

Copyright © 1971, by the author(s).
All rights reserved.

Permission to make digital or hard copies of all or part of this work for personal or classroom use is granted without fee provided that copies are not made or distributed for profit or commercial advantage and that copies bear this notice and the full citation on the first page. To copy otherwise, to republish, to post on servers or to redistribute to lists, requires prior specific permission.

CONVOLUTION FEEDBACK SYSTEMS

by

C. A. Desoer and F. M. Callier

Memorandum No. ERL-M312

19 July 1971

ELECTRONICS RESEARCH LABORATORY

**College of Engineering
University of California, Berkeley
94720**

Convolution Feedback Systems

by

C. A. DESOER and F. M. CALLIER

Department of Electrical Engineering and Computer Sciences
and the Electronics Research Laboratory,
University of California, Berkeley, California 94720

Abstract

This paper considers multi-input multi-output feedback systems characterized by $y = G * e$ and $e = u - y$. Theorem I shows that if the closed loop impulse response H is stable in the sense that $H \in \mathcal{A}^{n \times n}(\sigma)$, then $\hat{G}(s) = \hat{P}(s)[\hat{Q}(s)]^{-1}$ where $\hat{P}(s)$, $\hat{Q}(s)$ are also in $\mathcal{A}^{n \times n}(\sigma)$. Theorem II gives necessary and sufficient conditions for $H \in \mathcal{A}^{n \times n}(\sigma)$. Finally Theorem III gives necessary and sufficient condition for stability when $\hat{G}(s)$ has a finite number of multiple poles in $\text{Re } s \geq \sigma$: the case where the leading term of the Laurent expansion at each of these poles is singular is treated in detail.

I. Introduction

This paper considers linear time-invariant feedback systems with n inputs and n outputs. As it will become apparent, there is no loss of generality in taking the feedback to be unity. The input u , output y and error e are functions from \mathbb{R}_+ , (defined as $[0, \infty)$), to \mathbb{R}^n or corresponding distributions on \mathbb{R}_+ . The open loop system is of the convolution type so that we have

$$(1) \quad y = G * e$$

$$(2) \quad e = u - y$$

G is an $n \times n$ matrix whose elements are distributions on \mathbb{R}_+ . We use \mathcal{G} to denote the map $\mathcal{G}: e \mapsto G * e$.

We shall repeatedly use the convolution algebra $\mathcal{A}(\sigma)$ [1,2]: f is said to be in $\mathcal{A}(\sigma)$ iff $f(t) = 0$ for $t < 0$ and

$$(3) \quad f(t) = f_a(t) + \sum_0^{\infty} f_i \delta(t - t_i)$$

where $f_a(t) e^{-\sigma t} \in L^1(0, \infty)$, $f_i \in \mathbb{R}$ for all i , $\sum_0^{\infty} |f_i| e^{-\sigma t_i} < \infty$ and

$0 = t_0 < t_1 < t_2 \dots$. Thus f is a distribution of order 0 with support on \mathbb{R}_+ . An n -vector v ($n \times n$ matrix A) is said to be in $\mathcal{A}^n(\sigma)$ ($\mathcal{A}^{n \times n}(\sigma)$)

iff all its elements are in $\mathcal{A}(\sigma)$. Let \hat{f} denote the Laplace transform of f : f belongs to the convolution algebra $\mathcal{A}(\sigma)$ if and only if \hat{f} belongs to the algebra $\hat{\mathcal{A}}(\sigma)$ (with pointwise product). Similarly, $\hat{v} \in \hat{\mathcal{A}}^n(\sigma)$, $\hat{A} \in \hat{\mathcal{A}}^{n \times n}(\sigma)$.

Recently M. Vidyasagar [5] has shown that the class of systems (1), (2) where

$$(4) \quad \hat{G}(s) = \hat{P}(s)[\hat{Q}(s)]^{-1}$$

with $\hat{P}, \hat{Q} \in \hat{\mathcal{A}}^{n \times n}(\sigma)$ is very useful for distributed networks for example, and he extended some stability results of Desoer, Wu, Baker, Vakharia and Lam [1,2,3,10]. In Theorem I below we prove that, under very mild assumptions on G and on the closed loop system, if the closed loop impulse response $H \in \mathcal{A}^{n \times n}(\sigma)$ then \hat{G} is of the form (4). Theorem I is also an extension of a result of Nasburg and Baker [4]: the extension is in two directions, first, the n -input n -output case is considered and, second, the requirements on G are greatly relaxed. Theorem II is a straightforward extension of a result of [4]: it shows the importance of the systems considered by Vidyasagar in the sense that $H \in \mathcal{A}^{n \times n}(\sigma)$ if and only if \hat{G} is of the form (4). Finally Theorem III gives the necessary and sufficient conditions for stability of the closed loop system when \hat{G} is of the form (4) with a finite number of poles of finite order in $\text{Re } s \geq \sigma$. This theorem culminates a series of investigations starting with [1,2,3,7]. Note that except for [7], all previous work could only prove sufficiency.

II. The relation between G and H .

Theorem I

Let G be an $n \times n$ matrix whose elements are distributions with support on \mathbb{R}_+ . Suppose that in a neighborhood of the origin, say $V \subset \mathbb{R}$, G includes

at most δ -functions (i.e. on V , it is a distribution of at most order 0).

For the system defined by (1) and (2), assume that the closed loop response H exists and is uniquely defined by

$$(5) \quad H + G*H = G.$$

Under these conditions, if $H \in \mathcal{A}^{n \times n}(\sigma)$, then

(a) G is Laplace transformable and for some $\bar{\sigma} \geq \sigma$, $G \in \mathcal{A}^{n \times n}(\bar{\sigma})$.

(b) \hat{G} is of the form

$$(6) \quad \hat{G}(s) = \hat{P}(s)[\hat{Q}(s)]^{-1} \text{ for } \text{Re } s > \sigma$$

where $\hat{P}(\cdot)$ and $\hat{Q}(\cdot) \in \mathcal{A}^{n \times n}(\sigma)$.

(c) \hat{G} can at most have a countable number of poles in $\text{Re } s > \sigma$.

Comment. This theorem shows that under mild conditions on G regarding its behavior near $t = 0$; once the closed loop system is well-defined and "stable", then \hat{G} is necessarily of the form (6), can at most have poles in the strip $\sigma < \text{Re } s \leq \bar{\sigma}$ and is analytic for $\text{Re } s \geq \bar{\sigma}$.

Proof.

(a) By assumption, H is of the form

$$H(t) = H_a(t) + \sum_{i=0}^{\infty} H_i \delta(t-t_i)$$

where $0 = t_0 < t_1 < t_2 < \dots$. By assumption G can at most have an impulse at the origin. By the Abelian theorem of the Laplace transform [11] and

the properties of distributions, if G has an impulse G_0 at $t = 0$, $\hat{G}(s) \rightarrow G_0$ as $s \rightarrow \infty$ with $\text{Re } s \rightarrow \infty$. Clearly from (5), if G_0 is the zero matrix, then $H_0 = 0$. If $G_0 \neq 0$, then by balancing impulses at the origin in (5) we have $(I + G_0)H_0 = G_0$. By assumption H , hence H_0 , is uniquely defined by (5), hence $\det(I + G_0) \neq 0$. Furthermore by direct calculation, $(I - H_0)(I + G_0) = I$, so that $\det[I - H_0] \neq 0$.

The function $I - \hat{H}(s)$ is analytic and bounded for $\text{Re } s > \sigma$, and tends to $I - H_0$ as $s \rightarrow \infty$ with $\text{Re } s \rightarrow \infty$. Consequently, there exists a $\bar{\sigma} \geq \sigma$ such that

$$(7) \quad \inf_{\text{Re } s > \bar{\sigma}} |\det[I - \hat{H}(s)]| > 0 .$$

From (5), if G had a Laplace transform, we would have $\hat{H} + \hat{G}\hat{H} = \hat{G}$. Now by (7), $\hat{H}(s)[I - \hat{H}(s)]^{-1} \in \mathcal{A}^{n \times n}(\bar{\sigma})$, so $\hat{G}(s)$ is equal to that function, by the uniqueness of the convolution algebra of distributions on \mathbb{R}_+ .

(b) Since $\hat{H}(s)$ is analytic for $\text{Re } s > \sigma$, $[I - \hat{H}(s)]^{-1}$ has at most a countable number of poles in $\text{Re } s > \sigma$ and by analytic continuation

$$(8) \quad \hat{G}(s) = \hat{H}(s)[I - \hat{H}(s)]^{-1} \text{ for } \text{Re } s > \sigma .$$

Choose $\hat{P}(s) = \hat{H}(s)$, $\hat{Q}(s) = I - \hat{H}(s)$. Thus (b) and (c) have been established. □

Remark. It is important to reflect on the fact that under the conditions of Theorem I, we have

$$[I + \hat{G}(s)][I - \hat{H}(s)] = I \text{ for } \text{Re } s > \sigma .$$

This expression emphasizes the symmetrical role played by \mathbb{H} and \mathbb{G} : \mathbb{H} is obtained from \mathbb{G} by a negative feedback of \mathbb{I} ; \mathbb{G} is obtained from \mathbb{H} by a negative feedback of $-\mathbb{I}$ (to cancel the preceding one!).

Theorem II

Let G be an $n \times n$ matrix whose elements are Laplace transformable distributions with support in \mathbb{R}_+ . For the system defined by (1) and (2), assume that the closed loop transfer function \hat{H} is well-defined for almost all s in the half plane of convergence of \hat{G} ; i.e.

$$(9) \quad \hat{H}(s) = \hat{G}(s)[\mathbb{I} + \hat{G}(s)]^{-1}$$

for almost all s in the half-plane of convergence of $\hat{G}(\cdot)$. Under these conditions,

$$(10) \quad H \in \mathcal{A}^{n \times n}(\sigma)$$

if and only if there exists $\hat{P}, \hat{Q} \in \mathcal{A}^{n \times n}(\sigma)$ such that

$$(11) \quad \hat{G}(s) = \hat{P}(s)[\hat{Q}(s)]^{-1}$$

and

$$(12) \quad \inf_{\operatorname{Re} s > \sigma} |\det[\hat{P}(s) + \hat{Q}(s)]| > 0.$$

Proof:

Necessity. From (9) by algebra

$$\hat{G}(s) = \hat{H}(s)[\mathbb{I} - \hat{H}(s)]^{-1} \text{ for } \operatorname{Re} s \geq \sigma$$

Choose $\hat{P}(s) = \hat{H}(s) \in \hat{\mathcal{A}}^{n \times n}(\sigma)$ and $\hat{Q}(s) = I - \hat{H}(s) \in \hat{\mathcal{A}}^{n \times n}(\sigma)$, by (10).
 Since $\hat{P} + \hat{Q} = I$, (12) holds.

Sufficiency. From (9) and (11)

$$\hat{H}(s) = \hat{P}(s)[\hat{P}(s) + \hat{Q}(s)]^{-1}.$$

In view of (12) $\hat{H} \in \hat{\mathcal{A}}^{n \times n}(\sigma)$ as the product of two elements of $\hat{\mathcal{A}}^{n \times n}(\sigma)$.
 □

Remark. It is clear from (11) that a given \hat{G} does not define the ordered pair (\hat{P}, \hat{Q}) uniquely; for example, they might have a matrix as right common factor. In order to be able to express the condition (12) in a form which depends on \hat{G} only, we impose the Vidyasagar no-cancellation condition (N)

[5]: the ordered pair (a, b) where $a, b: \mathbb{C} \rightarrow \mathbb{C}$ is said to satisfy the no-cancellation condition on a set $A \subset \mathbb{C}$ iff, for all sequences $\{s_k\}$ in A , $a(s_k) \rightarrow 0$ implies that $\liminf |b(s_k)| > 0$.

It is then easy to show that, [5], if $(\det \hat{Q}(s), \det[\hat{P}(s) + \hat{Q}(s)])$ satisfies (N) on $\text{Re } s \geq \sigma$, then (12) is equivalent to $\inf_{\text{Re } s \geq \sigma} |\det[I + \hat{G}(s)]| > 0$.

III. Necessary and Sufficient Conditions for Stability.

We consider first and in detail the case where \hat{G} has a single pole p of order m in $\text{Re } s \geq \sigma$. The extension to the case of a finite number of poles is straightforward.

We consider the open loop transfer function

$$(13) \quad \hat{G}(s) = \sum_{i=0}^{m-1} R_i (s-p)^{-m+i} + \hat{G}_0(s)$$

where $\operatorname{Re} p \geq \sigma$, $\hat{G}_0 \in \mathcal{A}^{n \times n}(\sigma)$, $r_0 \triangleq \operatorname{rank} \hat{R}_0 \leq n$ and R_i ($i = 0, 1, \dots, m-1$) are $n \times n$ matrices with complex coefficients. We start by pointing out some facts which will streamline the proof.

Fact 1. Let

$$(14) \quad \hat{R}\left(\frac{1}{s+a}\right) \triangleq \left(\sum_{i=0}^{m-1} R_i (s-p)^{-m+i} \right) \left(\frac{s-p}{s+a} \right)^m; \quad a \triangleq 1-\sigma$$

then $\hat{R}\left(\frac{1}{s+a}\right)$ is $n \times n$ complex polynomial matrix in $\left(\frac{1}{s+a}\right)$ of degree m . This is obvious by considering the Laurent expansion of $\hat{R}\left(\frac{1}{s+a}\right)$ about $s = -a$.

Fact 2. (Smith canonical form [12]). For the $n \times n$ polynomial matrix $\hat{R}\left(\frac{1}{s+a}\right)$ there exist unimodular (i.e. with nonzero constant determinant) polynomial matrices in $\left(\frac{1}{s+a}\right)$ viz. $\hat{P}\left(\frac{1}{s+a}\right)$ and $\hat{Q}\left(\frac{1}{s+a}\right)$, such that:

$$(15) \quad \hat{Q}\left(\frac{1}{s+a}\right) \hat{R}\left(\frac{1}{s+a}\right) \hat{P}\left(\frac{1}{s+a}\right) =$$

$$\operatorname{diag} \left\{ \underbrace{\hat{a}_1\left(\frac{1}{s+a}\right), \dots, \hat{a}_j\left(\frac{1}{s+a}\right), \dots, \hat{a}_r\left(\frac{1}{s+a}\right)}_r, \underbrace{0, 0, \dots, 0}_{n-r} \right\}$$

where i) $r = \operatorname{rank}$ of $\hat{R}\left(\frac{1}{s+a}\right) =$ order of the largest minor of $\hat{R}\left(\frac{1}{s+a}\right)$ whose determinant is not equal to the zero polynomial;

ii) $\hat{a}_j\left(\frac{1}{s+a}\right)$ $j = 1, 2, \dots, r$ are the invariant polynomials of $\hat{R}\left(\frac{1}{s+a}\right)$ and each polynomial $\hat{a}_j(\cdot)$ divides $\hat{a}_{j+1}(\cdot)$, $j = 1, 2, \dots, r-1$;

iii) the diagonal matrix in the R.H.S. of (15) can be obtained by elementary operations.

Fact 3. The polynomial matrices $\hat{P}\left(\frac{1}{s+a}\right)$ and $\hat{Q}\left(\frac{1}{s+a}\right) \in \hat{A}^{n \times n}(\sigma)$ and their inverses are polynomial matrices in $\left(\frac{1}{s+a}\right)$ also in $\hat{A}^{n \times n}(\sigma)$.

Fact 4.

Let $\hat{a}_j(\cdot)$ $j = 1, 2, \dots, r$ be as in (15) and let r_0 be the rank of R_0 , then

(a)

$$(16) \quad \begin{cases} \hat{a}_j(1/(p+a)) = 0 \text{ for } r_0 + 1 \leq j \leq r \text{ by definition of } r_0; \\ \hat{a}_j(1/(p+a)) \neq 0 \text{ for } 1 \leq j \leq r_0; \end{cases}$$

(b)

$$(17) \quad \hat{a}_j\left(\frac{1}{s+a}\right) = \hat{b}_j\left(\frac{1}{s+a}\right)\left(\frac{s-p}{s+a}\right)^{c_j} \text{ for } r_0 + 1 \leq j \leq r$$

where c_j is the order of the zero of $\hat{a}_j(\cdot)$ at $s = p$;

$\hat{b}_j(\cdot)$ is a polynomial with

$$(18) \quad \hat{b}_j(1/(p+a)) \neq 0, \text{ (see [13]), and}$$

$$1 \leq c_{r_0+1} \leq c_{r_0+2} \leq \dots \leq c_r.$$

Proof. Set $s=p$ in (15) and note that the L.H.S. becomes $\hat{Q}(1/(p+a)) R_0(p+a)^{-m} \hat{P}(1/(p+a))$. Since $\hat{P}(\cdot)$ and $\hat{Q}(\cdot)$ are unimodular, exactly $(r-r_0)$ polynomials $\hat{a}_j(\cdot)$ are zero at $s=p$. By ii) of (15) $\hat{a}_j(1/(p+a)) = 0$ for $r_0 + 1 \leq j \leq r$. Hence (16) and (17) follow with the properties of the latter as a consequence of ii) of (15). □

Remark. Note that the exponents c_j in (17) may, for some j , be larger

than m (in fact $c_r \leq rm$).

Therefore, since the c_j are monotonically increasing and since $c_j - m$ may be of any sign, partition the index set $K = \{r_0+1, r_0+2, \dots, r\}$ into

$$(19) \quad K_- = \{r_0+1, r_0+2, \dots, \alpha\} = \{j \mid 1 \leq c_j < m\}$$

$$(20) \quad K_0 = \{\alpha+1, \alpha+2, \dots, \beta\} = \{j \mid c_j = m\}$$

$$(21) \quad K_+ = \{\beta+1, \beta+2, \dots, r\} = \{j \mid c_j > m\}$$

We are now ready for Theorem III.

Theorem III.

Let $\hat{G}(s)$ be given by (13) and let $\hat{P}(\frac{1}{s+a})$ and $\hat{Q}(\frac{1}{s+a})$ be the polynomial matrices defined in (15). Suppose that the index-sets K_-, K_0, K_+ , as defined in (19)-(21), are not empty.

Consider the partitioning

$$(22) \quad \hat{Q}\left(\frac{1}{s+a}\right) [I + \hat{G}_0(s)] \hat{P}\left(\frac{1}{s+a}\right) = \begin{matrix} & \begin{matrix} \alpha & n-\alpha \end{matrix} \\ \begin{matrix} \alpha \\ n-\alpha \end{matrix} & \left[\begin{array}{c|c} \hat{L}_{11}(s) & \hat{L}_{12}(s) \\ \hline \hat{L}_{21}(s) & \hat{L}_{22}(s) \end{array} \right] \end{matrix}$$

and let $\hat{\delta}_j(\cdot)$ be the polynomials defined in (17). Under these conditions,

$$(10) \quad H \in \mathcal{A}^{n \times n}(\sigma)$$

if and only if

$$(23) \quad \inf_{\text{Res} > \sigma} |\det[I + \hat{G}(s)]| > 0$$

and

$$(C) \quad \det\{\hat{L}_{22}(p) + \text{diag}[\delta_{\alpha+1}(1/(p+a)), \dots, \delta_{\beta}(1/(p+a)), 0, 0, \dots, 0]\} \neq 0.$$

Proof.

Sufficiency. Since $I - \hat{H}(s) = [I + \hat{G}(s)]^{-1}$, we need only to show that

$$(24) \quad [I + \hat{G}(s)]^{-1} \in \hat{\mathcal{A}}^{n \times n}(\sigma).$$

By fact 3, (24) is equivalent to

$$\left\{ \hat{Q}\left(\frac{1}{s+a}\right) [I + \hat{G}(s)] P\left(\frac{1}{s+a}\right) \right\}^{-1} \in \hat{\mathcal{A}}^{n \times n}(\sigma).$$

Introduce now the following multiplier:

$$(25) \quad \hat{M}(s) \triangleq \underbrace{\text{diag}\{\hat{z}(s)^m, \hat{z}(s)^m, \dots, \hat{z}(s)^m\}}_{r_0} \underbrace{\hat{z}(s)^{m-c_{r_0+1}}, \hat{z}(s)^{m-c_{r_0+2}}, \dots, \hat{z}(s)^{m-c_{\alpha}}}_{\alpha - r_0} \underbrace{\{1, \dots, 1\}}_{n - \alpha}$$

with

$$(26) \quad \hat{z}(s) = \frac{s-p}{s+a} \in \hat{\mathcal{A}}(\sigma).$$

By (19) and (26)

$$(27) \quad \hat{M}(s) \in \hat{\mathcal{A}}^{n \times n}(\sigma).$$

Remark that

$$\left\{ \hat{Q}\left(\frac{1}{s+a}\right) [I+\hat{G}(s)] \hat{P}\left(\frac{1}{s+a}\right) \right\}^{-1} = \hat{M}(s) \hat{N}(s)^{-1} \quad \text{where}$$

$$(28) \quad \hat{N}(s) \triangleq \left\{ \hat{Q}\left(\frac{1}{s+a}\right) [I+\hat{G}(s)] \hat{P}\left(\frac{1}{s+a}\right) \right\} \hat{M}(s) .$$

Clearly by (27) we are done if we can show that

$$\hat{N}(s)^{-1} \in \hat{\mathcal{A}}^{n \times n}(\sigma) .$$

Therefore by a reasoning of [2], we prove that $\hat{N}(s) \in \hat{\mathcal{A}}^{n \times n}(\sigma)$ and

$$\inf_{\text{Res} > \sigma} |\det \hat{N}(s)| > 0 .$$

Rewrite (25), therefore

$$(29) \quad \hat{M}(s) = \hat{z}(s)^m \hat{\Delta}(s) \quad \text{where}$$

$$(30) \quad \hat{\Delta}(s) \triangleq \text{diag} \left\{ \underbrace{1, 1, \dots, 1}_{r_0}, \underbrace{\hat{z}(s)^{-c_{r_0+1}}, \hat{z}(s)^{-c_{r_0+2}}, \dots, \hat{z}(s)^{-c_\alpha}}_{\alpha - r_0}, \underbrace{\hat{z}(s)^{-m}, \hat{z}(s)^{-m}, \dots, \hat{z}(s)^{-m}}_{n - \alpha} \right\} .$$

By (28), (13), (29), (30), (26), (14), (15), (17) and (20), we obtain

$$(31) \quad \hat{N}(s) = \hat{N}_1(s) + \hat{N}_2(s) \quad \text{where}$$

(a)

$$(32) \quad \hat{N}_1(s) = \hat{D}_1(s) \oplus \hat{D}_2(s) \quad \text{with}$$

$$(33) \quad \hat{D}_1(s) = \text{diag} \left\{ \underbrace{\hat{a}_1\left(\frac{1}{s+a}\right), \hat{a}_2\left(\frac{1}{s+a}\right), \dots, \hat{a}_{r_0}\left(\frac{1}{s+a}\right)}_{r_0}, \underbrace{\hat{b}_{r_0+1}\left(\frac{1}{s+a}\right), \hat{b}_{r_0+2}\left(\frac{1}{s+a}\right), \dots, \hat{b}_\alpha\left(\frac{1}{s+a}\right)}_{\alpha-r_0} \right\}$$

$$(34) \quad \hat{D}_2(s) = \text{Diag} \left\{ \underbrace{\hat{b}_{\alpha+1}\left(\frac{1}{s+a}\right), \hat{b}_{\alpha+2}\left(\frac{1}{s+a}\right), \dots, \hat{b}_\beta\left(\frac{1}{s+a}\right)}_{\beta-\alpha}, \underbrace{\hat{b}_{\beta+1}\left(\frac{1}{s+a}\right) \hat{z}(s)^{c_{\beta+1}-m}}_{r-\beta}, \dots, \underbrace{\hat{b}_r\left(\frac{1}{s+a}\right) \hat{z}(s)^{c_r-m}, 0, 0, \dots, 0}_{n-r} \right\}$$

and (b)

$$(35) \quad \hat{N}_2(s) = \hat{Q}\left(\frac{1}{s+a}\right) [I + \hat{G}_0(s)] \hat{P}\left(\frac{1}{s+a}\right) \hat{M}(s) .$$

Immediately

$$(36) \quad \hat{N}(s) \in \hat{A}^{n \times n}(\sigma) .$$

$\hat{N}_1(s) \in \hat{A}^{n \times n}(\sigma)$ because all its elements $\in \hat{A}(\sigma)$ (indeed all its nonzero elements are polynomials in $\left(\frac{1}{s+a}\right)$ because there are no negative powers of $\hat{z}(s)$ by (21)) and $\hat{N}_2(s) \in \hat{A}^{n \times n}(\sigma)$ by fact 3, (13) and (27). Finally by (23) and since $\hat{P}\left(\frac{1}{s+a}\right)$ and $\hat{Q}\left(\frac{1}{s+a}\right)$ are unimodular

$$\inf_{\text{Res} > \underline{\sigma}} |\det \hat{Q}\left(\frac{1}{s+a}\right) [I + \hat{G}(s)] \hat{P}\left(\frac{1}{s+a}\right)| > 0 .$$

Hence, since by (25)-(26) $\det \hat{M}(s)$ has only one zero for $\operatorname{Re} s \geq \sigma$ i.e. at p , we obtain with (28)

$$(37) \quad \inf_{s \in U} |\det \hat{N}(s)| > 0$$

where U is the half plane $\operatorname{Re} s \geq \sigma$ with a small neighborhood of p deleted. Consider now $\det \hat{N}(p)$.

Remark that by (35), (22) and (25)-(26)

$$(38) \quad \hat{N}_2(s) = \alpha \left\{ \begin{array}{c|c} \overbrace{\hat{K}_{11}(s)}^{\alpha} & \hat{L}_{12}(s) \\ \hline \hat{K}_{21}(s) & \hat{L}_{22}(s) \end{array} \right\}$$

with

$$(39) \quad \hat{K}_{11}(p) = 0$$

$$(40) \quad \hat{K}_{21}(p) = 0 .$$

Thus by (31), (32), (38)-(40)

$$\det \hat{N}(p) = \det \hat{D}_1(p) \det [\hat{L}_{22}(p) + \hat{D}_2(p)] \quad \text{with}$$

by (33), (16) and (18)

$$(41) \quad \det \hat{D}_1(p) \neq 0$$

and by (34), (18), (26) and (21)

$$(42) \quad \det[\hat{L}_{22}(p) + \hat{D}_2(p)] = \\ \det\{\hat{L}_{22}(p) + \text{diag}[\hat{\delta}_{\alpha+1}(1/(p+a)), \dots, \hat{\delta}_{\beta}(1/(p+a)), 0, \dots, 0]\}$$

which is nonzero by (C). Hence

$$(43) \quad \det \hat{N}(p) \neq 0 .$$

Since $\hat{N}(s)$ is continuous in $\text{Re } s \geq \sigma$, (36), (37) and (43) imply that

$$[\hat{N}(s)]^{-1} \in \hat{\mathcal{A}}^{n \times n}(\sigma) . \quad \text{Q.E.D.} \quad \square$$

Necessity. $\hat{H} \in \hat{\mathcal{A}}^{n \times n}(\sigma)$ by assumption.

(23) follows immediately by [6].

To establish (C) we use contradiction. So by (42) suppose that

$\det[\hat{L}_{22}(p) + \hat{D}_2(p)] = 0$. We are going to show that, for some input $u \in L_n^{2\sigma}[0, \infty)$ (i.e. $u(t) e^{-\sigma t} \in L_n^2[0, \sigma)$), the system defined by (1)-(2) has an error e and thus also an output $y = u - e$ not in $L_n^{2\sigma}[0, \infty)$. This is a contradiction because $u \in L_n^{2\sigma}[0, \infty)$ and $H \in \hat{\mathcal{A}}^{n \times n}(\sigma)$ imply $y = H * u \in L_n^{2\sigma}[0, \infty)$, [1] [2].

The Laplace transforms of e and u are related by

$$(44) \quad [I + \hat{G}(s)]\hat{e}(s) = \hat{u}(s) .$$

Multiply (44) on the left by $\hat{Q}(\frac{1}{s+a})$ and define the n -vectors $\bar{e}(s)$ and $\bar{u}(s)$ by

$$(45) \quad \hat{P}(\frac{1}{s+a}) \hat{M}(s) \bar{e}(s) = \hat{e}(s)$$

$$(46) \quad \hat{Q}(\frac{1}{s+a}) \hat{u}(s) = \bar{u}(s) .$$

By (44)-(46) and (28) obtain

$$(47) \quad \hat{N}(s) \bar{e}(s) = \bar{u}(s) .$$

Because $\det[\hat{L}_{22}(p) + \hat{D}_2(p)] = 0$ we can pick a nonzero vector $\eta \in \mathbb{C}^{n-\alpha}$ in the null space of $[\hat{L}_{22}(p) + \hat{D}_2(p)]$, hence

$$(48) \quad [\hat{L}_{22}(p) + \hat{D}_2(p)]\eta = 0 .$$

Pick now the vector $\xi \in \mathbb{C}^\alpha$ such that

$$(49) \quad \xi \triangleq - [\hat{D}_1(p)]^{-1} \hat{L}_{12}(p)\eta$$

which is well defined because of (41) and the fact that all elements of \hat{L}_{12} are in $\hat{A}(\sigma)$.

Hence with

$$(50) \quad \bar{e}(s) = \frac{1}{s-p} \begin{pmatrix} \xi \\ \eta \end{pmatrix} ,$$

and

$$(51) \quad \bar{u}(s) = \begin{pmatrix} \bar{u}_1(s) \\ \bar{u}_2(s) \end{pmatrix} \begin{matrix} \} \alpha \\ \} n-\alpha \end{matrix} ,$$

and (47), (31), (32), (38), we obtain

$$(52) \quad \bar{u}_1(s) = \{ [\hat{D}_1(s) + \hat{K}_{11}(s)]\xi + \hat{L}_{12}(s)\eta \} / (s-p)$$

$$(53) \quad \bar{u}_2(s) = \{ \hat{K}_{21}(s)\xi + [\hat{D}_2(s) + \hat{L}_{22}(s)]\eta \} / (s-p) .$$

All the components of the numerators of (52) and (53) are in $\hat{\mathcal{A}}(\sigma)$; by virtue of (39)-(40) and (48)-(49) they have at least a first order zero at p . Therefore $\bar{u}_1(s)$ and $\bar{u}_2(s)$ are well behaved and bounded at $s = p$. Thus $\bar{u}(s)$ is analytic for $\text{Re } s > \sigma$, bounded on $\text{Re } s \geq \sigma$ and as $|\omega| \rightarrow \infty$:

$$|\bar{u}(\text{Re } s + j\omega)| \text{ is at most } O\left(\frac{1}{\omega}\right) \text{ for any fixed } \text{Re } s \geq \sigma.$$

It follows therefore that the components of $\bar{u}(s)$ are the Laplace transforms of elements of $L_n^{2\sigma}[0, \infty)$ [14]. From fact 3 and (46) we conclude that the same is true for the components of $\hat{u}(s)$, hence

$$(54) \quad u \in L_n^{2\sigma}[0, \infty) .$$

Finally by (45), (50), (25)-(26) and since $\eta \neq 0$ and $\hat{P}\left(\frac{1}{s+a}\right)$ is unimodular, there exists at least one component of $\hat{e}(s)$ which has a nonzero residue at p .

Thus

$$(55) \quad e \notin L_n^{2\sigma}[0, \infty)$$

and by (54) and (55) we have established a contradiction. Q.E.D.

□

Remarks.

- 1) The theorem above describes in detail what happens when K_- , K_0 , K_+ are nonempty. When one or more of these sets are empty the required modifications of (C) and of the multiplier $\hat{M}(s)$ are straightforward.
- 2) In case there are l poles at p_1, p_2, \dots, p_l of order m_1, m_2, \dots, m_l with real part larger than or equal to σ , one uses a product of

multipliers like $\hat{M}(s)$, one for each pole. Condition (C) is used only to check that $\det \hat{N}(s)$ does not vanish at $s = p$. Therefore for the more general case an appropriate condition (C) is required at each pole.

3) We have checked that these techniques can be applied in a straightforward manner for the discrete-time case, thus providing a generalization to the work of Desoer, Wu and Lam [8,9,10].

References.

- [1] C. A. Desoer and M. Y. Wu, "Stability of linear time-invariant systems," IEEE Transactions on Circuit Theory, CT-15, pp. 245-250, Sept. 1968.
- [2] C. A. Desoer and M. Y. Wu, "Stability of multiple-loop feedback linear time-invariant systems," Jour. Math. Anal. and Appl., 23, pp. 121-130, June 1968.
- [3] R. A. Baker and D. J. Vakharia, "Input-output stability of linear time-invariant systems," IEEE Transactions on Automatic Control, AC-15, pp. 316-319, June 1970.
- [4] R. E. Nasburg and R. A. Baker, "Stability of linear time-invariant distributed parameter single-loop feedback systems," (to appear).
- [5] M. Vidyasagar, "Input-output stability of a broad class of linear time-invariant systems," (in press).
- [6] C. A. Desoer and M. Vidyasagar, "General necessary conditions for input-output stability," IEEE Proceedings (in press).
- [7] C. A. Desoer and F. L. Lam, "Recent results concerning the input-output properties of linear time-invariant systems," ERL-Memorandum #M-295, Electronics Research Laboratory, University of California, Berkeley, Jan. 1971, (to appear in IEEE Transactions on Circuit Theory).
- [8] C. A. Desoer and M. Y. Wu, "Input-output properties of linear discrete systems, Part I," Jour. Franklin Institute, 290, pp. 11-24, July 1970.

- [9] C. A. Desoer and M. Y. Wu, "Input-output properties of multiple-input, multiple-output nonlinear discrete systems, Part II," Jour. Franklin Institute, 290, 2, pp. 85-101, Aug. 1970.
- [10] C. A. Desoer and F. L. Lam, "Stability of linear time-invariant discrete systems," IEEE Proceedings, 58, pp. 1841-1843, Nov. 1970.
- [11] D. V. Widder, "The Laplace transform," Princeton U. Press, 1941, (Chap. IV, sec. 1).
- [12] F. R. Gantmacher, "Matrix Theory," vol. I, Chelsea, N. Y., 1959, (pp. 130-145, esp.).
- [13] K. Hoffman and R. Kunze, "Linear Algebra," Prentice-Hall, Englewood Cliffs, N. J., 1961, (pp. 108-128, esp.).
- [14] R.E.A.C. Paley and N. Wiener, "The Fourier transform in the complex domain," Amer. Math. Soc. Coll. Public., 19, New Providence, 1934 (Theorem V p. 8, esp.).