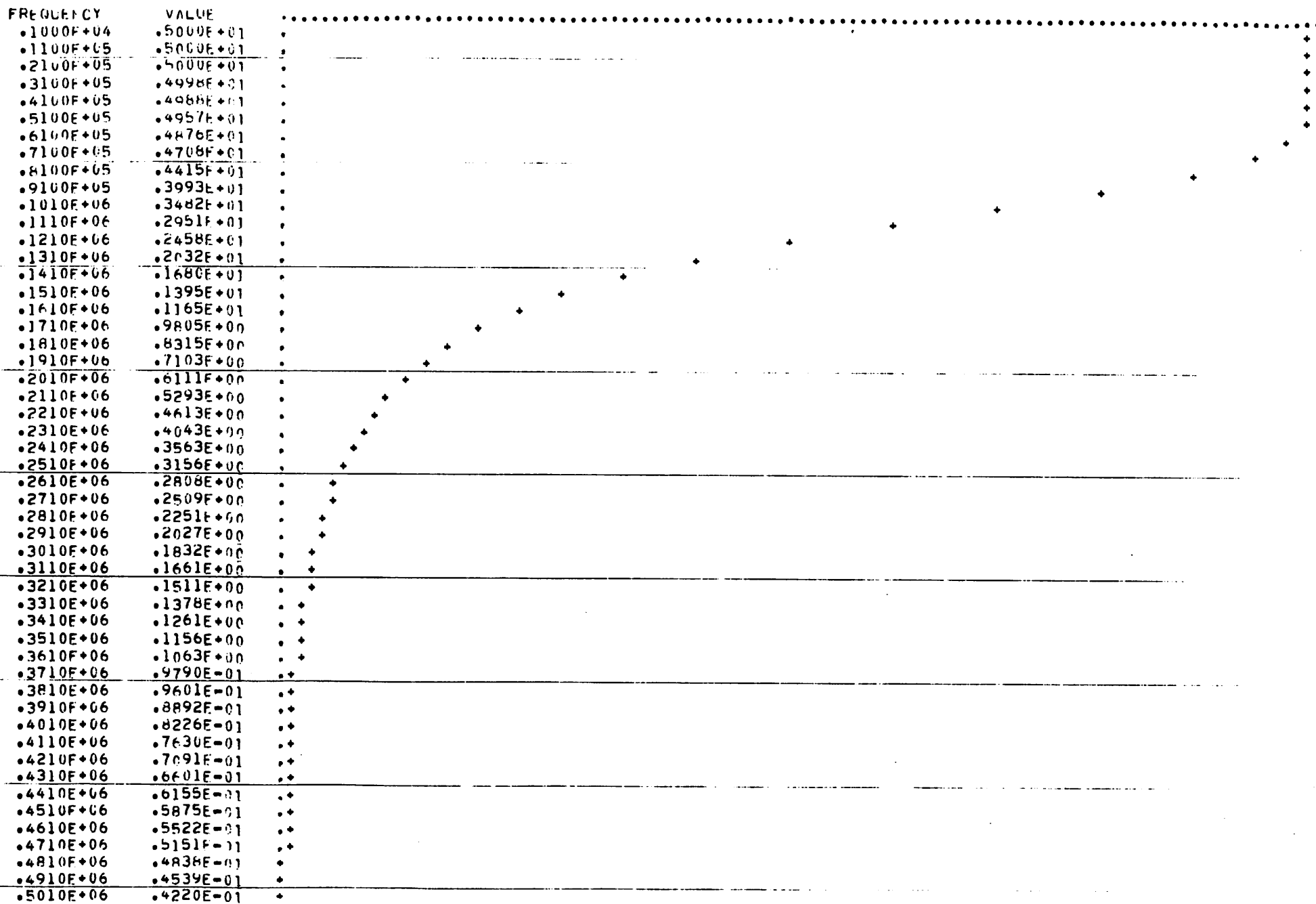
(NGRAPH = 5)

The required data cards for this problem may be seen on a subsequent page. The central processor time, for this problem was 9.86 seconds.

BRANCH 1 CURRENT MAGNITUDE PLOT

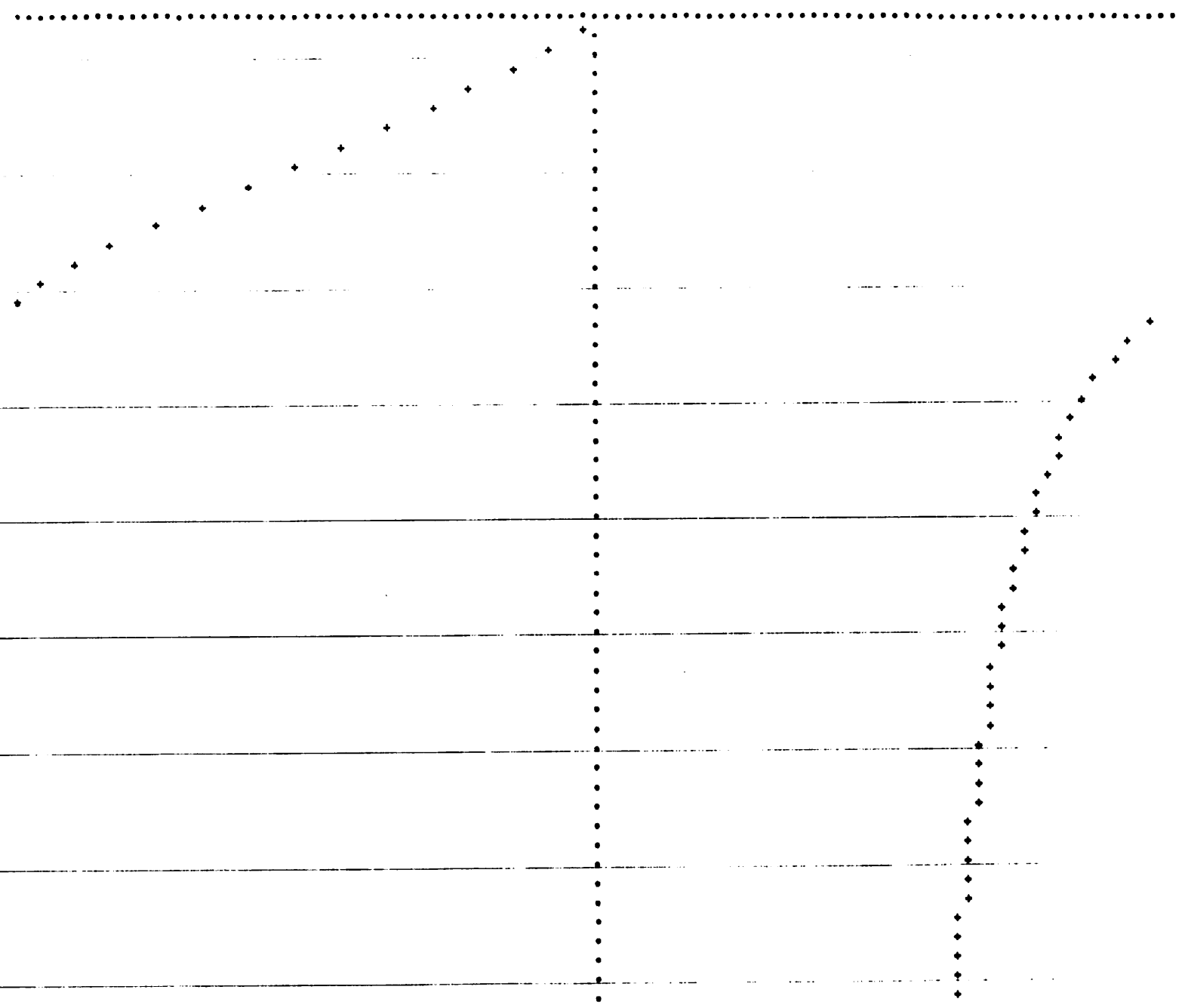


BRANCH 6 VOLTAGE MAGNITUDE PLOT



BRANCH 6 VOLTAGE PHASE PLOT

FREQUENCY	VALUE
.1000E+04	-.1175E+01
.1100E+05	-.1203E+02
.2100E+05	-.2425E+02
.3100E+05	-.3615E+02
.4100E+05	-.4853E+02
.5100E+05	-.6160E+02
.6100E+05	-.7555E+02
.7100E+05	-.9044E+02
.8100E+05	-.1060E+03
.9100E+05	-.1216E+03
.1010E+06	-.1364E+03
.1110E+06	-.1498E+03
.1210E+06	-.1614E+03
.1310E+06	-.1713E+03
.1410E+06	-.1797E+03
.1510E+06	.1732E+03
.1610E+06	.1672E+03
.1710E+06	.1619E+03
.1810E+06	.1574E+03
.1910E+06	.1534E+03
.2010E+06	.1499E+03
.2110E+06	.1468E+03
.2210E+06	.1440E+03
.2310E+06	.1415E+03
.2410E+06	.1392E+03
.2510E+06	.1371E+03
.2610E+06	.1351E+03
.2710E+06	.1334E+03
.2810E+06	.1318E+03
.2910E+06	.1303E+03
.3010E+06	.1289E+03
.3110E+06	.1276E+03
.3210E+06	.1263E+03
.3310E+06	.1252E+03
.3410E+06	.1241E+03
.3510E+06	.1231E+03
.3610E+06	.1222E+03
.3710E+06	.1213E+03
.3810E+06	.1202E+03
.3910E+06	.1194E+03
.4010E+06	.1187E+03
.4110E+06	.1180E+03
.4210E+06	.1173E+03
.4310E+06	.1166E+03
.4410E+06	.1160E+03
.4510E+06	.1150E+03
.4610E+06	.1146E+03
.4710E+06	.1140E+03
.4810E+06	.1136E+03
.4910E+06	.1131E+03
.5010E+06	.1133E+03



Specifications:

- Initial frequency = 0.0 (WSTART = 0.0)
- Final frequency = 2.0 (WEND = 2.0)
- Frequency additive step = 0.1 (FREQADD = 0.1)
- Desire error to fall below 10^{-5} or a maximum of 39 iterative minimization steps at each frequency point. (EPS = 10^{-5} , NITT = 39)
- Do not desire tree and \hat{F} to be outputted (NCONT = 0)
- Desire internal automatic current scaling (NSCALE \neq 1, SCALE is arbitrary)
- Do not desire an optimal tree to be selected at each frequency (NITREE \neq 1)
- Do not desire testing for ideal resonance (IDRES = 1)
- Desire the following graphical outputs

Branch 3 current magnitude	}	(NGRAPH = 5)
Branch 5 voltage phase		
Branch 8 voltage magnitude		
Branch 11 current magnitude		
Branch 13 current phase		

The required data cards, for this problem, may be seen on a subsequent page. The central processor time for this problem was 10.37 seconds.

Comments on sample problem #2:

Note that when $\omega = 0$, only the outer resistor loop can carry a current (since all the C's are open circuits). Thus, the current in branch 11 is 1/3 amp, and the voltage across branch 8 is zero. The results of CANDOFD agree with the above. As $\omega \rightarrow \infty$, the current in branch 11 will again tend toward 1/3, and the voltage across branch 8 will again tend toward zero (since the L's now become open circuits).

Comparison of different solution schemes for sample problem #2:

* implies that the indicated scheme has been implemented

CP indicates the central processor time, in seconds

AUTO indicates automatic, internal current scaling

TREE indicates that an optimal tree is to be selected at each frequency point

OUT indicates that tree and \underline{F} outputs are desired

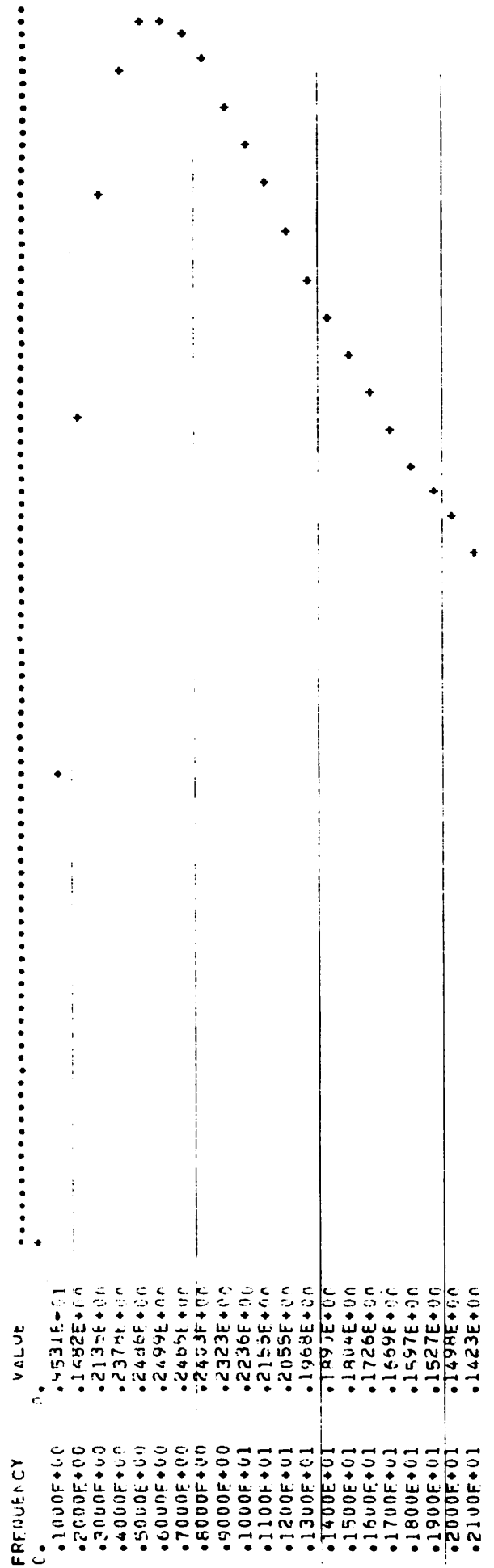
RES indicates that a test for resonances is desired

Scheme	AUTO	TREE	RES	OUT	CP
1	*	*	*	*	13.54
2	*	*	*		10.98
3	*	*			10.91
4	*				10.37
5	SCALE = 1.0	*			10.64
6	SCALE = 1.0				10.30

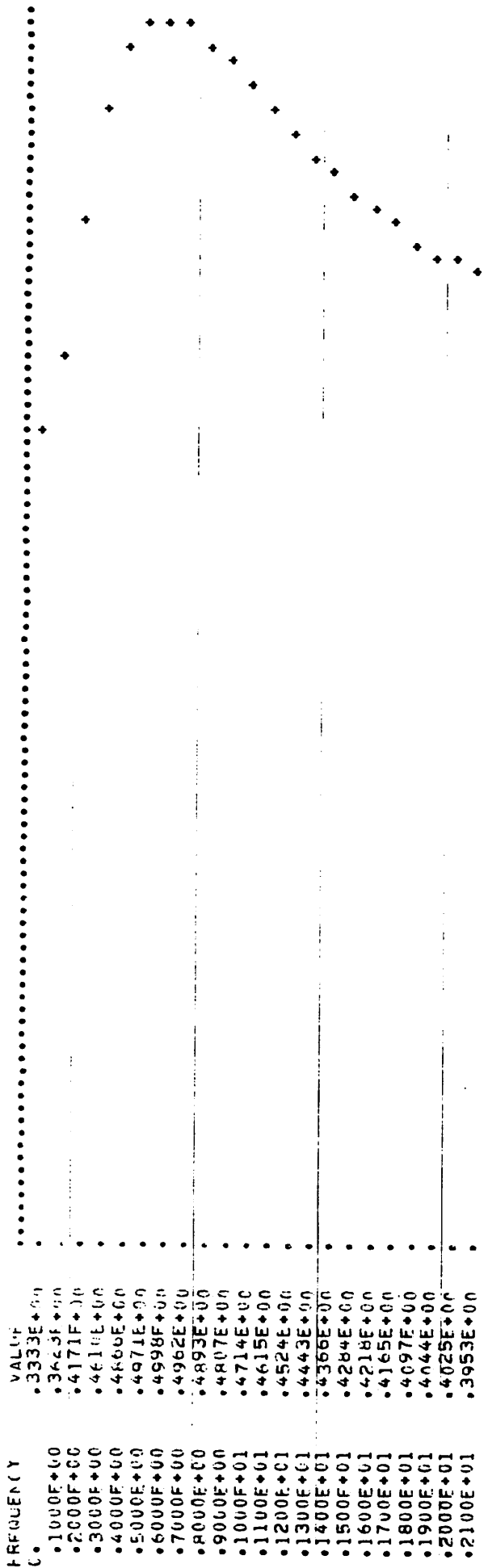
It should be noted from a comparison of the 1st and 2nd schemes, that the CP time for outputting the tree and \underline{F} information is substantial, and hence should not be used unless it is specifically desired. Even though schemes 4 has a CP time slightly below that of scheme 3, its results are significantly less accurate.

In general, scheme 3 is the recommended procedure when one does not have a clear idea of what the network signals will be.

BRANCH H VOLTAGE MAGNITUDE PLOT

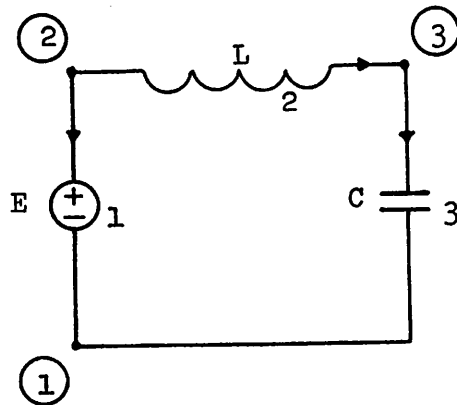


BRANCH 11 CURRENT MAGNITUDE PLOT



Sample Problem #3:

Ideal L-C filter section:



$$L = 1.0$$

$$C = 1.0$$

$$NN = 3$$

$$NB = 3$$

$$|E| = 1.0, \angle E = 0.0$$

Specifications:

- Initial frequency = 0.0 (WSTART = 0.0)
- Final frequency = 2.5 (WEND = 2.5)
- Frequency additive factor = 0.1 (FREQADD = 0.1)
- Desire error to fall below 10^{-5} (EPS = 10^{-5})
- Desire tree and \hat{F} to be outputted (NCONT = 1)
- Desire internal automatic current scaling (SCALE is arbitrary, NSCALE \neq 1)
- Desire an optimal tree to be selected at each frequency point (NTREE = 1)
- Desire the following graphical outputs
 - Branch 1 voltage magnitude
 - Branch 2 current magnitude
 - Branch 2 current phase
 - Branch 3 voltage magnitude
 - Branch 3 voltage phase

$$(NGRAPH = 5)$$

The required data cards for this problem may be seen on a subsequent page. The central processor time for this problem was 2.61 seconds.

Comments on Sample Problem #3:

Note the XXX's on the graphical outputs indicating an ideal resonance at $\omega = 1.0$.

BRANCH 3 VOLTAGE MAGNITUDE PLOT

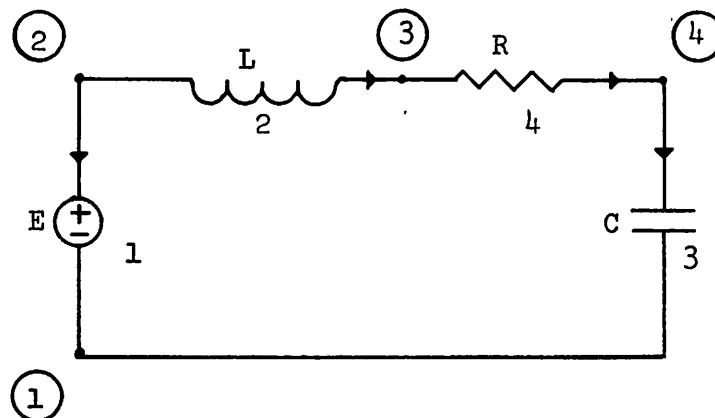
FREQUENCY	VALUE	
0.	.1000E+01
.1000E+00	.1019E+01	.
.2000E+00	.1042E+01	.
.3000E+00	.1099E+01	.
.4000E+00	.1190E+01	.
.5000E+00	.1333E+01	.
.6000E+00	.1562E+01	.
.7000E+00	.1961E+01	.
.8000E+00	.2778E+01	.
.9000E+00	.5263E+01	.
.1050E+01	.9750E+01	XXXXXXXXXXXX
.1100E+01	.4762E+01	.
.1200E+01	.2273E+01	.
.1300E+01	.1449E+01	.
.1400E+01	.1042E+01	.
.1500E+01	.8000E+00	.
.1600E+01	.6410E+00	.
.1700E+01	.5291E+00	.
.1800E+01	.4464E+00	.
.1900E+01	.3831E+00	.
.2000E+01	.3333E+00	.
.2100E+01	.2933E+00	.
.2200E+01	.2604E+00	.
.2300E+01	.2331E+00	.
.2400E+01	.2101E+00	.
.2500E+01	.1905E+00	.
.2600E+01	.1736E+00	.

BRANCH 3 VOLTAGE PHASE PLOT

FREQUENCY	VALUE	PHASE PLOT
0.	0.
.1000E+00	0.	+
.2000E+00	0.	+
.3000E+00	0.	+
.4000E+00	0.	+
.5000E+00	0.	+
.6000E+00	0.	+
.7000E+00	0.	+
.8000E+00	0.	+
.9000E+00	0.	+
.1050E+01	-.1800E+03	+ XXXXXXXXXXXX
.1100E+01	-.1800E+03	+
.1200E+01	-.1800E+03	+
.1300E+01	-.1800E+03	+
.1400E+01	-.1800E+03	+
.1500E+01	-.1800E+03	+
.1600E+01	-.1800E+03	+
.1700E+01	-.1800E+03	+
.1800E+01	-.1800E+03	+
.1900E+01	-.1800E+03	+
.2000E+01	-.1800E+03	+
.2100E+01	-.1800E+03	+
.2200E+01	-.1800E+03	+
.2300E+01	-.1800E+03	+
.2400E+01	-.1800E+03	+
.2500E+01	-.1800E+03	+
.2600E+01	-.1800E+03	+

Sample Problem #4:

Lossy L-C filter section:



$$R = 0.1$$

$$L = 1.0$$

$$C = 1.0$$

$$NN = 4$$

$$NB = 4$$

$$|E| = 1.0, \quad \angle E = 0.0$$

Specifications

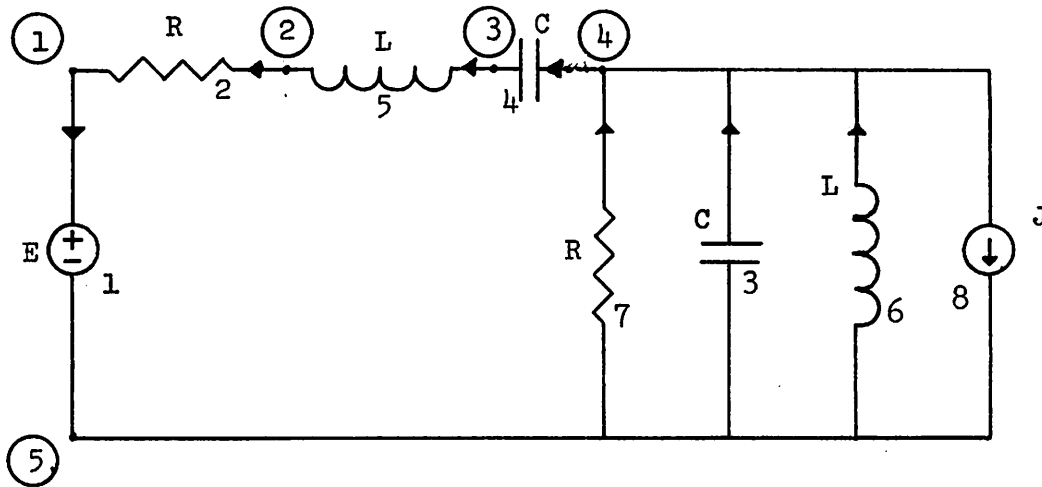
- Initial frequency = 0.0 (WSTART = 0.0)
- Final frequency = 2.5 (WEND = 2.5)
- Frequency additive factor = 0.1 (FREQADD = 0.1)
- Desire error to fall below 10^{-5} (EPS = 10^{-5})
- Do not desire tree and \underline{F} to be outputted (NCONT = 0)
- Desire internal automatic current scaling (SCALE is arbitrary,
NSCALE \neq 1)
- Desire an optimal tree to be selected at each frequency point
(NITREE = 1)
- Desire the following graphical outputs.

BRANCH 3 VOLTAGE MAGNITUDE PLOT

FREQUENCY	VALUE
.100E+00	.100E+01
.101E+01	.101E+01
.200E+00	.104E+01
.300E+00	.109E+01
.400E+00	.118E+01
.500E+00	.133E+01
.600E+00	.155E+01
.700E+00	.194E+01
.800E+00	.271E+01
.900E+00	.475E+01
.100E+01	.100E+02
.110E+01	.421E+01
.120E+01	.219E+01
.130E+01	.142E+01
.140E+01	.103E+01
.150E+01	.794E+00
.160E+01	.637E+00
.170E+01	.527E+00
.180E+01	.445E+00
.190E+01	.382E+00
.200E+01	.332E+00
.210E+01	.292E+00
.220E+01	.260E+00
.230E+01	.232E+00
.240E+01	.209E+00
.250E+01	.190E+00
.260E+01	.173E+00

BRANCH 3 VOLTAGE PHASE PLOT

FREQUENCY	VALUE
0.	0.
.1000E+00	-.5788E+00
.2000E+00	-.1193E+01
.3000E+00	-.1888E+01
.4000E+00	-.2726E+01
.5000E+00	-.3814E+01
.6000E+00	-.5356E+01
.7000E+00	-.7815E+01
.8000E+00	-.1253E+02
.9000E+00	-.2535E+02
.1000E+01	-.9000E+02
.1100E+01	-.1524E+03
.1200E+01	-.1647E+03
.1300E+01	-.1693E+03
.1400E+01	-.1717E+03
.1500E+01	-.1732E+03
.1600E+01	-.1741E+03
.1700E+01	-.1749E+03
.1800E+01	-.1754E+03
.1900E+01	-.1758E+03
.2000E+01	-.1762E+03
.2100E+01	-.1765E+03
.2200E+01	-.1767E+03
.2300E+01	-.1769E+03
.2400E+01	-.1771E+03
.2500E+01	-.1773E+03
.2600E+01	-.1774E+03

Sample Problem #5:

$$|E| = |J| = 10.0$$

$$R = 1.0$$

$$\angle E = \angle J = 0^\circ$$

$$L = 0.05$$

$$NN = 5, NB = 8$$

$$C = 0.05$$

Specifications:

- Initial frequency = 0.0 (WSTART = 0.0)
- Final frequency = 50.0 (WEND = 50.0)
- Frequency additive factor = 1.0 (FREQADD = 1.0)
- Desire error to fall below 10^{-4} or a maximum of 24 iterative minimization steps at each frequency point. (EPS = 10^{-4} , NITR = 24)
- Do not desire tree and \underline{F} to be outputted (NCONT = 0)
- Desire internal automatic current scaling. (SCALE is arbitrary, NSCALE \neq 1)
- Desire an optimal tree to be selected at each frequency point (NIFREE = 1)

- Desire the following graphical outputs

Branch 1 current magnitude

Branch 2 voltage magnitude

Branch 5 current magnitude

Branch 5 current phase

Branch 8 current magnitude

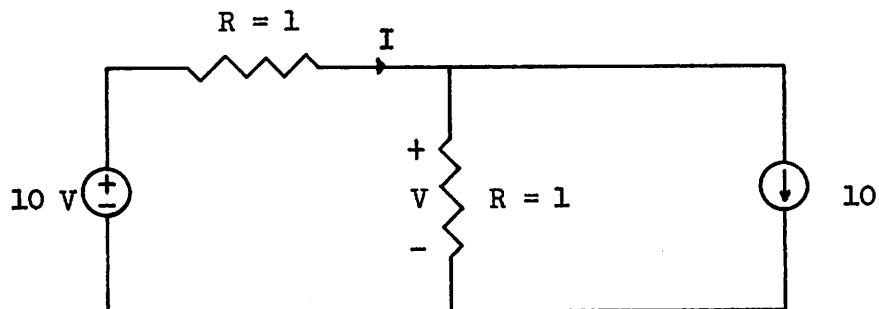
(NGRAPH = 5)

- Desire internal testing for resonances (IDRES \neq 1)

The required data cards for this problem may be seen on a subsequent page. The central processor time for this problem was 10.56 seconds.

Comments on sample problem #5:

At $\omega = 20.0$, the L-C series and parallel combinations act as short and open circuits respectfully. Thus, our network reduces to



The solution to this network is

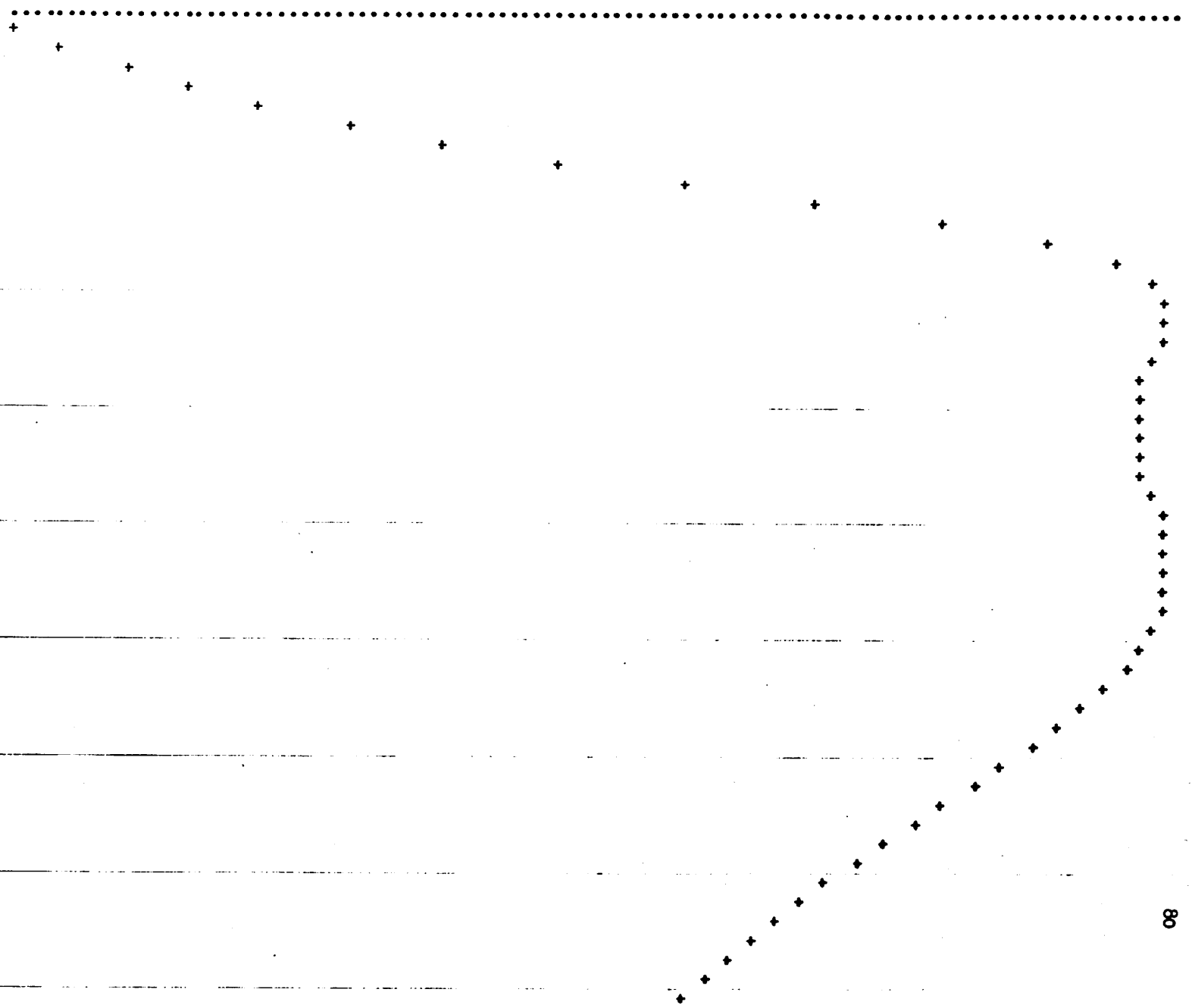
$$V = 0 \text{ volts}$$

$$I = 10 \text{ amps}$$

This is borne out by CANDOFD.

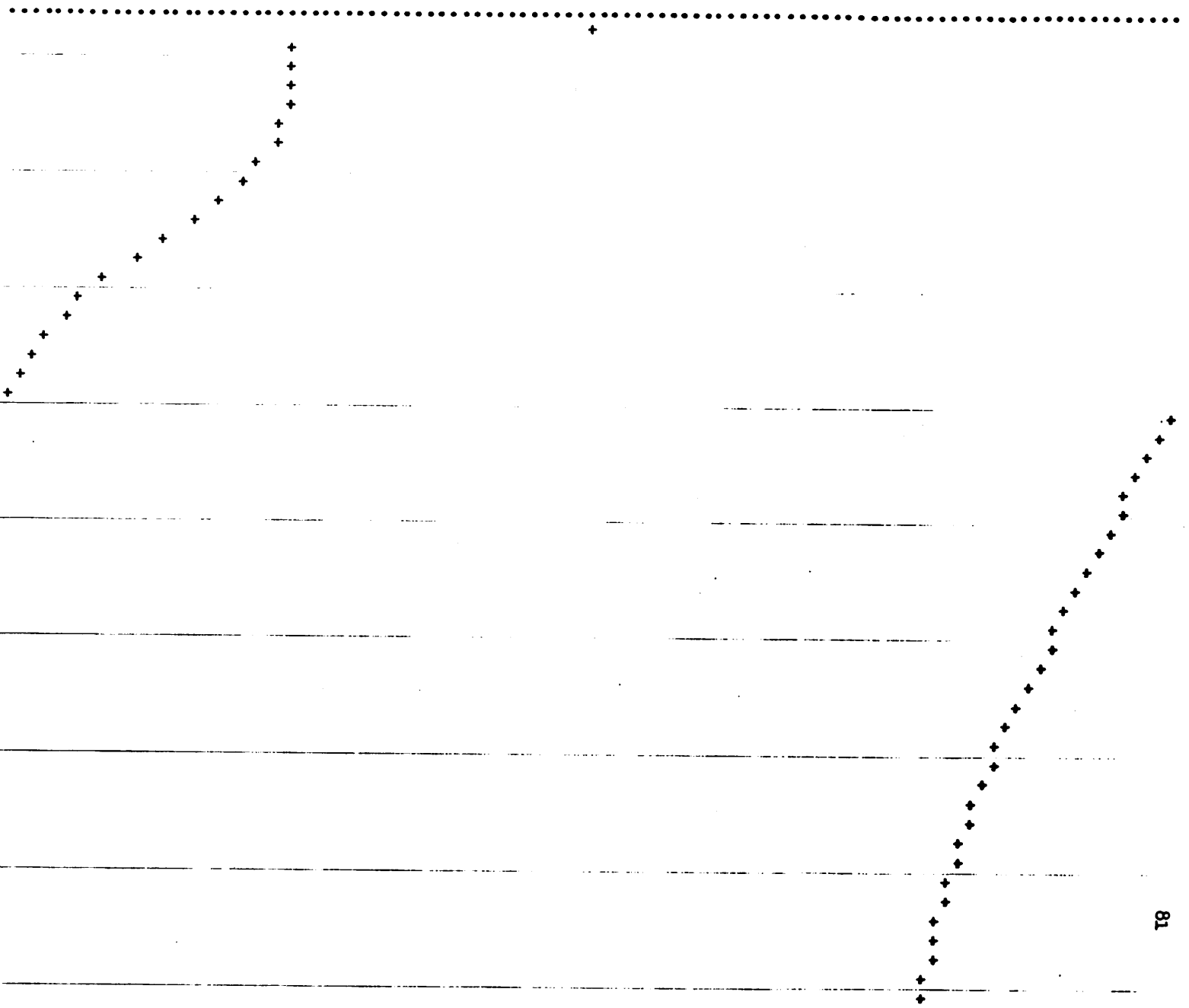
BRANCH 5 CURRENT MAGNITUDE PLOT

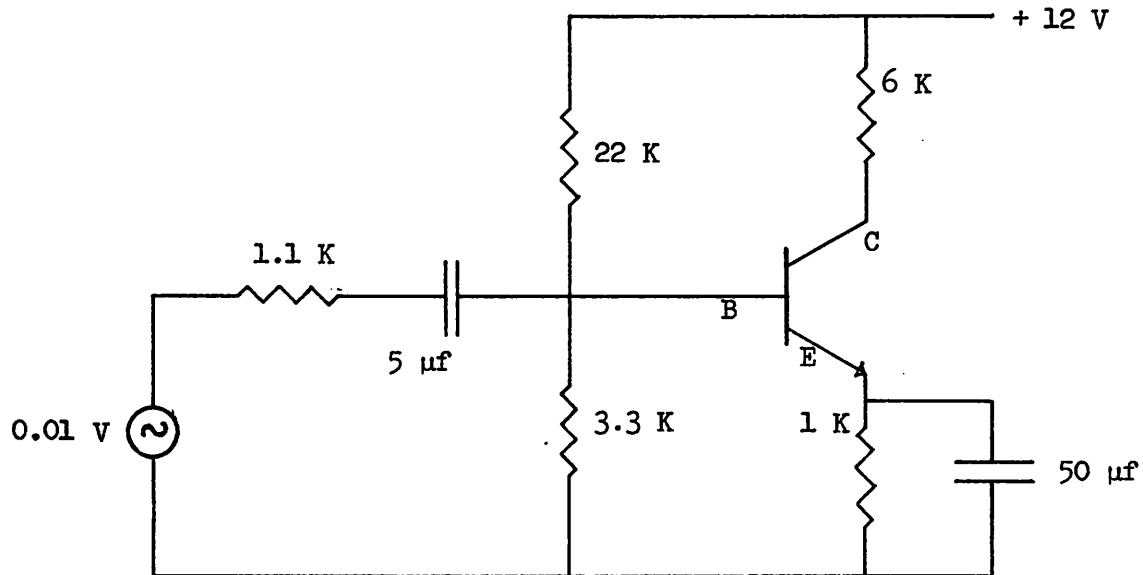
FREQUENCY	VALUE
.1000E+1	.5038E+00
.2000E+1	.1030E+01
.3000E+1	.1603E+01
.4000E+1	.2248E+01
.5000E+1	.2992E+01
.6000E+1	.3859E+01
.7000E+1	.4862E+01
.8000E+1	.5989E+01
.9000E+1	.7177E+01
.1000E+02	.8305E+01
.1100E+02	.9228E+01
.1200E+02	.9851E+01
.1300E+02	.1018E+02
.1400E+02	.1029E+02
.1500E+02	.1027E+02
.1600E+02	.1020E+02
.1700E+02	.1012E+02
.1800E+02	.1005E+02
.1900E+02	.1001E+02
.2000E+02	.1000E+02
.2100E+02	.1001E+02
.2200E+02	.1004E+02
.2300E+02	.1008E+02
.2400E+02	.1014E+02
.2500E+02	.1020E+02
.2600E+02	.1025E+02
.2700E+02	.1028E+02
.2800E+02	.1029E+02
.2900E+02	.1028E+02
.3000E+02	.1023E+02
.3100E+02	.1016E+02
.3200E+02	.1005E+02
.3300E+02	.9905E+01
.3400E+02	.9733E+01
.3500E+02	.9533E+01
.3600E+02	.9312E+01
.3700E+02	.9074E+01
.3800E+02	.8824E+01
.3900E+02	.8567E+01
.4000E+02	.8305E+01
.4100E+02	.8044E+01
.4200E+02	.7785E+01
.4300E+02	.7531E+01
.4400E+02	.7284E+01
.4500E+02	.7045E+01
.4600E+02	.6813E+01
.4700E+02	.6594E+01
.4800E+02	.6383E+01
.4900E+02	.6181E+01
.5000E+02	.5989E+01



BRANCH 5 CURRENT PHASE PLOT

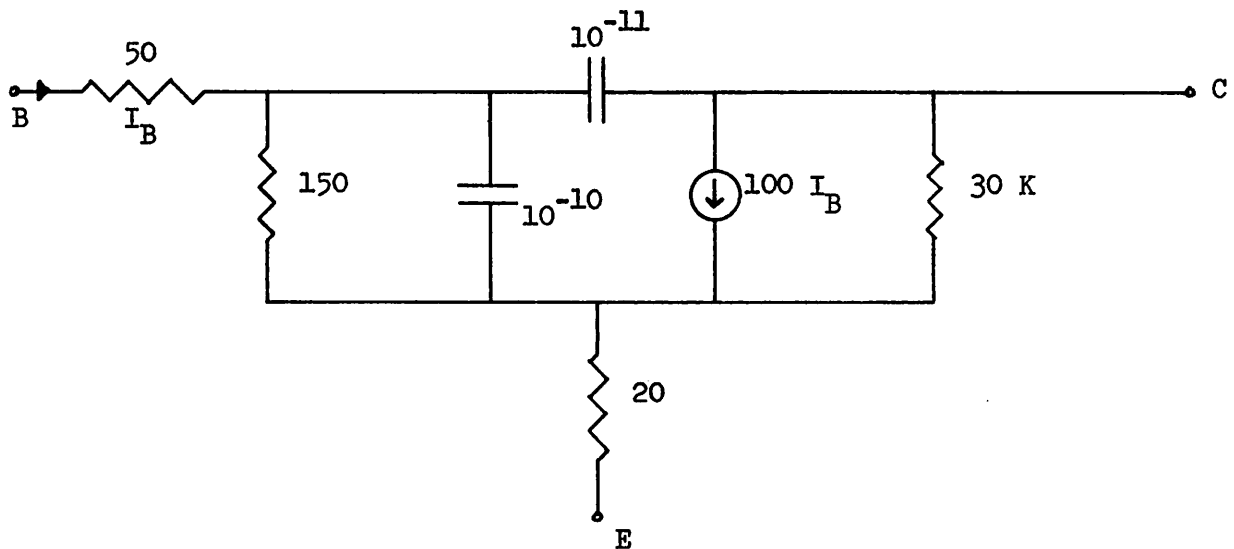
FREQUENCY C.	VALUE D.
.1000E+01	-.9003E+02
.2000E+01	-.9023E+02
.3000E+01	-.9079E+02
.4000E+01	-.9191E+02
.5000E+01	-.9380E+02
.6000E+01	-.9672E+02
.7000E+01	-.1009E+03
.8000E+01	-.1066E+03
.9000E+01	-.1137E+03
.1000E+02	-.1221E+03
.1100E+02	-.1312E+03
.1200E+02	-.1401E+03
.1300E+02	-.1482E+03
.1400E+02	-.1553E+03
.1500E+02	-.1612E+03
.1600E+02	-.1661E+03
.1700E+02	-.1702E+03
.1800E+02	-.1738E+03
.1900E+02	-.1770E+03
.2000E+02	.1800E+03
.2100E+02	.1772E+03
.2200E+02	.1744E+03
.2300E+02	.1717E+03
.2400E+02	.1689E+03
.2500E+02	.1661E+03
.2600E+02	.1632E+03
.2700E+02	.1601E+03
.2800E+02	.1570E+03
.2900E+02	.1539E+03
.3000E+02	.1507E+03
.3100E+02	.1475E+03
.3200E+02	.1443E+03
.3300E+02	.1411E+03
.3400E+02	.1380E+03
.3500E+02	.1350E+03
.3600E+02	.1322E+03
.3700E+02	.1294E+03
.3800E+02	.1268E+03
.3900E+02	.1244E+03
.4000E+02	.1221E+03
.4100E+02	.1200E+03
.4200E+02	.1180E+03
.4300E+02	.1161E+03
.4400E+02	.1144E+03
.4500E+02	.1128E+03
.4600E+02	.1114E+03
.4700E+02	.1100E+03
.4800E+02	.1088E+03
.4900E+02	.1076E+03
.5000E+02	.1065E+03



Sample Problem #6:One stage amplifier:

where μf indicates microfarads and K indicates ohms $\times 10^3$.

The transistor is represented by the following small signal model:



With the above transistor model and appropriate controlled source modeling, the small signal equivalent of our network is;

- Desire the following graphical outputs

Branch 1 current magnitude	}	(NGRAPH = 4)
Branch 7 current magnitude		
Branch 14 current magnitude		
Branch 16 current magnitude		

The required data cards for this problem may be seen on a subsequent page. The central processor time, for this problem, was 69 seconds, i.e., approximately 2 seconds per frequency point. (Note that this is, essentially, an exact analysis.)

Comments on sample problem #6:

The mid-band voltage gain, given by CANDOFD, is

$$G_V = \frac{R_{16} \times I_{16}}{V_1} = \frac{6 \times 10^3 \times 0.22 \times 10^{-3}}{0.01} \cong 132$$

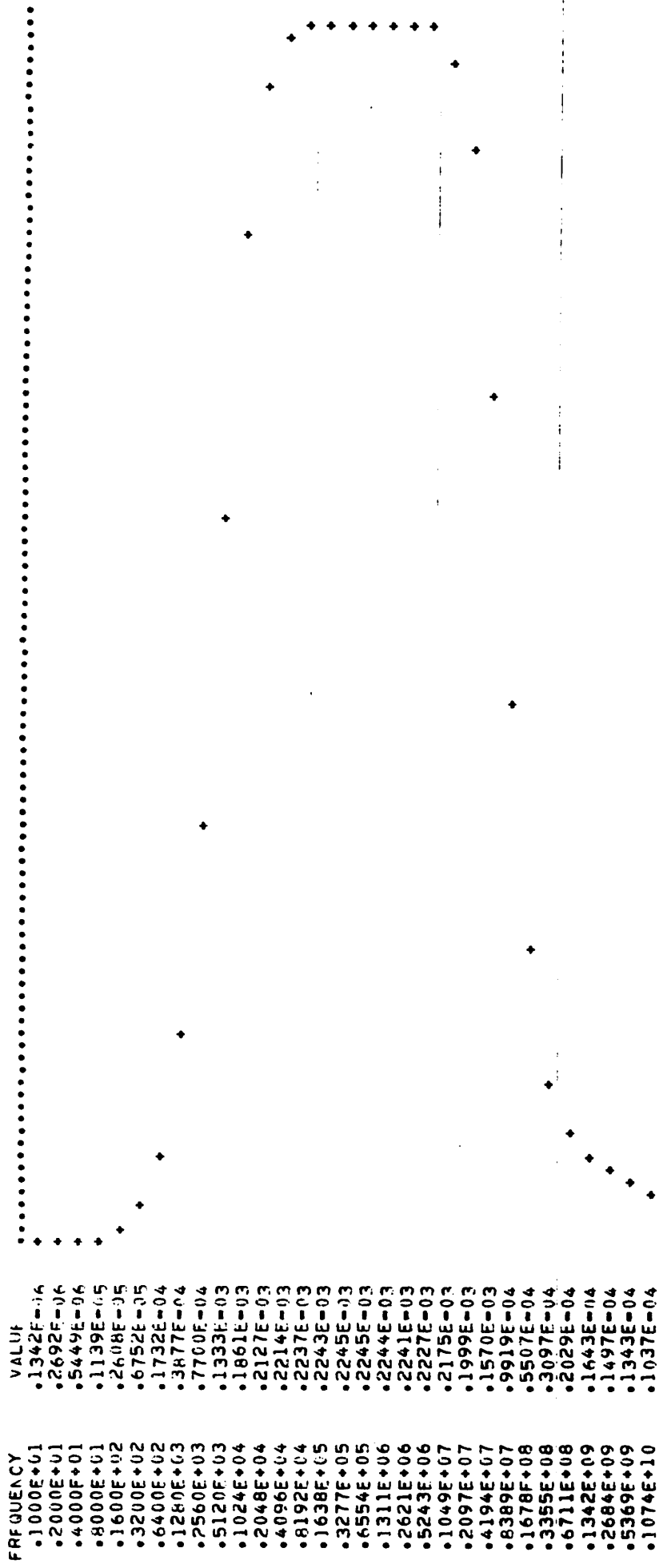
where $R_{16} = 6 \text{ K}$.

The mid-band power gain, given by CANDOFD, is

$$\begin{aligned} G_P &= \frac{R_{16} \times I_{16}^2}{I_1^2} = G_V \times \frac{I_{16}}{I_1} \\ &= 132 \times \frac{0.22 \times 10^{-3}}{0.44 \times 10^{-5}} \\ &\cong 6.6 \times 10^3 \end{aligned}$$

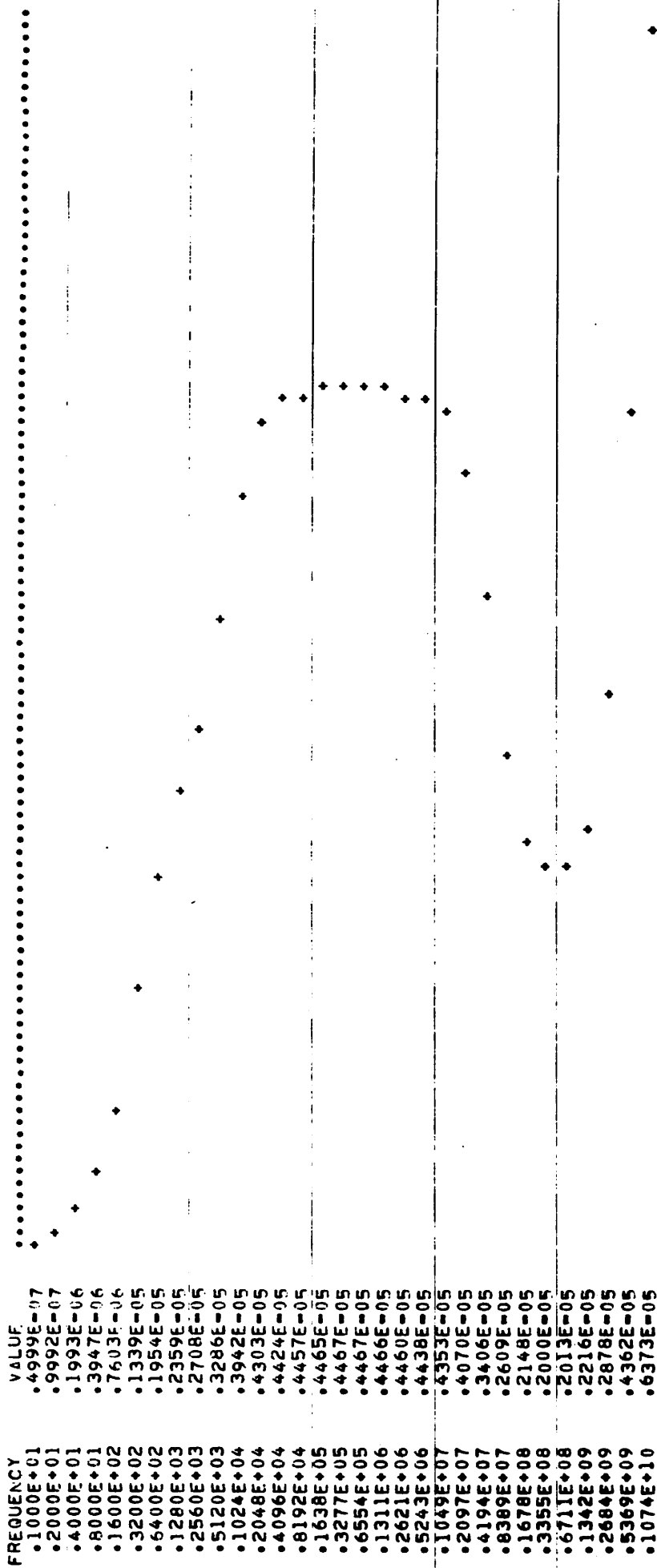
These results agree well with theoretical results for this amplifier.

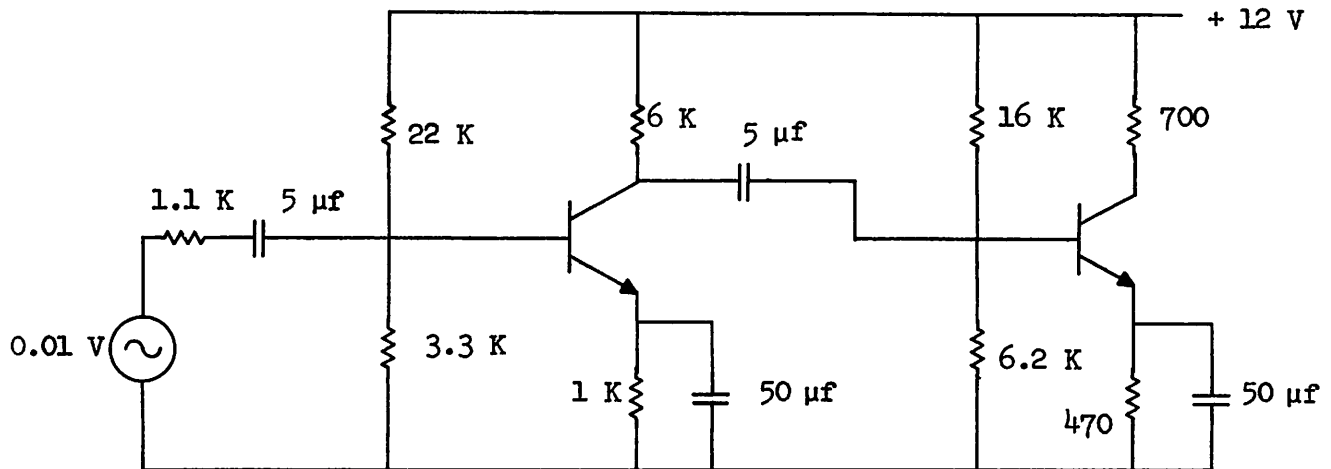
BRANCH 16 CURRENT MAGNITUDE PLOT



4 5

BRANCH 1 CURRENT MAGNITUDE PLOT



Sample Problem #7:Two Stage Audio Amplifier:

We will use the small signal transistor model described in sample problem #6.

Specifications:

- Initial frequency = 1.0 (WSTART = 1.0)
- Final frequency = 10^9 (WEND = 10^9)
- Frequency multiplicative factor = 10^3 (FACTOR = 10^3 , FREQADD = 0.0)
- Desire error to fall below 10^{-12} or a maximum of 60 iterative minimization steps at each frequency point. (EPS = 10^{-12} , NITT = 60)
- Do not desire tree and \hat{F} outputted (NCONT = 0)
- Desire an optimal tree to be selected at each frequency point
(NTREE = 1)
- Do not desire testing for ideal resonances (IDRES = 1)
- Desire internal, automatic current scaling (SCALE is arbitrary,
NSCALE \neq 1)

The central processor time for this problem was 56 seconds.

BRANCH		ABS VALUE =		PHASE =	
BRANCH	VOLTAGE	1	1000E-01	PHASE =	0.
BRANCH	CURRENT	1	49889E-07	PHASE =	-0.91107E+02
BRANCH	CURRENT	7	16302E-08	PHASE =	0.91317E+02
BRANCH	CURRENT	14	16302E-06	PHASE =	0.91317E+02
BRANCH	CURRENT	17	31241E-08	PHASE =	0.19030E+00
BRANCH	CURRENT	20	10072E-09	PHASE =	-0.82742E+02
BRANCH	CURRENT	22	42346E-09	PHASE =	-0.87670E+02
BRANCH	CURRENT	25	10072E-07	PHASE =	-0.82794E+02
BRANCH	VOLTAGE	30	51371E-05	PHASE =	0.10099E+03
BRANCH	CURRENT	30	10442E-07	PHASE =	0.96154E+02

FREQUENCY = 1000E+01

BRANCH		ABS VALUE =		PHASE =	
BRANCH	VOLTAGE	1	1000E-01	PHASE =	0.
BRANCH	CURRENT	1	37743E-05	PHASE =	-0.16399E+03
BRANCH	CURRENT	7	20042E-05	PHASE =	0.36456E+02
BRANCH	CURRENT	14	20042E-03	PHASE =	0.36456E+02
BRANCH	CURRENT	17	14709E-03	PHASE =	-0.13506E+03
BRANCH	CURRENT	20	93781E-04	PHASE =	-0.11875E+03
BRANCH	CURRENT	22	93780E-04	PHASE =	-0.11879E+03
BRANCH	CURRENT	25	93781E-02	PHASE =	-0.11875E+03
BRANCH	VOLTAGE	30	64106E+01	PHASE =	0.61285E+02
BRANCH	CURRENT	30	91580E-02	PHASE =	0.61285E+02

FREQUENCY = 1000E+04

FREQUENCY = .10000E+07

BRANCH	1	VOLTAGE	ABS VALUE =	.10000E-01	PHASE =	0.
BRANCH	1	CURRENT	ABS VALUE =	.42974E-05	PHASE =	-.17841E+03
BRANCH	7	CURRENT	ABS VALUE =	.24649E-05	PHASE =	.38127E+01
BRANCH	14	CURRENT	ABS VALUE =	.24649E-03	PHASE =	.38126E+01
BRANCH	17	CURRENT	ABS VALUE =	.19051E-03	PHASE =	-.17635E+03
BRANCH	20	CURRENT	ABS VALUE =	.12821E-03	PHASE =	-.17528E+03
BRANCH	22	CURRENT	ABS VALUE =	.15619E-03	PHASE =	.14852E+03
BRANCH	25	CURRENT	ABS VALUE =	.12821E-01	PHASE =	-.17528E+03
BRANCH	30	VOLTAGE	ABS VALUE =	.87637E+01	PHASE =	.43148E+01
BRANCH	30	CURRENT	ABS VALUE =	.12520E-01	PHASE =	.43148E+01

FREQUENCY = .10000E+10

BRANCH	1	VOLTAGE	ABS VALUE =	.10000E-01	PHASE =	0.
BRANCH	1	CURRENT	ABS VALUE =	.57985E-05	PHASE =	.13248E+03
BRANCH	7	CURRENT	ABS VALUE =	.62134E-05	PHASE =	-.71947E+02
BRANCH	14	CURRENT	ABS VALUE =	.62134E-03	PHASE =	-.71947E+02
BRANCH	17	CURRENT	ABS VALUE =	.12148E-03	PHASE =	-.68116E+02
BRANCH	20	CURRENT	ABS VALUE =	.12285E-03	PHASE =	-.76623E+02
BRANCH	22	CURRENT	ABS VALUE =	.79876E-03	PHASE =	.26153E+02
BRANCH	25	CURRENT	ABS VALUE =	.12285E-01	PHASE =	-.76623E+02
BRANCH	30	VOLTAGE	ABS VALUE =	.12951E+01	PHASE =	.22496E+02
BRANCH	30	CURRENT	ABS VALUE =	.18502E-02	PHASE =	.22495E+02

APPENDIX D:

CANDOFD FORTRAN IV LISTING

(CDC 6400)

	PROGRAM CANDOFD(INPUT,OUTPUT)	A	1
	COMPLEX E,CONST,V,C,GRAD	A	2
	COMPLEX CONDTEM	A	3
	COMPLEX GRAF	A	4
	COMPLEX TEMP,PAR,DUMMY	A	5
	COMPLEX X,Y	A	6
	COMPLEX CONT,S	A	7
	INTEGER CNTSOR	A	8
	INTEGER CONTYPE	A	9
	INTEGER CONTTEM	A	10
	INTEGER TEMP	A	11
	INTEGER TYPE	A	12
	COMMON /BLOCK1/ NNP(200),NP(200),IDRES	A	13
	COMMON /BLOCK2/ IBRAN(200),LEAV(200),LENT(200)	A	14
	COMMON /BLOCK3/ TYPE(200),VALUE(200),F(49,151),NOEL(7)	A	15
	COMMON /BLOCK4/ ITBRAN(200),LEAVT(200),LENTT(200)	A	16
	COMMON /BLOCK5/ IOUT(200),ITEST(200),NOUT(200),ITTEST(200)	A	17
	COMMON /BLOCK6/ TEMP(7),E(200),GRAD(200),CONST(200)	A	18
	COMMON /BLOCK7/ ALPHA,FUNCT,NIC,FACTOR,EPS,NN,NB	A	19
	COMMON /BLOCK8/ V(200),C(200)	A	20
	COMMON /BLOCK9/ CONTYPE(200),KONBRAN(200)	A	21
	COMMON /BLOCK10/ AMAG(200),APHASE(200),WSTART,WEND,NRES	A	22
	COMMON /BLOCK11/ CONT(200),S(200)	A	23
	COMMON /BLOCK12/ CONTTEM(200),CONDTEM(200),KONTEM(200)	A	24
	COMMON /BLOCK13/ X(200),Y(200),ITN	A	25
	COMMON /BLOCK14/ KP(5),W,FREQADD	A	26
	COMMON /BLOCK15/ NIT,JOUT,NGRAPH,NALLOUT	A	27
	COMMON /BLOCK16/ GRAF(200,5),JGRAPH(5),IGRAPH(5),NPRINT,SCALE,NITT	A	28
	COMMON /BLOCK19/ CNTSOR(200)	A	29
	COMMON /BLOCK20/ NCONT,KLOOP,KPP	A	30
	COMMON /BLOCK21/ NST(5)	A	31
	COMMON /BLOCK23/ TEMP,PAR,DUMMY(200)	A	32
	COMMON /BLOCK24/ SCAL,NSCALE,NTREE	A	33
	DIMENSION HH(54,54)	A	34
	NDATA=0	A	35
1	NIC=0	A	36
	NRES=0	A	37
	IF (NDATA.EQ.0) GO TO 2	A	38
	CALL SECOND (T)	A	39
	TO=T-TU	A	40
	PRINT 21, TO	A	41
2	CALL SECOND (TU)	A	42
	CALL READIN	A	43
	NDATA=NDATA+1	A	44
	KLOOP=0	A	45
	CALL PT	A	46
	IF (KLOOP.EQ.1) GO TO 1	A	47
	NB2=(NB-NOEL(1)-NOEL(7))*2	A	48
	CALL FCSM	A	49
	NLS=NN-1	A	50
	CALL INITIAL	A	51
	NPRINT=0	A	52
	IF (FREQADD) 3,4,3	A	53
3	W=WSTART-FREQADD	A	54
	GO TO 6	A	55

4	W=WSTART/FACTOR	A	56
5	W=W*FACTOR	A	57
	NTRES=0	A	58
	GO TO 7	A	59
6	W=W+FREQADD	A	60
	NTRES=0	A	61
7	NIC=NIC+1	A	62
	IF (NIC.EQ.1) GO TO 8	A	63
	IF (NTREE.NE.1) GO TO 8	A	64
	CALL PT	A	65
	IF (KPP.EQ.0) GO TO 8	A	66
	CALL FCSM	A	67
8	CALL KONST	A	68
	NIT=0	A	69
	ITN=0	A	70
	KRES=0	A	71
	IF (NIC.EQ.1) SCAL=1.0	A	72
	DO 9 I=NN,NB	A	73
	C(I)=C(I)*SCALE/SCAL	A	74
9	CONTINUE	A	75
10	CALL CALCAL	A	76
	IF (NITT.EQ.0) GO TO 11	A	77
	IF (NIT-NITT) 11,17,17	A	78
11	CALL ERROR	A	79
	IF (IDRES.EQ.1) GO TO 13	A	80
	IF (KRES.EQ.0) FRES=FUNCT	A	81
	IF (KRES.EQ.0) GO TO 12	A	82
	A=ABS(FRES-FUNCT)	A	83
	B=1.0E-09*FUNCT	A	84
	IF (A.LT.B) NRES=1	A	85
	FRES=FUNCT	A	86
	IF (NTRES.EQ.1) GO TO 13	A	87
	IF (NRES.EQ.1) GO TO 19	A	88
	GO TO 13	A	89
12	KRES=1	A	90
13	IF (FUNCT-EPS) 17,17,14	A	91
14	CALL FNDGRAD	A	92
	ITN=ITN+1	A	93
	CALL FLPOW (HH,NB,NB2)	A	94
	NIT=NIT+1	A	95
	CALL FNDALPH	A	96
	CALL CNGVARS	A	97
	IF (IDRES.EQ.1) GO TO 10	A	98
	DO 15 I=1,NLS	A	99
	A=REAL(V(I))	A	100
	B=AIMAG(V(I))	A	101
	IF (ABS(A).GT.1.0E+30) GO TO 17	A	102
	IF (ABS(B).GT.1.0E+30) GO TO 17	A	103
15	CONTINUE	A	104
	DO 16 I=NN,NB	A	105
	B=AIMAG(C(I))	A	106
	A=REAL(C(I))	A	107
	IF (ABS(B).GT.1.0E+30) GO TO 17	A	108
	IF (ABS(A).GT.1.0E+30) GO TO 17	A	109
16	CONTINUE	A	110

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GO TO 10 A 111
17 NPRINT=NPRINT+1 A 112
  IF (NGRAPH.GE.1) CALL GRAPH A 113
  IF (NALLOUT.GE.1) CALL ALLOUT A 114
  IF (JOUT.GE.1) CALL READOUT A 115
  IF (NGRAPH.EQ.201) GO TO 1 A 116
  IF (NRES.EQ.1) GO TO 20 A 117
  IF (WEND-W) 1,1,18 A 118
18 IF (NIC.EQ.1) CALL INITIAL A 119
  IF (FREQADD) 6,5,6 A 120
19 DW=0.05*W A 121
  W=W+DW A 122
  NIC=NIC-1 A 123
  NTRES=1 A 124
  GO TO 7 A 125
20 W=W-DW A 126
  NRES=0 A 127
  GO TO 18 A 128
C A 129
C A 130
21 FORMAT (1H1,20X,*THE CENTRAL PROCESSOR TIME FOR THIS PROBLEM = * A 131
1,E15.5,* SECONDS*) A 132
END A 133-
```

	SUBROUTINE READIN	B	1
C	NB IS THE NUMBER OF BRANCHES	B	2
C	NN IS THE NUMBER OF NODES	B	3
C		B	4
	COMPLEX GRAF	B	5
	INTEGER CONTYPE	B	6
	INTEGER CONTTEM	B	7
	INTEGER TYPE	B	8
	COMMON /BLOCK1/ NNP(200),NP(200),IDRES	B	9
	COMMON /BLOCK2/ IBRAN(200),LEAV(200),LENT(200)	B	10
	COMMON /BLOCK3/ TYPE(200),VALUE(200),F(49,151),NOEL(7)	B	11
	COMMON /BLOCK4/ ITBRAN(200),LEAVT(200),LENTT(200)	B	12
	COMMON /BLOCK5/ IOUT(200),ITEST(200),NOUT(200),ITTEST(200)	B	13
	COMMON /BLOCK7/ ALPHA,FUNCT,NIC,FACTOR,EPS,NN,NB	B	14
	COMMON /BLOCK9/ CONTYPE(200),KONBRAN(200)	B	15
	COMMON /BLOCK10/ AMAG(200),APHASE(200),WSTART,WEND	B	16
	COMMON /BLOCK14/ KP(5),W,FREQADD	B	17
	COMMON /BLOCK15/ NIT,JOUT,NGRAPH,NALLOUT	B	18
	COMMON /BLOCK16/ GRAF(200,5),JGRAPH(5),IGRAPH(5),NPRINT,SCALE,NITT	B	19
	COMMON /BLOCK20/ NCONT,KLOOP,KPP	B	20
	COMMON /BLOCK24/ SCAL,NSCALE,NTREE	B	21
	IJP=0	B	22
	READ 12, NN,NB,WSTART,WEND,FACTOR,EPS,NCONT,NTREE	B	23
	IF (NN.EQ.0) STOP	B	24
	READ 13, NGRAPH,NALLOUT,JOUT,SCALE,NITT,FREQADD,NSCALE,IDRES	B	25
	PRINT 14, EPS,WSTART,WEND,FACTOR	B	26
	PRINT 15, NN,NB,SCALE,NITT,FREQADD,NSCALE,IDRES	B	27
	PRINT 16, NTREE	B	28
	DO 1 I=1,NB	B	29
	READ 17, TYPE(I),IBRAN(I),CONTYPE(I),KONBRAN(I),LEAV(I),LENT(I),VA	B	30
	1LUE(I),AMAG(I),APHASE(I)	B	31
1	CONTINUE	B	32
	IF (NGRAPH.EQ.0) GO TO 2	B	33
	READ 18, ((IGRAPH(I),JGRAPH(I),ITTEST(I)),I=1,NGRAPH)	B	34
2	IF (NALLOUT.GE.1.OR.NGRAPH.GE.1) GO TO 3	B	35
	READ 19, ((IOUT(I),ITEST(I)),I=1,JOUT)	B	36
3	PRINT 20	B	37
	PRINT 21	B	38
	DO 4 I=1,NB	B	39
	PRINT 22, IBRAN(I),TYPE(I),VALUE(I),LEAV(I),LENT(I)	B	40
4	CONTINUE	B	41
	PRINT 27	B	42
	DO 8 I=1,NB	B	43
	IF (TYPE(I).EQ.1HE) GO TO 6	B	44
	IF (TYPE(I).EQ.1HJ) GO TO 5	B	45
	GO TO 8	B	46
5	A=7HCURRENT	B	47
	GO TO 7	B	48
6	A=7HVOLTAGE	B	49
7	PRINT 26, IBRAN(I),A,AMAG(I),APHASE(I)	B	50
8	CONTINUE	B	51
	DO 11 I=1,NB	B	52
	IF (CONTYPE(I).NE.1H) GO TO 9	B	53
	GO TO 11	B	54
9	IF (IJP.EQ.0) PRINT 23	B	55

	IJP=1	B	56
	IF (TYPE(I).EQ.1HI) GO TO 10	B	57
	A=7HCURRENT	B	58
	IF (CONTYPE(I).EQ.1HV) A=7HVOLTAGE	B	59
	PRINT 24, A,IBRAN(I),KONBRAN(I)	B	60
	GO TO 11	B	61
10	A=7HCURRENT	B	62
	IF (CONTYPE(I).EQ.1HV) A=7HVOLTAGE	B	63
	PRINT 25, A,IBRAN(I),KONBRAN(I)	B	64
11	CONTINUE	B	65
	W=WSTART	B	66
	RETURN	B	67
C		B	68
C		B	69
12	FORMAT (I5,I5,E15.3,E15.3,2E15.3,2I5)	B	70
13	FORMAT (3I5,E15.5,I5,E15.5,I5,I5)	B	71
14	FORMAT (1H1,////,40X,*ERROR CRITERION = *,E15.5,////,40X,* INITIAL	B	72
	1 FREQUENCY = *,E15.5, * FINAL FREQUENCY = *, E15.5,////,40X,* FACTO	B	73
	2R = *,E15.5)	B	74
15	FORMAT (1H0,////,30X,* NUMBER OF NODES = *,I5,* NUMBER OF BRANC	B	75
	1HES = *,I5,////,30X,* SCALE FACTOR = *,E15.5,////,30X,*DESIRED NUM	B	76
	2BER OF ITERATIONS AT EACH FREQUENCY POINT = *,I5,////,30X,*FREQUENC	B	77
	3Y ADDITIVE STEP = *,E15.5,////,30X,* NSCALE = *,I5,////,30X,* IDRES	B	78
	4 = *,I5)	B	78A
16	FORMAT (1H0,////,35X,*NTREE = *,I5)	B	79
17	FORMAT (A1,I3,1X,A1,I3,1X,I2,1X,I2,5X,3E15.5)	B	80
18	FORMAT (3I5)	B	81
19	FORMAT (2I5)	B	82
20	FORMAT (1H1,55X,*THIS IS THE GIVEN NETWORK*)	B	83
21	FORMAT (1H0,////,55X,*UNITS ARE OHMS, FARADS AND HENRYS *)	B	84
22	FORMAT (1H0,10X,*BRANCH NUMBER *,5X,I3,4X,* IS A *,A1,* OF	B	85
	1 VALUE *,E12.5,* LEAVING NODE *,I3,* AND ENTERING NODE *,I3)	B	86
23	FORMAT (1H0,////,30X,*CONTROLLED SOURCES *,///)	B	87
24	FORMAT (1H0,10X,A7,* CONTROLLED VOLTAGE SOURCE*,I3,* IS CONTROLLED	B	88
	1 BY BRANCH*,I5)	B	89
25	FORMAT (1H0,10X,A7,* CONTROLLED CURRENT SOURCE*,I3,* IS CONTROLLED	B	90
	1 BY BRANCH*,I5)	B	91
26	FORMAT (1H0,10X,*BRANCH*,I4,3X,* IS AN INDEPENDENT *,A7,* SOURCE*,	B	92
	110X,*MAGNITUDE = *,E15.5,10X,*PHASE = *,E15.5)	B	93
27	FORMAT (1H1,30X,*INDEPENDENT SOURCES*,///)	B	94
	END	B	95-

	SUBROUTINE PT	C	1
	COMPLEX V,C	C	2
	COMPLEX TEMP11,TEMP12	C	3
	COMPLEX GRAF	C	4
	INTEGER CNTSOR	C	5
	INTEGER CONTTEM	C	6
	INTEGER TYTEMP	C	7
	INTEGER TEMPO1,TEMPO2,TEMPO3,TEMPO4,TEMPO6,TEMPO9	C	8
	INTEGER TEMP	C	9
	INTEGER TYPE	C	10
	INTEGER CONTYPE	C	11
	COMMON /BLOCK1/ NNP(200),NP(200)	C	12
	COMMON /BLOCK2/ IBRAN(200),LEAV(200),LENT(200)	C	13
	COMMON /BLOCK3/ TYPE(200),VALUE(200),F(49,151),NOEL(7)	C	14
	COMMON /BLOCK4/ ITBRAN(200),LEAVT(200),LENTT(200)	C	15
	COMMON /BLOCK5/ IOUT(200),ITEST(200),NOUT(200),ITTEST(200)	C	16
	COMMON /BLOCK6/ TEMP(7),ITBB(200),TYTEMP(200),VALTEM(200)	C	17
	COMMON /BLOCK7/ ALPHA,FUNCT,NIC,FACTOR,EPS,NN,NB	C	18
	COMMON /BLOCK8/ V(200),C(200)	C	19
	COMMON /BLOCK9/ CONTYPE(200),KONBRAN(200)	C	20
	COMMON /BLOCK10/ AMAG(200),APHASE(200),WSTART,WEND	C	21
	COMMON /BLOCK12/ CONTTEM(200),AMTEM(200),APH(200),KONTEM(200)	C	22
	COMMON /BLOCK14/ KP(5),W,FREQADD	C	23
	COMMON /BLOCK15/ NIT,JOUT,NGRAPH,NALLOUT	C	24
	COMMON /BLOCK16/ GRAF(200,5),JGRAPH(5),IGRAPH(5),NPRINT,SCALE,NITT	C	25
	COMMON /BLOCK19/ CNTSOR(200)	C	26
	COMMON /BLOCK20/ NCONT,KLOOP,KPP	C	27
	COMMON /BLOCK21/ NST(5)	C	28
	COMMON /BLOCK23/ CONST(200)	C	29
	COMMON /BLOCK24/ SCAL,NSCALE,NTREE	C	30
	KK=NN-1	C	31
	KKTT=0	C	32
	IF (NIC.EQ.0) GO TO 1	C	33
	GO TO 6	C	34
1	KM=0	C	35
	TEMP(1)=1HE	C	36
	TEMP(2)=1HV	C	37
	TEMP(3)=1HC	C	38
	TEMP(4)=1HR	C	39
	TEMP(5)=1HL	C	40
	TEMP(6)=1HI	C	41
	TEMP(7)=1HJ	C	42
	DO 4 K=1,7	C	43
	KT=0	C	44
	DO 3 I=1,NB	C	45
	IF (TYPE(I).EQ.TEMP(K)) GO TO 2	C	46
	GO TO 3	C	47
2	KM=KM+1	C	48
	ITBRAN(KM)=I	C	49
	TYTEMP(KM)=TEMP(K)	C	50
	VALTEM(KM)=VALUE(I)	C	51
	KT=KT+1	C	52
	CONTTEM(KM)=CONTYPE(I)	C	53
	AMTEM(KM)=AMAG(I)	C	54
	APH(KM)=APHASE(I)	C	55

	KONTEM(KM)=KONBRAN(I)	C	56
3	CONTINUE	C	57
	NOEL(K)=KT	C	58
4	CONTINUE	C	59
	DO 5 I=1,NB	C	60
	K=ITBRAN(I)	C	61
	LEAVT(I)=LEAV(K)	C	62
	LENTT(I)=LENT(K)	C	63
	TYPE(I)=TYTEMP(I)	C	64
	VALUE(I)=VALTEM(I)	C	65
	CONTYPE(I)=CONTTEM(I)	C	66
	AMAG(I)=AMTEM(I)	C	67
	APHASE(I)=APH(I)	C	68
	KONBRAN(I)=KONTEM(I)	C	69
5	CONTINUE	C	70
	MM=NB	C	71
	N=1	C	72
	GO TO 7	C	73
6	MM=NB-NOEL(7)	C	74
	N=NOEL(1)+1	C	75
7	M=MM-1	C	76
	DO 8 I=N,MM	C	77
	IF (TYPE(I).EQ.1HE) CONST(I)=-1.0	C	78
	IF (TYPE(I).EQ.1HV) CONST(I)=1.0E-50	C	79
	IF (TYPE(I).EQ.1HC) GO TO 9	C	80
	IF (TYPE(I).EQ.1HR) CONST(I)=VALUE(I)	C	81
	IF (TYPE(I).EQ.1HL) CONST(I)=VALUE(I)*W	C	82
	IF (TYPE(I).EQ.1HI) CONST(I)=1.0E+50	C	83
	IF (TYPE(I).EQ.1HJ) CONST(I)=1.0E+52	C	84
8	CONTINUE	C	85
	GO TO 10	C	86
9	A=VALUE(I)*W	C	87
	IF (A.EQ.0.0) CONST(I)=1.0E+51	C	88
	IF (A.NE.0.0) CONST(I)=1.0/A	C	89
	GO TO 8	C	90
10	KPP=0	C	91
	IF (N.GT.M) GO TO 16	C	92
	DO 15 I=N,M	C	93
	KP=0	C	94
	AMIN=CONST(I)	C	95
	K=I+1	C	96
	DO 12 J=K,MM	C	97
	IF (CONST(J).LT.AMIN) GO TO 11	C	98
	GO TO 12	C	99
11	AMIN=CONST(J)	C	100
	NP(I)=J	C	101
	KP=1	C	102
	KPP=1	C	103
12	CONTINUE	C	104
	IF (KP.EQ.0) GO TO 15	C	105
	J=NP(I)	C	106
	TEMPO1=ITBRAN(I)	C	107
	TEMPO2=LEAVT(I)	C	108

	TEMPO3=LENTT(I)	C 109
	TEMPO4=TYPE(I)	C 110
	TEMPO5=VALUE(I)	C 111
	TEMPO6=CONTYPE(I)	C 112
	TEMP20=AMAG(I)	C 113
	TEMP21=APHASE(I)	C 114
	TEMPO9=KONBRAN(I)	C 115
	TEMP10=CONST(I)	C 116
	IF (NIC.EQ.0) GO TO 13	C 117
	TEMP11=V(I)	C 118
	TEMP12=C(I)	C 119
13	ITBRAN(I)=ITBRAN(J)	C 120
	LEAVT(I)=LEAVT(J)	C 121
	LENTT(I)=LENTT(J)	C 122
	TYPE(I)=TYPE(J)	C 123
	VALUE(I)=VALUE(J)	C 124
	CONTYPE(I)=CONTYPE(J)	C 125
	AMAG(I)=AMAG(J)	C 126
	APHASE(I)=APHASE(J)	C 127
	KONBRAN(I)=KONBRAN(J)	C 128
	CONST(I)=CONST(J)	C 129
	IF (NIC.EQ.0) GO TO 14	C 130
	V(I)=V(J)	C 131
	C(I)=C(J)	C 132
14	ITBRAN(J)=TEMPO1	C 133
	LEAVT(J)=TEMPO2	C 134
	LENTT(J)=TEMPO3	C 135
	TYPE(J)=TEMPO4	C 136
	VALUE(J)=TEMPO5	C 137
	CONTYPE(J)=TEMPO6	C 138
	AMAG(J)=TEMP20	C 139
	APHASE(J)=TEMP21	C 140
	KONBRAN(J)=TEMPO9	C 141
	CONST(J)=TEMP10	C 142
	IF (NIC.EQ.0) GO TO 15	C 143
	V(J)=TEMP11	C 144
	C(J)=TEMP12	C 145
15	CONTINUE	C 146
16	IF (2.GT.KK) GO TO 32	C 147
	DO 26 I=2, KK	C 148
17	NE=LENTT(I)	C 149
	NL=LEAVT(I)	C 150
	NP(1)=I	C 151
	NNP(1)=1	C 152
	M=I-1	C 153
	MT=0	C 154
	KT=1	C 155
18	DO 20 JJJ=1, M	C 156
	J=M+1-JJJ	C 157
	IF (MT.EQ.J) GO TO 19	C 158
	IF (LEAVT(J).EQ.NE) GO TO 24	C 159
	IF (LENTT(J).EQ.NE) GO TO 25	C 160
19	CONTINUE	C 161
20	CONTINUE	C 162
21	IF (KT.EQ.1) GO TO 26	C 163

	KA=NNP(KT)	C 164
	KB=NP(KT)	C 165
	M=KB-1	C 166
	IF (M.EQ.0) GO TO 22	C 167
9	MT=NP(KT-1)	C 168
	KT=KT-1	C 169
	IF (KA.EQ.-1) GO TO 23	C 170
	NE=LEAVT(KB)	C 171
	GO TO 18	C 172
22	KT=KT-1	C 173
	GO TO 21	C 174
23	NE=LENTT(KB)	C 175
	GO TO 18	C 176
24	IF (LENTT(J).EQ.NL) GO TO 27	C 177
	KT=KT+1	C 178
	NE=LENTT(J)	C 179
	NP(KT)=J	C 180
	NNP(KT)=1	C 181
	MT=J	C 182
	M=I-1	C 183
	GO TO 18	C 184
25	IF (LEAVT(J).EQ.NL) GO TO 27	C 185
	NE=LEAVT(J)	C 186
	KT=KT+1	C 187
	NP(KT)=J	C 188
	NNP(KT)=-1	C 189
	MT=J	C 190
	M=I-1	C 191
	GO TO 18	C 192
26	CONTINUE	C 193
	GO TO 32	C 194
27	LAF=NOEL(1)	C 195
	IF (I.LE.LAF) GO TO 52	C 196
28	TEMPO1=ITBRAN(I)	C 197
	TEMPO2=LEAVT(I)	C 198
	TEMPO3=LENTT(I)	C 199
	TEMPO4=TYPE(I)	C 200
	TEMPO5=VALUE(I)	C 201
	TEMPO6=CONTYPE(I)	C 202
	TEMP20=AMAG(I)	C 203
	TEMP21=APHASE(I)	C 204
	TEMPO9=KONBRAN(I)	C 205
	IF (NIC.EQ.0) GO TO 29	C 206
	TEMP11=V(I)	C 207
	TEMP12=C(I)	C 208
29	NLB=NB-1	C 209
	IF (NIC.NE.0) NLB=NB-NOEL(7)-1	C 210
	DO 30 JN=I,NLB	C 211
	JE=JN+1	C 212
30	ITBRAN(JN)=ITBRAN(JE)	C 213
	LEAVT(JN)=LEAVT(JE)	C 214
	LENTT(JN)=LENTT(JE)	C 215
	TYPE(JN)=TYPE(JE)	C 216

	VALUE(JN)=VALUE(JE)	C 217
	CONTYPE(JN)=CONTYPE(JE)	C 218
	AMAG(JN)=AMAG(JE)	C 219
	APHASE(JN)=APHASE(JE)	C 220
	KONBRAN(JN)=KONBRAN(JE)	C 221
	IF (NIC.EQ.0) GO TO 30	C 222
	V(JN)=V(JE)	C 223
	C(JN)=C(JE)	C 224
30	CONTINUE	C 225
	NZ=NB	C 226
	IF (NIC.NE.0) NZ=NB-NOEL(7)	C 227
	ITBRAN(NZ)=TEMPO1	C 228
	LEAVT(NZ)=TEMPO2	C 229
	LENTT(NZ)=TEMPO3	C 230
	TYPE(NZ)=TEMPO4	C 231
	VALUE(NZ)=TEMPO5	C 232
	CONTYPE(NZ)=TEMPO6	C 233
	AMAG(NZ)=TEMP20	C 234
	APHASE(NZ)=TEMP21	C 235
	KONBRAN(NZ)=TEMPO9	C 236
	IF (NIC.EQ.0) GO TO 31	C 237
	V(NZ)=TEMP11	C 238
	C(NZ)=TEMP12	C 239
31	IF (KKTT.EQ.1) GO TO 62	C 240
	GO TO 17	C 241
32	CONTINUE	C 242
	IF (NIC.NE.0) GO TO 36	C 243
	DO 33 L=1, KK	C 244
	IF (TYPE(L).EQ.TEMP(7)) GO TO 53	C 245
33	CONTINUE	C 246
	I=NN-1	C 246A
	DO 35 J=NN, NB	C 247
	I=I+1	C 247A
	IF (TYPE(I).EQ.TEMP(7)) GO TO 34	C 248
	GO TO 35	C 249
34	KKTT=1	C 250
	GO TO 28	C 251
62	I=I-1	C 251A
35	CONTINUE	C 252
36	DO 37 I=1, NB	C 253
	K=ITBRAN(I)	C 254
	NP(I)=IBRAN(K)	C 255
37	CONTINUE	C 256
	DO 38 I=1, NB	C 257
	CNTSOR(I)=0	C 258
38	CONTINUE	C 259
	IF (JOUT.EQ.0) GO TO 41	C 260
	DO 40 I=1, JOUT	C 261
	K=0	C 262
39	K=K+1	C 263
	IF (IOUT(I).NE.NP(K)) GO TO 39	C 264
	NOUT(I)=K	C 265
40	CONTINUE	C 266
41	IF (NGRAPH.EQ.0) GO TO 44	C 267
	DO 43 I=1, NGRAPH	C 268

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K=0
42 K=K+1
   IF (IGRAPH(I).NE.NP(K)) GO TO 42
   NST(I)=K
43 CONTINUE
44 N=NOEL(1)+1
   M=NB-NOEL(7)
   DO 50 I=N,M
   IF (TYPE(I).EQ.1HI) GO TO 45
   IF (TYPE(I).EQ.1HV) GO TO 45
   GO TO 50
45 K=KONBRAN(I)
   L=0
46 L=L+1
   IF (L.GT.NOEL(1)) GO TO 47
   IF (NP(L).NE.K) GO TO 46
   GO TO 49
47 L=NB-NOEL(7)
48 L=L+1
   IF (NP(L).NE.K) GO TO 48
49 CNTSOR(I)=L
   CNTSOR(L)=I
50 CONTINUE
   IF (NCONT.EQ.0) GO TO 54
   PRINT 56
   PRINT 57, ((I,NP(I)),I=1,KK)
   PRINT 58, ((I,NP(I)),I=NN,NB)
   PRINT 59
   DO 51 I=1,NB
   PRINT 55, NP(I),TYPE(I),VALUE(I),LEAVT(I),LENTT(I)
51 CONTINUE
   RETURN
52 PRINT 60
   KLOOP=1
   RETURN
53 PRINT 61
   KLOOP=1
54 RETURN
C
C
55 FORMAT (1H0,10X,*BRANCH NUMBER *,5X,I3,4X,* IS A *,A1,* OF
1 VALUE *,E12.5,* LEAVING NODE *,I3,* AND ENTERING NODE *,I3)
56 FORMAT (1H1,20X,*CORRESPONDENCE BETWEEN ORIGINAL TOPOLOGY AND NEW
1 TOPOLOGY*,///)
57 FORMAT (1H0,30X,*TREE BRANCH*,I4,5X,*CORRESPONDS TO BRANCH *,I4)
58 FORMAT (1H0,33X,* LINK *,2X,I4,5X,*CORRESPONDS TO BRANCH *,I4)
59 FORMAT (1H1)
12X,* COND = *,E10.3)
60 FORMAT (1H0,30X,*THERE IS A VOLTAGE LOOP IN NETWORK*)
61 FORMAT (1H0,30X,*THERE IS A CURRENT SOURCE CUTSET IN NETWORK*)
END

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C 310
C 311
C 312
C 313
C 314
C 315
C 316
C 317
C 318
C 319-

	SUBROUTINE FCSM	D	1
	INTEGER TYPE	D	2
	COMMON /BLOCK1/ NNP(200),NP(200)	D	3
	COMMON /BLOCK3/ TYPE(200),VALUE(200),F(49,151),NOEL(7)	D	4
	COMMON /BLOCK4/ ITBRAN(200),LEAVT(200),LENTT(200)	D	5
	COMMON /BLOCK7/ ALPHA,FUNCT,NIC,FACTOR,EPS,NN,NB	D	6
	COMMON /BLOCK20/ NCONT,KLOOP,KPP	D	7
C	SET MATRIX TO ZERO	D	8
	KK=NN-1	D	9
	NLB=NB-NN+1	D	10
	DO 1 I=1,NLB	D	11
	DO 1 J=1,KK	D	12
	F(J,I)=0.0	D	13
1	CONTINUE	D	14
	DO 8 I=NN,NB	D	15
	NE=LENTT(I)	D	16
	NL=LEAVT(I)	D	17
	IO=I-NN+1	D	18
C	NP STORES THE PREVIOUS TREE BRANCH	D	19
C	NNP STORES ITS DIRECTION	D	20
	NP(1)=I	D	21
	NNP(1)=1	D	22
	M=1	D	23
	KT=1	D	24
2	DO 3 J=M,KK	D	25
	IF (J.EQ.NP(KT)) GO TO 3	D	26
	IF (LEAVT(J).EQ.NE) GO TO 6	D	27
	IF (LENTT(J).EQ.NE) GO TO 7	D	28
3	CONTINUE	D	29
4	M=NP(KT)+1	D	30
	KA>NNP(KT)	D	31
	KB=NP(KT)	D	32
	F(KB,IO)=0.0	D	33
	KT=KT-1	D	34
	IF (M.GE.NN) GO TO 4	D	35
	IF (KA.EQ.1) GO TO 5	D	36
	NE=LENTT(KB)	D	37
	GO TO 2	D	38
5	NE=LEAVT(KB)	D	39
	GO TO 2	D	40
6	F(J,IO)=-1.0	D	41
	KT=KT+1	D	42
	IF (LENTT(J).EQ.NL) GO TO 8	D	43
	NE=LENTT(J)	D	44
	NP(KT)=J	D	45
	NNP(KT)=1	D	46
	M=1	D	47
	GO TO 2	D	48
7	F(J,IO)=1.0	D	49
	KT=KT+1	D	50
	IF (LEAVT(J).EQ.NL) GO TO 8	D	51
	NE=LEAVT(J)	D	52
	NP(KT)=J	D	53
	NNP(KT)=-1	D	54
	M=1	D	55

	GO TO 2	D	56
8	CONTINUE	D	57
	IF (NCONT.EQ.0) GO TO 16	D	58
	NLB=NB-NN+1	D	59
	KK=NN-1	D	60
	NCOUNT=-1	D	61
	NT=NLB	D	62
	NPAGE=1	D	63
	PRINT 20, NPAGE	D	64
	PRINT 19	D	65
	PRINT 18, (J,J=1,NLB)	D	66
9	NT=NT-25	D	67
	NCOUNT=NCOUNT+1	D	68
	NS1=1+25*NCOUNT	D	69
	NS2=25+25*NCOUNT	D	70
	IF (NT.LE.0) GO TO 11	D	71
	DO 10 I=1,KK	D	72
	PRINT 17, I,(F(I,J),J=NS1,NS2)	D	73
10	CONTINUE	D	74
	NPAGE=NPAGE+1	D	75
	GO TO 9	D	76
11	IF (NT.EQ.0) GO TO 12	D	77
	GO TO 13	D	78
12	NS3=1	D	79
	NS4=25	D	80
	GO TO 14	D	81
13	NT=NT+25	D	82
	NS3=NS1	D	83
	NS4=NS1+NT-1	D	84
14	CONTINUE	D	85
	DO 15 I=1,KK	D	86
	PRINT 17, I,(F(I,J),J=NS3,NS4)	D	87
15	CONTINUE	D	88
16	RETURN	D	89
C		D	90
C		D	91
17	FORMAT (1H0,I3,3X,25(2X,F3.0))	D	92
18	FORMAT (1H0,7X,25(2X,I2,1X))	D	93
19	FORMAT (///)	D	94
20	FORMAT (1H1,30X,* F PORTION OF FUNDAMENTAL CUTSET MATRIX*,5X,*PAG	D	95
	1E*,I5,///)	D	96
	END	D	97-

	SUBROUTINE VOLCUR (I,D)	E	1
	COMPLEX D	E	2
3	COMMON /BLOCK10/ AMAG(200),APHASE(200),WSTART,WEND	E	3
	PI=180.0/3.14159265	E	4
	IF (APHASE(I).LT.-180.) APHASE(I)=360.0+APHASE(I)	E	5
	IF (APHASE(I).GT.180.) APHASE(I)=-360.+APHASE(I)	E	6
	IF (APHASE(I)) 1,5,6	E	7
1	IF (APHASE(I)+90.0) 2,3,4	E	8
2	TH=(180.0+APHASE(I))/PI	E	9
	A=-COS(TH)*AMAG(I)	E	10
	B=-SIN(TH)*AMAG(I)	E	11
	D=CMPLX(A,B)	E	12
	RETURN	E	13
3	B=-AMAG(I)	E	14
	D=CMPLX(0.0,B)	E	15
	RETURN	E	16
4	TH=-APHASE(I)/PI	E	17
	A=COS(TH)*AMAG(I)	E	18
	B=-SIN(TH)*AMAG(I)	E	19
	D=CMPLX(A,B)	E	20
	RETURN	E	21
5	A=AMAG(I)	E	22
	D=CMPLX(A,0.0)	E	23
	RETURN	E	24
6	IF (APHASE(I)-90.0) 7,8,9	E	25
7	TH=APHASE(I)/PI	E	26
	A=COS(TH)*AMAG(I)	E	27
	B=SIN(TH)*AMAG(I)	E	28
	D=CMPLX(A,B)	E	29
	RETURN	E	30
8	B=AMAG(I)	E	31
	D=CMPLX(0.0,B)	E	32
	RETURN	E	33
9	TH=(180.0-APHASE(I))/PI	E	34
	A=-AMAG(I)*COS(TH)	E	35
	B=AMAG(I)*SIN(TH)	E	36
	D=CMPLX(A,B)	E	37
	RETURN	E	38
	END	E	39-

	SUBROUTINE INITIAL	F	1
	COMPLEX V,C	F	2
	COMPLEX GRAF	F	3
	INTEGER TYPE	F	4
	COMMON /BLOCK3/ TYPE(200),VALUE(200),F(49,151),NOEL(7)	F	5
	COMMON /BLOCK7/ ALPHA,FUNCT,NIC,FACTOR,EPS,NN,NB	F	6
	COMMON /BLOCK8/ V(200),C(200)	F	7
	COMMON /BLOCK10/ AMAG(200),APHASE(200),WSTART,WEND	F	8
	COMMON /BLOCK16/ GRAF(200,5),JGRAPH(5),IGRAPH(5),NPRINT,SCALE,NITT	F	9
	N=NN-1	F	10
	IF (NIC.EQ.0) GO TO 4	F	11
	IF (WSTART.NE.0.0) RETURN	F	12
1	DO 2 I=1,N	F	13
	V(I)=CMPLX(0.0,0.0)	F	14
	IF (TYPE(I).EQ.1HE) CALL VOLCUR (I,V(I))	F	15
2	CONTINUE	F	16
	DO 3 I=NN,NB	F	17
	C(I)=CMPLX(0.0,0.0)	F	18
	IF (TYPE(I).EQ.1HJ) CALL VOLCUR (I,C(I))	F	19
	IF (NIC.GT.0) C(I)=C(I)*SCALE	F	20
3	CONTINUE	F	21
	RETURN	F	22
4	IF (WSTART.NE.0.0) GO TO 1	F	23
	DO 5 I=1,N	F	24
	V(I)=CMPLX(0.0,0.0)	F	25
	IF (TYPE(I).EQ.1HE) V(I)=CMPLX(AMAG(I),0.0)	F	26
5	CONTINUE	F	27
	DO 6 I=NN,NB	F	28
	C(I)=CMPLX(0.0,0.0)	F	29
	IF (TYPE(I).EQ.1HJ) C(I)=CMPLX(AMAG(I),0.0)	F	30
6	CONTINUE	F	31
	RETURN	F	32
	END	F	33-

SUBROUTINE CALCAL

COMPLEX E,CONST,V,C,GRAD

INTEGER TYPE

COMMON /BLOCK3/ TYPE(200),VALUE(200),F(49,151),NOEL(7)

COMMON /BLOCK7/ ALPHA,FUNCT,NIC,FACTOR,EPS,NN,NB

COMMON /BLOCK8/ V(200),C(200)

N=NN-1

DO 1 I=1,N

C(I)=CMPLX(0.0,0.0)

DO 1 J=NN,NB

C(I)=C(I)-F(I,J-N)*C(J)

CONTINUE

DO 2 I=NN,NB

V(I)=CMPLX(0.0,0.0)

DO 2 J=1,N

V(I)=V(I)+F(J,I-N)*V(J)

CONTINUE

RETURN

END

G 1
G 2
G 3
G 4
G 5
G 6
G 7
G 8
G 9
G 10
G 11
G 12
G 13
G 14
G 15
G 16
G 17
G 18
G 19-

	SUBROUTINE KONST	H	1
C	CALCULATION OF CONSTANTS RELATED TO ELEMENT VALUES FOR USE IN BRAN	H	2
C	RELATIONS AND GRADIENT	H	3
	COMPLEX E,CONST,V,C,GRAD	H	4
	COMPLEX GRAF	H	5
	INTEGER CONTYPE	H	6
	INTEGER TEMP	H	7
	INTEGER TYPE	H	8
	COMMON /BLOCK3/ TYPE(200),VALUE(200),F(49,151),NOEL(7)	H	9
	COMMON /BLOCK6/ TEMP(7),E(200),GRAD(200),CONST(200)	H	10
	COMMON /BLOCK7/ ALPHA,FUNCT,NIC,FACTOR,EPS,NN,NB	H	11
	COMMON /BLOCK8/ V(200),C(200)	H	12
	COMMON /BLOCK9/ CONTYPE(200),KONBRAN(200)	H	13
	COMMON /BLOCK14/ KP(5),W,FREQADD	H	14
	COMMON /BLOCK16/ GRAF(200,5),JGRAPH(5),IGRAPH(5),NPRINT,SCALE,NITT	H	15
	COMMON /BLOCK24/ SCAL,NSCALE,NTREE	H	16
	SCAL=SCALE	H	17
	IF (NSCALE.EQ.1) GO TO 3	H	18
	KW=0	H	19
	DO 2 I=NN,NB	H	20
	IF (TYPE(I).EQ.1HJ) GO TO 2	H	21
	IF (TYPE(I).EQ.1HV) GO TO 2	H	22
	IF (TYPE(I).EQ.1HI) GO TO 2	H	23
	IF (TYPE(I).EQ.1HC) A=W*VALUE(I)	H	24
	IF (TYPE(I).EQ.1HR) A=1.0/VALUE(I)	H	25
	IF (W.EQ.0.0) GO TO 1	H	26
	IF (TYPE(I).EQ.1HL) A=1.0/(W*VALUE(I))	H	27
1	IF (KW.EQ.0) SCALE=A	H	28
	KW=1	H	29
	IF (A.GT.SCALE) SCALE=A	H	30
2	CONTINUE	H	31
	IF (KW.EQ.0) SCALE=1.0	H	32
	IF (SCALE.EQ.0.0) SCALE=1.0	H	33
	SCALE=1.0/SCALE	H	34
3	N=NN-1	H	35
	DO 9 I=1,N	H	36
	IF (TYPE(I).EQ.1HV) GO TO 4	H	37
	IF (TYPE(I).EQ.1HC) GO TO 5	H	38
	IF (TYPE(I).EQ.1HR) GO TO 6	H	39
	IF (TYPE(I).EQ.1HL) GO TO 7	H	40
	IF (TYPE(I).EQ.1HI) GO TO 8	H	41
	GO TO 9	H	42
4	CONST(I)=CMPLX(VALUE(I),0.0)	H	43
	IF (CONTYPE(I).EQ.1HI) CONST(I)=CONST(I)/SCALE	H	44
	GO TO 9	H	45
5	IF (W.EQ.0.0) GO TO 9	H	46
	A=-1.0/(VALUE(I)*SCALE*W)	H	47
	CONST(I)=CMPLX(0.0,A)	H	48
	GO TO 9	H	49
6	A=VALUE(I)/SCALE	H	50
	CONST(I)=CMPLX(A,0.0)	H	51
	GO TO 9	H	52
7	A=VALUE(I)*W/SCALE	H	53
	CONST(I)=CMPLX(0.0,A)	H	54
	GO TO 9	H	55

8	CONST(I)=CMPLX(VALUE(I),0.0)	H	56
	IF (CONTYPE(I).EQ.1HV) CONST(I)=CONST(I)*SCALE	H	57
9	CONTINUE	H	58
	DO 15 I=NN,NB	H	59
	IF (TYPE(I).EQ.1HV) GO TO 10	H	60
	IF (TYPE(I).EQ.1HC) GO TO 11	H	61
	IF (TYPE(I).EQ.1HR) GO TO 12	H	62
	IF (TYPE(I).EQ.1HL) GO TO 13	H	63
	IF (TYPE(I).EQ.1HI) GO TO 14	H	64
	GO TO 15	H	65
10	CONST(I)=CMPLX(VALUE(I),0.0)	H	66
	IF (CONTYPE(I).EQ.1HI) CONST(I)=CONST(I)/SCALE	H	67
	GO TO 15	H	68
11	A=VALUE(I)*SCALE*W	H	69
	CONST(I)=CMPLX(0.0,A)	H	70
	GO TO 15	H	71
12	A=SCALE/VALUE(I)	H	72
	CONST(I)=CMPLX(A,0.0)	H	73
	GO TO 15	H	74
13	IF (W.EQ.0.0) GO TO 15	H	75
	A=-SCALE/(VALUE(I)*W)	H	76
	CONST(I)=CMPLX(0.0,A)	H	77
	GO TO 15	H	78
14	CONST(I)=CMPLX(VALUE(I),0.0)	H	79
	IF (CONTYPE(I).EQ.1HV) CONST(I)=CONST(I)*SCALE	H	80
15	CONTINUE	H	81
	RETURN	H	82
	END	H	83-

```

SUBROUTINE ERROR
INTEGER CNTSOR
COMPLEX E,CONST,V,C,GRAD
INTEGER TEMP
INTEGER CONTYPE
INTEGER TYPE
COMMON /BLOCK3/ TYPE(200),VALUE(200),F(49,151),NOEL(7)
COMMON /BLOCK6/ TEMP(7),E(200),GRAD(200),CONST(200)
COMMON /BLOCK7/ ALPHA,FUNCT,NIC,FACTOR,EPS,NN,NB
COMMON /BLOCK8/ V(200),C(200)
COMMON /BLOCK9/ CONTYPE(200),KONBRAN(200)
COMMON /BLOCK14/ KP(5),W,FREQADD
COMMON /BLOCK19/ CNTSOR(200)
M=NOEL(1)+1
N=NB-NOEL(7)
DO 10 I=M,N
IF (TYPE(I).EQ.1HV) GO TO 8
IF (TYPE(I).EQ.1HI) GO TO 6
IF (I.GE.NN) GO TO 3
IF (W.EQ.0.0) GO TO 2
1 E(I)=-V(I)+C(I)*CONST(I)
GO TO 10
2 IF (TYPE(I).NE.1HC) GO TO 1
E(I)=CMPLX(0.0,0.0)
V(I)=CMPLX(0.0,0.0)
GO TO 10
3 IF (W.EQ.0.0) GO TO 5
4 E(I)=-C(I)+V(I)*CONST(I)
GO TO 10
5 IF (TYPE(I).NE.1HL) GO TO 4
E(I)=CMPLX(0.0,0.0)
C(I)=CMPLX(0.0,0.0)
GO TO 10
6 K=CNTSOR(I)
IF (CONTYPE(I).NE.1HV) GO TO 7
E(I)=-C(I)+V(K)*CONST(I)
GO TO 10
7 E(I)=-C(I)+C(K)*CONST(I)
GO TO 10
8 K=CNTSOR(I)
IF (CONTYPE(I).NE.1HV) GO TO 9
E(I)=-V(I)+V(K)*CONST(I)
GO TO 10
9 E(I)=-V(I)+C(K)*CONST(I)
10 CONTINUE
FUNCT=0.0
DO 11 I=M,N
FUNCT=FUNCT+E(I)*CONJG(E(I))
11 CONTINUE
FUNCT=FUNCT/2
RETURN
END
SUBROUTINE CNGVARS
C THIS SUBROUTINE CHANGES THE CURRENTS ALONG THE GRADIENT BY THE
C CONSTANT ALPHA

```

```

I 1
I 2
I 3
I 4
I 5
I 6
I 7
I 8
I 9
I 10
I 11
I 12
I 13
I 14
I 15
I 16
I 17
I 18
I 19
I 20
I 21
I 22
I 23
I 24
I 25
I 26
I 27
I 28
I 29
I 30
I 31
I 32
I 33
I 34
I 35
I 36
I 37
I 38
I 39
I 40
I 41
I 42
I 43
I 44
I 45
I 46
I 47
I 48
I 49
I 50
I 51
I 52-
J 1
J 2
J 3

```

```
COMPLEX V,C
COMPLEX CONT,S
INTEGER TYPE
COMMON /BLOCK3/ TYPE(200),VALUE(200),F(49,151),NOEL(7)
COMMON /BLOCK7/ ALPHA,FUNCT,NIC,FACTOR,EPS,NN,NB
COMMON /BLOCK8/ V(200),C(200)
COMMON /BLOCK11/ CONT(200),S(200)
M=NOEL(1)+1
N=NB-NOEL(7)
IEND=NN-1
IF (M.GT.IEND) GO TO 2
DO 1 I=M,IEND
V(I)=V(I)-ALPHA*S(I)
CONTINUE
1 IF (NN.GT.N) RETURN
DO 3 I=NN,N
C(I)=C(I)-ALPHA*S(I)
CONTINUE
3 RETURN
END
```

```
J 4
J 5
J 6
J 7
J 8
J 9
J 10
J 11
J 12
J 13
J 14
J 15
J 16
J 17
J 18
J 19
J 20
J 21
J 22
J 23-
```

	SUBROUTINE FNDGRAD	K	1
	• THIS SUBROUTINE CALCULATES THE GRADIENT VECTOR	K	2
C	INTEGER CNTSOR	K	3
	COMPLEX TEMPOR,DUMMY	K	4
	COMPLEX E,CONST,V,C,GRAD	K	5
	INTEGER CONTYPE	K	6
	INTEGER TEMP	K	7
	INTEGER TYPE	K	8
	COMMON /BLOCK3/ TYPE(200),VALUE(200),F(49,151),NOEL(7)	K	9
	COMMON /BLOCK6/ TEMP(7),E(200),GRAD(200),CONST(200)	K	10
	COMMON /BLOCK7/ ALPHA,FUNCT,NIC,FACTOR,EPS,NN,NB	K	11
	COMMON /BLOCK8/ V(200),C(200)	K	12
	COMMON /BLOCK9/ CONTYPE(200),KONBRAN(200)	K	13
	COMMON /BLOCK14/ KP(5),W,FREQADD	K	14
	COMMON /BLOCK19/ CNTSOR(200)	K	15
	COMMON /BLOCK23/ TEMPOR(200),DUMMY(200)	K	16
	M=NOEL(1)+1	K	17
	MM=NB-NOEL(7)	K	18
	N=NN-1	K	19
	DO 7 I=1,N	K	20
	IF (TYPE(I).EQ.1HE) GO TO 3	K	21
	IF (TYPE(I).EQ.1HI) GO TO 5	K	22
	IF (TYPE(I).EQ.1HV) GO TO 6	K	23
	IF (W.EQ.0.0) GO TO 2	K	24
1	TEMPOR(I)=-E(I)	K	25
	A=REAL(E(I))*REAL(CONST(I))+AIMAG(E(I))*AIMAG(CONST(I))	K	26
	B=-REAL(E(I))*AIMAG(CONST(I))+AIMAG(E(I))*REAL(CONST(I))	K	27
	DUMMY(I)=CMPLX(A,B)	K	28
	GO TO 7	K	29
2	IF (TYPE(I).NE.1HC) GO TO 1	K	30
	TEMPOR(I)=CMPLX(0.0,0.0)	K	31
	DUMMY(I)=CMPLX(0.0,0.0)	K	32
	GO TO 7	K	33
3	K=CNTSOR(I)	K	34
	IF (K.EQ.0) GO TO 4	K	35
	TEMPOR(I)=CMPLX(0.0,0.0)	K	36
	DUMMY(I)=CONST(K)*E(K)	K	37
	GO TO 7	K	38
4	TEMPOR(I)=CMPLX(0.0,0.0)	K	39
	DUMMY(I)=CMPLX(0.0,0.0)	K	40
	GO TO 7	K	41
5	TEMPOR(I)=CMPLX(0.0,0.0)	K	42
	DUMMY(I)=-E(I)	K	43
	GO TO 7	K	44
6	TEMPOR(I)=-E(I)	K	45
	DUMMY(I)=CMPLX(0.0,0.0)	K	46
7	CONTINUE	K	47
	DO 14 I=NN,NB	K	48
	IF (TYPE(I).EQ.1HJ) GO TO 10	K	49
	IF (TYPE(I).EQ.1HV) GO TO 12	K	50
	IF (TYPE(I).EQ.1HI) GO TO 13	K	51
	IF (W.EQ.0.0) GO TO 9	K	52
8	DUMMY(I)=-E(I)	K	53
	A=REAL(E(I))*REAL(CONST(I))+AIMAG(E(I))*AIMAG(CONST(I))	K	54
	B=-REAL(E(I))*AIMAG(CONST(I))+AIMAG(E(I))*REAL(CONST(I))	K	55

	TEMPAR(I)=CMPLX(A,B)	K	56
	GO TO 14	K	57
9	IF (TYPE(I).NE.1HL) GO TO 8	K	58
	TEMPAR(I)=CMPLX(0.0,0.0)	K	59
	DUMMY(I)=CMPLX(0.0,0.0)	K	60
	GO TO 14	K	61
10	K=CNTSOR(I)	K	62
	IF (K.EQ.0) GO TO 11	K	63
	TEMPAR(I)=CONST(K)*E(K)	K	64
	DUMMY(I)=CMPLX(0.0,0.0)	K	65
	GO TO 14	K	66
11	TEMPAR(I)=CMPLX(0.0,0.0)	K	67
	DUMMY(I)=CMPLX(0.0,0.0)	K	68
	GO TO 14	K	69
12	TEMPAR(I)=-E(I)	K	70
	DUMMY(I)=CMPLX(0.0,0.0)	K	71
	GO TO 14	K	72
13	TEMPAR(I)=CMPLX(0.0,0.0)	K	73
	DUMMY(I)=-E(I)	K	74
14	CONTINUE	K	75
	IF (M.GT.N) GO TO 16	K	76
	DO 15 I=M,N	K	77
	GRAD(I)=TEMPAR(I)	K	78
	DO 15 J=NN,NB	K	79
15	GRAD(I)=GRAD(I)+F(I,J-N)*TEMPAR(J)	K	80
16	IF (NN.GT.MM) GO TO 18	K	81
	DO 17 I=NN,MM	K	82
	GRAD(I)=DUMMY(I)	K	83
	DO 17 J=1,N	K	84
17	GRAD(I)=GRAD(I)-F(J,I-N)*DUMMY(J)	K	85
18	IF (W.EQ.0.0) GO TO 19	K	86
	RETURN	K	87
19	IF (M.GT.N) GO TO 21	K	88
	DO 20 I=M,N	K	89
	IF (TYPE(I).EQ.1HC) GRAD(I)=CMPLX(0.0,0.0)	K	90
20	CONTINUE	K	91
21	IF (NN.GT.MM) RETURN	K	92
	DO 22 I=NN,MM	K	93
	IF (TYPE(I).EQ.1HL) GRAD(I)=CMPLX(0.0,0.0)	K	94
22	CONTINUE	K	95
	RETURN	K	96
	END	K	97-

	SUBROUTINE FNDALPH	L	1
	THIS SUBROUTINE FINDS THE DISTANCE WE SHOULD GO ALONG THE GRADIENT	L	2
	COMPLEX E,CONST,V,C,GRAD	L	3
	COMPLEX CONT,S	L	4
	INTEGER TYPE	L	5
	INTEGER TEMP	L	6
	COMMON /BLOCK3/ TYPE(200),VALUE(200),F(49,151),NOEL(7)	L	7
	COMMON /BLOCK6/ TEMP(7),E(200),GRAD(200),CONST(200)	L	8
	COMMON /BLOCK7/ ALPHA,FUNCT,NIC,FACTOR,EPS,NN,NB	L	9
	COMMON /BLOCK8/ V(200),C(200)	L	10
	COMMON /BLOCK11/ CONT(200),S(200)	L	11
	FZERO=FUNCT	L	12
	M=NOEL(1)+1	L	13
	N=NB-NOEL(7)	L	14
	MM=NN-1	L	15
	PROD=0.0	L	16
	IF (M.GT.MM) GO TO 2	L	17
	DO 1 I=M,MM	L	18
	PROD=PROD+REAL(GRAD(I))*REAL(S(I))+AIMAG(GRAD(I))*AIMAG(S(I))	L	19
	CONT(I)=V(I)	L	20
	V(I)=V(I)-S(I)	L	21
1	CONTINUE	L	22
2	IF (NN.GT.N) GO TO 4	L	23
	DO 3 I=NN,N	L	24
	PROD=PROD+REAL(GRAD(I))*REAL(S(I))+AIMAG(GRAD(I))*AIMAG(S(I))	L	25
	CONT(I)=C(I)	L	26
	C(I)=C(I)-S(I)	L	27
3	CONTINUE	L	28
4	CALL CALCAL	L	29
	CALL ERROR	L	30
	FGRAD=FUNCT	L	31
	IF (M.GT.MM) GO TO 6	L	32
	DO 5 I=M,MM	L	33
	V(I)=CONT(I)	L	34
5	CONTINUE	L	35
6	IF (NN.GT.N) GO TO 8	L	36
	DO 7 I=NN,N	L	37
	C(I)=CONT(I)	L	38
7	CONTINUE	L	39
8	A=PROD+FGRAD-FZERO	L	40
	IF (A.EQ.0.0) A=1.0E-10	L	41
	ALPHA=PROD/A	L	42
	ALPHA=ALPHA/2.0	L	43
	RETURN	L	44
	END	L	45-

	SUBROUTINE FLPOW (HH,NB,NB2)	M	1
	COMPLEX CONT,S	M	2
	COMPLEX E,GRAD,CONST	M	3
	COMPLEX GRADR	M	4
	INTEGER TEMP	M	5
	INTEGER TYPE	M	6
	COMMON /BLOCK3/ TYPE(200),VALUE(200),F(49,151),NOEL(7)	M	7
	COMMON /BLOCK6/ TEMP(7),E(200),GRAD(200),CONST(200)	M	8
	COMMON /BLOCK7/ ALPHA	M	9
	COMMON /BLOCK11/ CONT(200),S(200)	M	10
	COMMON /BLOCK13/ GRADR(200),DUMMY(400),ITN	M	11
	COMMON /BLOCK23/ SS(400),GB(400)	M	12
	DIMENSION HH(NB2,NB2)	M	13
	IF (ITN.GT.1) GO TO 3	M	14
	M=NOEL(1)	M	15
	NB1=NB2/2	M	16
	DO 2 I=1,NB1	M	17
	II=I+NB1	M	18
	GRADR(I+M)=GRAD(I+M)	M	19
	DO 1 J=1,NB2	M	20
	HH(I,J)=0.0	M	21
1	HH(II,J)=0.0	M	22
	HH(II,II)=1.0	M	23
	HH(I,II)=1.0	M	24
2	S(I+M)=GRAD(I+M)	M	25
	RETURN	M	26
3	GHG=0.0	M	27
	SG=0.0	M	28
	DO 4 I=1,NB1	M	29
	II=I+NB1	M	30
	L=I+M	M	31
	SS(I)=REAL(S(L))*ALPHA	M	32
	SS(II)=AIMAG(S(L))*ALPHA	M	33
	GB(I)=REAL(GRAD(L))-REAL(GRADR(L))	M	34
	GB(II)=AIMAG(GRAD(L))-AIMAG(GRADR(L))	M	35
4	SG=SG+SS(I)*GB(I)+SS(II)*GB(II)	M	36
	DO 5 I=1,NB2	M	37
	DUMMY(I)=0.0	M	38
	DO 5 J=1,NB2	M	39
	GHG=GHG+GB(I)*GB(J)*HH(I,J)	M	40
5	DUMMY(I)=DUMMY(I)+HH(I,J)*GB(J)	M	41
	DO 6 I=1,NB2	M	42
	DO 6 J=I,NB2	M	44*
	HH(I,J)=HH(I,J)-SS(I)*SS(J)/SG-DUMMY(I)*DUMMY(J)/GHG	M	45
6	HH(J,I)=HH(I,J)	M	46
	DO 7 I=1,NB1	M	47
	II=I+NB1	M	48
	L=I+M	M	49
	GB(I)=REAL(GRAD(L))	M	50
	GB(II)=AIMAG(GRAD(L))	M	51
	DO 8 I=1,NB2	M	52
	SS(I)=0.0	M	53A
	DO 8 J=1,NB2	M	53B
8	SS(I)=SS(I)+HH(I,J)*GB(J)	M	54
	DO 9 I=1,NB1	M	55

```
II=I+NB1  
GRADB(I+M)=GRAD(I+M)  
A=SS(I)  
B=SS(II)  
S(I+M)=CMPLX(A,B)  
RETURN  
END
```

```
M 56  
M 57  
M 58  
M 59  
M 60  
M 61  
M 62-
```

	SUBROUTINE READOUT	N	1
	COMPLEX V,C	N	2
	COMPLEX VAROUT	N	3
	COMPLEX GRAF	N	4
	INTEGER TYPE	N	5
	COMMON /BLOCK3/ TYPE(200),VALUE(200),F(49,151),NOEL(7)	N	6
	COMMON /BLOCK5/ IOUT(200),ITEST(200),NOUT(200),ITTEST(200)	N	7
	COMMON /BLOCK7/ ALPHA,FUNCT,NIC,FACTOR,EPS,NN,NB	N	8
	COMMON /BLOCK8/ V(200),C(200)	N	9
	COMMON /BLOCK10/ AMAG(200),APHASE(200),WSTART,WEND,NRES	N	10
	COMMON /BLOCK14/ KP(5),W,FREQADD	N	11
	COMMON /BLOCK15/ NIT,JOUT,NGRAPH,NALLOUT	N	12
	COMMON /BLOCK16/ GRAF(200,5),JGRAPH(5),IGRAPH(5),NPRINT,SCALE,NITT	N	13
	COMMON /BLOCK23/ VAROUT(200),VABS(200),VPHASE(200)	N	14
	PI=180./3.14159265	N	15
	N=NN-1	N	16
	DO 7 I=1,JOUT	N	17
	K=NOUT(I)	N	18
	IF (NOUT(I).LT.NN) GO TO 5	N	19
	IF (ITEST(I).EQ.1) GO TO 1	N	20
	VAROUT(I)=V(K)	N	21
	GO TO 2	N	22
1	VAROUT(I)=C(K)/SCALE	N	23
2	VABS(I)=SQRT(VAROUT(I)*CONJG(VAROUT(I)))	N	24
	IF (REAL(VAROUT(I)).EQ.0.0) GO TO 4	N	25
	VPHASE(I)=ATAN(AIMAG(VAROUT(I))/REAL(VAROUT(I)))*PI	N	26
	IF (VPHASE(I).LE.0.0.AND.REAL(VAROUT(I)).LT.0.0) GO TO 3	N	27
	IF (VPHASE(I).GT.0.0.AND.REAL(VAROUT(I)).LT.0.0) VPHASE(I)=VPHASE(I)	N	28
	1I)-180.0	N	29
	GO TO 7	N	30
3	VPHASE(I)=VPHASE(I)+180.0	N	31
	GO TO 7	N	32
4	VPHASE(I)=90.0	N	33
	IF (AIMAG(VAROUT(I)).LT.0.0) VPHASE(I)=-VPHASE(I)	N	34
	GO TO 7	N	35
5	IF (ITEST(I).EQ.0) GO TO 6	N	36
	VAROUT(I)=C(K)/SCALE	N	37
	GO TO 2	N	38
6	VAROUT(I)=V(K)	N	39
	GO TO 2	N	40
7	IF (VABS(I).EQ.0.0) VPHASE(I)=0.0	N	41
	MRES=9H	N	42
	IF (NRES.EQ.1) MRES=9HRESONANCE	N	43
	PRINT 9, W,MRES	N	44
	DO 8 I=1,JOUT	N	45
	A=7HVOLTAGE	N	46
	IF (ITEST(I).EQ.1) A=7HCURRENT	N	47
	PRINT 10, IOUT(I),A,VABS(I),VPHASE(I)	N	48
8	CONTINUE	N	49
	RETURN	N	50
C		N	51
C		N	52
9	FORMAT (/////,10X,*FREQUENCY = *,E15.5,5X,A9)	N	53
10	FORMAT (1H0,10X,*BRANCH*,5X,I5,3X ,A7,10X,*ABS VALUE = *,E15.5,	N	54
	110X,* PHASE = *,E15.5)	N	55

END

	SUBROUTINE ALLOUT	0	1
	COMPLEX V,C	0	2
	COMPLEX A	0	3
	COMPLEX GRAF	0	4
	COMMON /BLOCK7/ ALPHA,FUNCT,NIC,FACTOR,EPS,NN,NB	0	5
	COMMON /BLOCK8/ V(200),C(200)	0	6
	COMMON /BLOCK14/ KP(5),W,FREQADD	0	7
	COMMON /BLOCK16/ GRAF(200,5),JGRAPH(5),IGRAPH(5),NPRINT,SCALE,NITT	0	8
	COMMON /BLOCK23/ A(200)	0	9
	N=NN-1	0	10
	PRINT 2, W	0	11
	DO 1 I=NN,NB	0	12
	A(I)=C(I)/SCALE	0	13
1	CONTINUE	0	14
	PRINT 3, (V(I),I=1,N)	0	15
	PRINT 4, (A(I),I=NN,NB)	0	16
	RETURN	0	17
C		0	18
C		0	19
2	FORMAT (1H0,///,* FREQUENCY = *,E15.5)	0	20
3	FORMAT(1H0,*THESE ARE THE TREE BRANCH VOLTAGES*,//,10(3X,E12.5))	0	21
4	FORMAT (1H0, * THESE ARE THE LINK CURRENTS *,//,10(3X,E12.5))	0	22
	END	0	23-

	SUBROUTINE GRAPH	P	1
	COMPLEX E,CONST,V,C,GRAD	P	2
	COMPLEX GRAF	P	3
	INTEGER ANP	P	4
	COMMON /BLOCK1/ ANP(200),NP(200)	P	5
	COMMON /BLOCK5/ FR(200),KR(200),NOUT(200),ITTEST(200)	P	6
	COMMON /BLOCK7/ ALPHA,FUNCT,NIC,FACTOR,EPS,NN,NB	P	7
	COMMON /BLOCK8/ V(200),C(200)	P	8
	COMMON /BLOCK10/ AMAG(200),APHASE(200),WSTART,WEND,NRES	P	9
	COMMON /BLOCK14/ KP(5),W,FREQADD	P	10
	COMMON /BLOCK15/ NIT,JOUT,NGRAPH,NALLOUT	P	11
	COMMON /BLOCK16/ GRAF(200,5),JGRAPH(5),IGRAPH(5),NPRINT,SCALE,NITT	P	12
	COMMON /BLOCK21/ NST(5)	P	13
	N=NN-1	P	14
	DO 2 I=1,NGRAPH	P	15
	K=NST(I)	P	16
	FR(NPRINT)=W	P	17
	KR(NPRINT)=0	P	18
	IF (NRES.EQ.1) KR(NPRINT)=1	P	19
	IF (JGRAPH(I).EQ.0) GO TO 1	P	20
	GRAF(NPRINT,I)=C(K)/SCALE	P	21
	GO TO 2	P	22
1	GRAF(NPRINT,I)=V(K)	P	23
2	CONTINUE	P	24
	IF (NPRINT.LT.200.AND.W.LT.WEND) RETURN	P	25
	DO 27 I=1,NGRAPH	P	26
	PRINT 28	P	27
	A=7HVOLTAGE	P	28
	IF (JGRAPH(I).EQ.1) A=7HCURRENT	P	29
	IF (ITTEST(I).EQ.1) GO TO 3	P	30
	PRINT 29, IGRAPH(I),A	P	31
	GO TO 19	P	32
3	PRINT 30, IGRAPH(I),A	P	33
	PI=180.0/3.14159265	P	34
	DO 7 J=1,NPRINT	P	35
	IF (REAL(GRAF(J,I)).EQ.0.0) GO TO 6	P	36
	A=ATAN(AIMAG(GRAF(J,I))/REAL(GRAF(J,I)))*PI	P	37
	B=REAL(GRAF(J,I))	P	38
	IF (A.LT.0.0) GO TO 4	P	39
	IF (A.GE.0.0) GO TO 5	P	40
4	IF (B.LT.0.0) A=A+180.	P	41
	GRAF(J,I)=CMPLX(A,0.0)	P	42
	GO TO 7	P	43
5	IF (B.LT.0.0) A=A-180.	P	44
	GRAF(J,I)=CMPLX(A,0.0)	P	45
	GO TO 7	P	46
6	A=90.0	P	47
	IF (AIMAG(GRAF(J,I)).EQ.0.0) A=0.0	P	48
	IF (AIMAG(GRAF(J,I)).LT.0.0) A=-A	P	49
	GRAF(J,I)=CMPLX(A,0.0)	P	50
7	CONTINUE	P	51
	AMAX=ABS(REAL(GRAF(1,I)))	P	52
	DO 8 J=2,NPRINT	P	53
	A=ABS(REAL(GRAF(J,I)))	P	54
	IF (A.GT.AMAX) AMAX=A	P	55

8	CONTINUE	P	56
	IF (AMAX.EQ.0.0) AMAX=90.0	P	57
	DO 9 J=1,NPRINT	P	58
	NP(J)=REAL(GRAF(J,I))/AMAX*50.0+51.0	P	59
9	CONTINUE	P	60
	DO 10 J=1,101	P	61
	ANP(J)=1H.	P	62
10	CONTINUE	P	63
	PRINT 31, (ANP(J),J=1,101)	P	64
	DO 11 J=1,102	P	65
	ANP(J)=1H	P	66
11	CONTINUE	P	67
	DO 14 KK=1,NPRINT	P	68
	ANP(51)=1H.	P	69
	LL=NP(KK)	P	70
	ANP(LL)=1H+	P	71
	A=REAL(GRAF(KK,I))	P	72
	W=FR(KK)	P	73
	IF (KR(KK).EQ.1) GO TO 15	P	74
12	PRINT 32, W,A,(ANP(J),J=1,102)	P	75
	ANP(LL)=1H	P	76
	IF (KR(KK).EQ.1) GO TO 17	P	77
13	CONTINUE	P	78
14	CONTINUE	P	79
	GO TO 27	P	80
15	DO 16 II=5,15	P	81
	ANP(II)=1HX	P	82
16	CONTINUE	P	83
	ANP(LL)=1H+	P	84
	IF (ITTEST(I).EQ.0) GO TO 24	P	85
	GO TO 12	P	86
17	DO 18 II=5,15	P	87
	ANP(II)=1H	P	88
18	CONTINUE	P	89
	IF (ITTEST(I).EQ.0) GO TO 25	P	90
	GO TO 13	P	91
19	AMAX=GRAF(1,I)*CONJG(GRAF(1,I))	P	92
	DO 20 J=2,NPRINT	P	93
	A=GRAF(J,I)*CONJG(GRAF(J,I))	P	94
	IF (A.GT.AMAX) AMAX=A	P	95
20	CONTINUE	P	96
	IF (AMAX.EQ.0.0) AMAX=1.0	P	97
	B=SQRT(AMAX)	P	98
	DO 21 J=1,NPRINT	P	99
	A=SQRT(GRAF(J,I)*CONJG(GRAF(J,I)))	P	100
	GRAF(J,I)=CMPLX(A,0.0)	P	101
	NP(J)=A*100.0/B+1	P	102
21	CONTINUE	P	103
	DO 22 J=1,101	P	104
	ANP(J)=1H.	P	105
22	CONTINUE	P	106
	PRINT 31, (ANP(J),J=1,101)	P	107
	DO 23 J=1,102	P	108

	ANP(J)=1H	P 109
23	CONTINUE	P 110
	DO 26 KK=1,NPRINT	P 111
	ANP(1)=1H.	P 112
	LL=NP(KK)	P 113
	ANP(LL)=1H+	P 114
	A=REAL(GRAF(KK,I))	P 115
	W=FR(KK)	P 116
	IF (KR(KK).EQ.1) GO TO 15	P 117
24	PRINT 32, W,A,(ANP(J),J=1,102)	P 118
	ANP(LL)=1H	P 119
	IF (KR(KK).EQ.1) GO TO 17	P 120
25	CONTINUE	P 121
26	CONTINUE	P 122
27	CONTINUE	P 123
	NGRAPH=201	P 124
	RETURN	P 125
C		P 126
C		P 127
28	FORMAT (1H1)	P 128
29	FORMAT (1H0,30X,*BRANCH*,I3,5X, A7 ,6X,*MAGNITUDE PLOT *)	P 129
30	FORMAT (1H0,30X,*BRANCH*,I3,5X, A7 ,6X,*PHASE PLOT *)	P 130
31	FORMAT (1H0,1X,*FREQUENCY*,6X,*VALUE*,6X,101A1)	P 131
32	FORMAT (1HZ,E11.4,2X,E11.4,3X,102A1)	P 132
	END	P 133-

APPENDIX EGeneral Comments Regarding Use of CANDOFD.

1. Number of nodes cannot exceed 50
2. Number of branches cannot exceed 200
3. Cannot exceed 5 graphical outputs
4. There is an internal limit of 200 output points per graph.
5. Branch and node numbering is arbitrary, but branch numbers cannot exceed three digits, and node numbers cannot exceed two digits.
6. Zero is not a valid node or branch number.
7. There must be at least one branch, which is not an independent source, in any given network.
8. CANDOFD can handle zero valued elements as long as there are no loops of short circuits and no cutsets of open circuits. In this context, zero valued R's and L's, and independent voltage sources are short circuits, and zero valued C's and independent current sources are open circuits.
9. If a DC network analysis is to be performed, there can be no independent current source - capacitance cutsets, and no independent voltage source - inductance loops.
10. We cannot have both of the variables FACTOR and FREQADD simultaneously equal to zero.
11. Core requirements:
 Since our Fletcher-Powell minimization scheme requires the storing of an $N \times N$ matrix, where

$$N = 2 * (NB - \# \text{ of independent sources})$$
 it would be inconvenient to always dimension the above matrix 400×400 .

Thus, to save core, one need only replace the DIMENSION statement in the MAIN program by

```
DIMENSION HH(N, N)
```

where N is the largest number to be encountered in the network (s) to be solved.

With the above dimensioning procedure, the core requirements for CANDOFD are, approximately,

$$29500 + N^2 \text{ (decimal)}$$

storage locations.

12. In general, one should make use of the automatic current scaling (NSCALE \neq 1) and of the optimal tree picking (NTREE = 1) algorithms at each frequency point.

13. In general, unless the network is lossless, one need not test for ideal resonances (IDRES = 1) and one need not output the tree and \underline{F} information (NCONT \neq 1)

14. Comments 12 and 13 will, in general, reduce the execution time required for the analysis of a network. Sample problem #2 indicates variations in CP time caused by different selections of the control variables discussed above.

15. The output voltage and current units are always volts and amperes respectively.

16. Whenever FREQADD \neq 0.0, CANDOFD assumes that the user wishes to increment the frequency additively.

17. An attempt to graph a signal which is identically zero over the entire frequency range of interest, will result in an arithmetic error during the execution of CANDOFD.

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- (3) R. A. Rohrer, private communication.