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ROBUST STABILITY UNDER ADDITIVE  
PERTURBATIONS: THE NONLINEAR CASE

by

A. Bhaya and C. A. Desoer

Memorandum No. UCB/ERL M85/4

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**Robust Stability under Additive  
Perturbations: The Nonlinear Case**

A. Bhaya and C. A. Desoer

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**Abstract**

We consider a MIMO nonlinear feedback system  $S(P, C)$  which is assumed to be  $S$ -stable. The plant  $P$  is subjected to an *arbitrary* additive (resp. multiplicative) perturbation  $\Delta P$  (resp.  $M$ ). We prove necessary and sufficient conditions for the  $S$ -stability of the perturbed system. As a special case, we obtain a generalization of our earlier result (on linear time-invariant systems) to linear time-varying systems.

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## I. Introduction

One of the main purposes of feedback is to reduce the sensitivity of the closed-loop system to changes in the plant, and it is very important to determine whether a feedback system remains stable after being subjected to changes in the plant. There is an abundant literature on this subject with various restrictions imposed on the nature of i) the plant: linear lumped [Des. 1], [Ast. 1], [Fra. 1] [Doy. 1]; linear distributed [Chen 1], [Chen 2]; nonlinear and time-varying [Zam. 1], [San. 1]), ii) the perturbation: a) stable perturbation: ([Ast. 1], [Fra. 1], [Cru. 1] [Pos. 1], [Zam. 2]) -all giving only sufficient conditions; b) a class of possibly *unstable linear* perturbations [Doy. 1], [Chen 1] with *necessary and sufficient conditions* (n.a.s.c.); c) *linear fractional* possibly unstable perturbations [Chen 2] with *n.a.s.c.*

In an earlier paper [Bha. 1], we considered MIMO *linear time-invariant* systems (lumped or distributed, discrete- or continuous-time) and a simple algebraic proof of a n.a.s.c. for stability of  ${}^1S(P, C)$ , (Fig. 1, solid lines only), when  $P$  is subjected to an additive perturbation  $\Delta P$  which is *proper* but *not necessary stable*. In this paper, we extend this result to MIMO *nonlinear* systems (Theorem 1, Sec. 3) and when we specialize this result to the linear case, (Sec. 4, Cor. 1), we obtain a generalization of the result of [Bha. 1] which applies to *time-varying* systems as well. We state and give a simple algebraic proof of a *necessary and sufficient condition* for  $\mathbf{S}$ -stability (see Sec. II for Defn. ) of the feedback system  ${}^1S(\tilde{P}, \tilde{C})$  under *arbitrary* additive (resp. multiplicative) perturbations  $\Delta \tilde{P}$ , (resp.  $\tilde{M}$ ), (i.e.,  $\Delta \tilde{P}$ , (resp.  $\tilde{M}$ ), is *not* required to be  $\mathbf{S}$ -stable).

In Section III we formalize the following intuitive argument: (the nominal system  ${}^1S(\tilde{P}, \tilde{C})$  is assumed  $\mathbf{S}$ -stable) a) the addition of  $\Delta \tilde{P}$  to  ${}^1S(\tilde{P}, \tilde{C})$  (as shown by the dotted lines in Fig. 1) creates a new loop; b) the "gain seen by  $\Delta \tilde{P}$ ,"

through  ${}^1S(\tilde{P}, \tilde{C})$  is equal to  $-Q_{\tilde{u}_2}$ , c) since the nominal system  ${}^1S(\tilde{P}, \tilde{C})$  is  $S$ -stable,  $Q_{\tilde{u}_2}$  is  $S$ -stable,  $\forall u_2 \in L_{\tilde{u}_2}^{n_2}$ , d) view the new loop as  ${}^1S(\Delta\tilde{P}, Q_{\tilde{u}_2})$  (Fig. 2): "clearly"  $\tilde{S}(\tilde{P}, \Delta\tilde{P}, \tilde{C})$  is  $S$ -stable  $\iff \forall u_2 \in L_{\tilde{u}_2}^{n_2}$ ,  ${}^1S(\Delta\tilde{P}, Q_{\tilde{u}_2})$  is  $S$ -stable.

## II. Definitions and Notations

U.t.c. means under these conditions. Let  $(L, \|\cdot\|)$  be a normed space of "time functions":  $T \rightarrow V$  where  $T$  is the time set (typically  $\mathbb{R}_+$  or  $\mathbb{N}$ ),  $V$  is a normed space typically  $\mathbb{R}$ ,  $\mathbb{R}^n$ ,  $\mathbb{C}^n$ ,  $\dots$ ) and  $\|\cdot\|$  is the chosen norm in  $L$ . Let  $L_{\tilde{u}_2}$  be the corresponding extended space (see e.g., [Des. 2]).

A function  $\varphi: \mathbb{R}_+ \rightarrow \mathbb{R}_+$  is said to belong to class  $K$  iff  $\varphi$  is  $\varphi$  continuous and increasing.  $\varphi$  is said to belong to class  $K_0$  iff  $\varphi \in K$  and  $\varphi(0) = 0$ . If  $\varphi_1$  and  $\varphi_2 \in K_0$ , then  $\varphi_1 + \varphi_2$  and  $\alpha \mapsto \varphi_1(\varphi_2(\alpha)) \in K_0$ . A nonlinear causal map  $H: L_{\tilde{u}_2}^{n_1} \rightarrow L_{\tilde{u}_2}^{n_2}$  is said to be  $S$ -stable iff  $\exists \varphi \in K$  s.t.  $\forall x \in L_{\tilde{u}_2}^{n_1}$ ,  $\forall T \in T$ ,

$$\|Hx\|_T \leq \varphi(\|x\|_T) .$$

$H$  is said to be incrementally  $S$ -stable (incr.  $S$ -stable) iff (i)  $H$  is  $S$ -stable,

(ii)  $\exists \tilde{\varphi} \in K_0$  s.t.  $\forall x, x' \in L_{\tilde{u}_2}^{n_1}$ ,  $\forall T \in T$ ,

$$\|Hx - Hx'\|_T \leq \tilde{\varphi}(\|x - x'\|_T)$$

Let  $\beta$ ,  $\gamma$ , and  $\tilde{\gamma}$  be constants: if  $\varphi(x) = \beta + \gamma x$ , then  $S$ -stability reduces to finite-gain stability [Saf. 1], [Des. 2]; if  $\tilde{\varphi}(x) = \tilde{\gamma}x$ , then incr.  $S$ -stability reduces to finite-gain incr. stability. It can be shown that if the nonlinear causal maps  $H_{\tilde{u}_2}^1$  and  $H_{\tilde{u}_2}^2$  are  $S$ -stable, (incr.  $S$ -stable), then  $H_{\tilde{u}_2}^1 + H_{\tilde{u}_2}^2$  and  $H_{\tilde{u}_2}^1 \circ H_{\tilde{u}_2}^2$  are  $S$ -stable, (incr.  $S$ -stable, resp.). (For simplicity, in what follows we drop the symbol " $\circ$ ")

denoting the composition of the maps.)

A feedback system is said to be *well-posed* iff the relation from the exogenous inputs into each subsystem<sup>†</sup> variable (i.e., subsystem input and subsystem output) is a well-defined nonlinear causal map between the corresponding extended spaces. More precisely, the system  ${}^1S(\underline{P}, \underline{C})$  of Fig. 1, where  $\underline{P}: L_{\tilde{v}}^{n_i} \rightarrow L_{\tilde{v}}^{n_o}$ ,  $\underline{C}: L_{\tilde{v}}^{n_o} \rightarrow L_{\tilde{v}}^{n_i}$  are causal maps, is said to be *well-posed* iff  $\underline{H}: (u_1, u_2) \mapsto (e_1, e_2, y_1, y_2)$  is well-defined and causal. Note that  ${}^1S(\underline{P}, \underline{C})$  is well-posed implies that  $(I + \underline{P}\underline{C})^{-1}$  and  $(I + \underline{C}\underline{P})^{-1}$  are well-defined and causal. We say that a well-posed nonlinear feedback system is *S-stable* (incr. *S-stable*) iff the map from the exogenous inputs to any subsystem variable is *S-stable* (incr. *S-stable*, resp.).

### III. Statement and Proof of the Theorems

We will need the following assumptions:

- A1.  $\underline{P}: L_{\tilde{v}}^{n_i} \rightarrow L_{\tilde{v}}^{n_o}$  and  $\underline{C}: L_{\tilde{v}}^{n_o} \rightarrow L_{\tilde{v}}^{n_i}$  are nonlinear causal maps between the appropriate extended spaces and the nominal system

$${}^1S(\underline{P}, \underline{C}) \text{ is well posed.} \tag{1}$$

A2.

$${}^1S(\underline{P}, \underline{C}) \text{ is S-stable.} \tag{2}$$

- A2'.  ${}^1S(\underline{P}, \underline{C})$  is *S-stable* and  $\underline{\Psi}_{\tilde{v}_1}: (u_1, u_2) \mapsto y_1$ , defined by  ${}^1S(\underline{P}, \underline{C})$ , is *incr. S-stable*.

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<sup>†</sup>By subsystem we mean any block of the block diagram of the feedback system.

<sup>†</sup>The meaning of  $(I + \underline{P}\underline{C})^{-1}$  deserves clarification: the map  $\underline{C}$  is composed with  $\underline{P}$  then the identity is added, and the resulting map is inverted. Although this formula has the same form as the linear case, it has a completely different interpretation.

A3. The additive perturbation  $\Delta P: L_{\mathbb{R}}^{n_1} \rightarrow L_{\mathbb{R}}^{n_0}$  is a nonlinear causal map subject only to the restriction that the perturbed system

$$\tilde{S}(P, \Delta P, C) \text{ be well-posed.} \quad (3)$$

**Remark 1:**  $P$ ,  $C$ , and  $\Delta P$  are nonlinear, causal maps subject only to (1), (2), and (3). None, some or all of these three maps may be unstable.

**Remark 2:** The nominal S-stable system  ${}^1S(P, C)$  with  $u_1, u_2$  as inputs and  $y_1, y_2$  as outputs, defines the S-stable maps  $\tilde{\Psi}_1: L_{\mathbb{R}}^{n_0} \times L_{\mathbb{R}}^{n_1} \rightarrow L_{\mathbb{R}}^{n_1}$  s.t.  $y_1 = \tilde{\Psi}_1(u_1, u_2) := Q_{u_2}(u_1)$  and  $\tilde{\Psi}_2: L_{\mathbb{R}}^{n_0} \times L_{\mathbb{R}}^{n_1} \rightarrow L_{\mathbb{R}}^{n_0}$  s.t.  $y_2 = \tilde{\Psi}_2(u_1, u_2) := H_{u_2}(u_1)$ .

**Theorem 1:** Let assumptions A1, A2 and A3 hold. U.t.c.,

(i)  $\tilde{S}(P, \Delta P, C)$  is S-stable  $\iff \forall u_2 \in L_{\mathbb{R}}^{n_1}, {}^1S(\Delta P, Q_{u_2})$  is S-stable.

(ii) If, instead of A2, the stronger assumption A2' holds and we let  $Q_0$  denote the partial map  $u_1 \rightarrow \tilde{\Psi}_1(u_1, 0)$ , then,

$$\tilde{S}(P, \Delta P, C) \text{ is S-stable} \iff {}^1S(P, Q_0) \text{ is S-stable} .$$

**Proof of (i):** ( $\Rightarrow$ ) Consider the system  $\tilde{S}(P, \Delta P, C)$  (see Fig. 1), and write its equations in terms of the outputs  $y_1, y_2$ , and  $y_3$ :

$$y_1 = C(u_1 - y_3 - y_2) \quad (4)$$

$$y_2 = P(u_2 + y_1) \quad (5)$$



$$y_3 = \Delta P(u_3 + y_1) \tag{6}$$

If, in Fig. 1, we delete the dotted part of the figure, we are left with  ${}^1S(\tilde{P}, \tilde{C})$  driven by  $u_1, u_2$  with outputs  $y_1, y_2$ :  ${}^1S(\tilde{P}, \tilde{C})$  is described by

$$y_1 = \tilde{C}(u_1 - y_2) \tag{7}$$

$$y_2 = \tilde{P}(u_2 + y_1) \tag{8}$$

By assumption,  ${}^1S(\tilde{P}, \tilde{C})$  is  $\mathbf{S}$ -stable, hence

$$y_i = \tilde{\Psi}_i(u_1, u_2) \tag{9}$$

where for  $i = 1, 2$ ,  $\tilde{\Psi}_i$  is a causal  $\mathbf{S}$ -stable map on  $L_{\mathbb{R}}^{n_1} \times L_{\mathbb{R}}^{n_2}$ . Using (9) in (4) and (5) we obtain a new equivalent representation for  $\tilde{S}(\tilde{P}, \Delta \tilde{P}, \tilde{C})$ :

$$y_1 = \tilde{\Psi}_1(u_1 - y_3, u_2) := Q_{u_2}(u_1 - y_3) \tag{10}$$

$$y_2 = \tilde{\Psi}_2(u_1 - y_3, u_2) \tag{11}$$

$$y_3 = \Delta P(u_3 + y_1) \tag{12}$$

Note that (10) and (12) involve only the outputs  $y_1$  and  $y_3$  and that (10) and (12) describe the system  ${}^1S(\Delta \tilde{P}, Q_{u_2})$  shown in Fig. 2. Now we obtain, successively:

$\tilde{S}(\tilde{P}, \Delta \tilde{P}, \tilde{C})$  is  $\mathbf{S}$ -stable hence the map  $(u_1, u_2, u_3) \mapsto (y_1, y_2, y_3)$  is  $\mathbf{S}$ -stable which implies that the map  $(u_1, u_2, u_3) \mapsto (y_1, y_3)$  defined by (10) and (12) is  $\mathbf{S}$ -stable, which, in turn, is equivalent to  $\forall u_2 \in L_{\mathbb{R}}^{n_2}$ ,  ${}^1S(\Delta \tilde{P}, Q_{u_2})$  is  $\mathbf{S}$ -stable.

( $\Leftarrow$ ) Assume that,  $\forall u_2 \in L_3^{n_1}$ ,  ${}^1S(\Delta P, Q_{\sim u_2})$  is  $\mathcal{S}$ -stable. Then, I) by the last equivalence, (10) and (12) specify, for the system  $\tilde{S}(P, \Delta P, C)$ , an  $\mathcal{S}$ -stable map from  $(u_1, u_2, u_3)$  to  $(y_1, y_3)$ ; II) since  ${}^1S(P, C)$  is  $\mathcal{S}$ -stable by assumption, the function  $\Psi_{\sim 2}$  in (11) is  $\mathcal{S}$ -stable; III) consequently, the *three* equations (10)-(12) define an  $\mathcal{S}$ -stable map from  $(u_1, u_2, u_3)$  to  $(y_1, y_2, y_3)$ ; equivalently  $\tilde{S}(P, \Delta P, C)$  is  $\mathcal{S}$ -stable.

**Proof of (ii):** ( $\Rightarrow$ ) By the theorem,  $\tilde{S}(P, \Delta P, C)$  is  $\mathcal{S}$ -stable, hence  $\forall u_2 \in L_3^{n_1}$   ${}^1S(\Delta P, Q_{\sim u_2})$  is  $\mathcal{S}$ -stable. In particular, for  $u_2 = 0$  we have that the *system*  ${}^1S(\Delta P, Q_{\sim 0})$  is  $\mathcal{S}$ -stable.

( $\Leftarrow$ ) By assumption  ${}^1S(\Delta P, Q_{\sim 0})$  is  $\mathcal{S}$ -stable.

We show that because  $(u_1, u_2) \rightarrow \Psi(u_1, u_2) =: Q_{\sim u_2}(u_1)$  is incr.  $\mathcal{S}$ -stable, then  $\forall u_2 \in L_3^{n_1}$ ,  ${}^1S(\Delta P, Q_{\sim u_2})$  is  $\mathcal{S}$ -stable. Then, by theorem 1, (i),  $\tilde{S}(P, \Delta P, C)$  is  $\mathcal{S}$ -stable.

Consider  ${}^1S(\Delta P, Q_{\sim u_2})$

$$y_1 = Q_{\sim u_2}(u_1 - y_3) \quad (13)$$

$$y_3 = \Delta P(u_3 + y_1) \quad (14)$$

We can rewrite  $Q_{\sim u_2}(u_1 - y_3)$  as follows

$$Q_{\sim u_2}(u_1 - y_3) = Q_{\sim 0}(u_1 - y_3) + \tilde{y}_1 = y_1^0 + \tilde{y}_1$$

where the last equalities define  $\tilde{y}_1$  and  $y_1^0$ , resp. Now  $\forall (u_1, u_2) \in L_3^{n_1} \times L_3^{n_2}$

and  $\forall T$ , we have:

$$\begin{aligned} \|\tilde{y}_1\|_T &\leq \|Q_{\tilde{u}_2}(u_1-y_3) - Q_{\tilde{u}_0}(u_1-y_3)\|_T = \|\Psi_{\tilde{u}_1}(u_1-y_3, u_2) - \Psi_{\tilde{u}_1}(u_1-y_3, 0)\|_T \\ &\leq \tilde{\varphi}(\|u_2\|_T) \end{aligned}$$

Since  $\tilde{\varphi} \in K_0 \subset K$ , for the system  ${}^1S(\Delta P, Q)$ , the map  $(u_1, u_2, u_3) \mapsto \tilde{y}_1$  is  $S$ -stable. (13) and (14) can be rewritten as:

$$y_1^o = Q_{\tilde{u}_0}(u_1-y_3) \tag{15}$$

$$y_3 = \Delta P(u_3 + \tilde{y}_1 + y_1^o) \tag{16}$$

The systems (13)-(14) and (15)-(16) are equivalent. Since  ${}^1S(\Delta P, Q)$  is  $S$ -stable and since  $(u_1, u_2, u_3) \mapsto \tilde{y}_1$  is  $S$ -stable we conclude that  $\forall u_2 \in L_{\tilde{u}_2}^{n_2}$   ${}^1S(\Delta P, Q_{\tilde{u}_2})$  is  $S$ -stable hence  $\tilde{S}(P, \Delta P, C)$  is  $S$ -stable, by Theorem 1, (i).

We replace assumption A3 by:

A3'. The multiplicative perturbation  $M = L_{\tilde{u}_2}^{n_2} \rightarrow L_{\tilde{u}_0}^{n_0}$  is a nonlinear causal map subject only to the restriction that the perturbed system  $\hat{S}(P, M, C)$  (see Fig. 3) be well-posed. Recall that  $y_2 = \Psi_{\tilde{u}_2}(u_1, u_2) := H_{\tilde{u}_2}(u_1)$  (see Remark 2). We now state:

**Theorem 2:** Let assumptions A1, A2 and A3' hold. U.t.c.  $\hat{S}(P, M, C)$  is  $S$ -stable  $\iff \forall u_2 \in L_{\tilde{u}_2}^{n_2}$ ,  ${}^1S(M, H_{\tilde{u}_2})$  ( see Fig. 2, caption ) is  $S$ -stable.

**Proof:** From Fig. 3, we write just as in Theorem 1.

$$y_1 = \tilde{\Psi}_{\tilde{1}}(u_1 - y_3, u_2) \quad (17)$$

$$y_2 = \tilde{\Psi}_{\tilde{2}}(u_1 - y_3, u_2) =: \tilde{H}_{\tilde{u}_2}(u_1 - y_3) \quad (18)$$

$$y_3 = \tilde{M}(u_3 + y_2) \quad (19)$$

Then, observe that (18) and (19) define  ${}^1S(\tilde{M}, \tilde{H}_{\tilde{u}_2})$  and follow the proof of Theorem 1, replacing  $\tilde{\Delta P}$  by  $\tilde{M}$ ,  $\tilde{Q}_{\tilde{u}_2}$  by  $\tilde{H}_{\tilde{u}_2}$  and  $\tilde{S}(\tilde{P}, \tilde{\Delta P}, \tilde{C})$  by  $\hat{S}(\tilde{P}, \tilde{M}, \tilde{C})$ .

#### IV. The Linear Case

Suppose that, in addition to assumption A1, A2, and A3 above,  $\tilde{P}$ ,  $\tilde{C}$ , and  $\tilde{\Delta P}$  are *linear*. (Note that we say linear, *not* linear and time-invariant; thus  $\tilde{P}: \mathbb{L}_{\tilde{u}}^{n_1} \rightarrow \mathbb{L}_{\tilde{y}}^{n_0}$  is a *linear map* (not necessarily represented by a transfer function), etc. ..). Consequently, S-stability and incr. S-stability are equivalent to finite-gain stability. By linearity,  $\tilde{\Psi}_{\tilde{1}}$  is linear and an easy calculation shows that

$$\tilde{\Psi}_{\tilde{1}}(u_1, u_2) = \tilde{Q}_{\tilde{0}} u_1 - \tilde{Q}_{\tilde{0}} \tilde{P} u_2 \quad (20)$$

where  $\tilde{Q}_{\tilde{0}}$  and  $\tilde{Q}_{\tilde{0}} \tilde{P}$  are finite-gain stable (by A2) linear maps. Thus the second part of Theorem 1 applies and we have: the linear system  $\tilde{S}(\tilde{P}, \tilde{\Delta P}, \tilde{C})$  is finite-gain stable  $\Leftrightarrow$  the linear system  ${}^1S(\tilde{\Delta P}, \tilde{Q}_{\tilde{0}})$  is finite-gain stable. Now for the closed loop system  ${}^1S(\tilde{\Delta P}, \tilde{Q}_{\tilde{0}})$  we have:

$$y_1 = \tilde{Q}_{\tilde{0}} (I - \tilde{M} \tilde{Q}_{\tilde{0}}) u_1 - \tilde{Q}_{\tilde{0}} \tilde{M} u_3 \quad (21)$$

$$y_3 = \underset{\sim}{M} \underset{\sim}{Q} u_1 + \underset{\sim}{M} u_3 \quad (22)$$

where  $\underset{\sim}{M} = \Delta \underset{\sim}{P} (I + \underset{\sim}{Q} \Delta \underset{\sim}{P})^{-1}$ . Since  $\underset{\sim}{Q}$  is finite-gain stable, (21), (22) show that  ${}^1S(\Delta \underset{\sim}{P}, \underset{\sim}{Q})$  is finite-gain stable if and only if  $\underset{\sim}{M} := \Delta \underset{\sim}{P} (I + \underset{\sim}{Q} \Delta \underset{\sim}{P})^{-1}$  is finite-gain stable. We state this as:

**Corollary 1:** Let  $\underset{\sim}{P}$ ,  $\underset{\sim}{C}$ ,  $\Delta \underset{\sim}{P}$  satisfy A1, A2, and A3 and let them be *linear* (but not necessarily time-invariant). U.t.c.,  $\tilde{S}(\underset{\sim}{P}, \Delta \underset{\sim}{P}, \underset{\sim}{C})$  is finite-gain stable  $\iff$  the linear map  $\Delta \underset{\sim}{P} (I + \underset{\sim}{Q} \Delta \underset{\sim}{P})^{-1}$  is finite-gain stable.

**Note:** Similar considerations apply to the case for multiplicative perturbations.

## References

- [Åst. 1] K. J. Åstrom, "Robustness of a design method based on assignment of poles and zeros," *IEEE Trans. Automat. Contr.*, Vol. AC-25, pp. 588-591, June 1980.
- [Bha. 1] A. Bhaya and C. A. Desoer, "Robust stability under additive perturbations," submitted to *IEEE Trans. on Automat. Contr.*, also ERL Memo No. UCB/ERL M84/110
- [Chen 1] M. J. Chen and C. A. Desoer, "Necessary and sufficient condition for robust stability of linear distributed feedback systems," *Int. J. Contr.*, Vol. 35, pp. 255-267, Feb. 1982.
- [Chen 2] M. J. Chen and C. A. Desoer, "Algebraic theory for robust stability of interconnected systems: necessary and sufficient conditions," *IEEE Trans. Automat. Contr.*, Vol. AC-29, No. 6, pp. 511-519, June 1984.
- [Cru. 1] J. B. Cruz, Jr., J. S. Freudenberg, and D. P. Looze, "A relationship between sensitivity and stability of multivariable feedback systems," *IEEE Trans. Automat. Contr.*, Vol. AC-26, pp. 66-74, Feb. 1981.
- [Des. 1] C. A. Desoer, F. M. Callier, and W. S. Chan, "Robustness of stability conditions for linear time-invariant feedback systems," *IEEE Trans. Automat. Contr.*, Vol. AC-22, pp. 586-590, Aug. 1977.
- [Des. 2] C. A. Desoer and M. Vidyasagar, "Feedback systems: Input-Output Properties," Academic Press, 1973.
- [Doy. 1] J. C. Doyle and G. Stein, "Multivariable feedback design: concepts for a classical/modern synthesis," *IEEE Trans. Automat. Contr.*, Vol. AC-26, pp. 4-16, Feb. 1981.
- [Fra. 1] B. A. Francis, "On robustness of the stability of feedback systems," *IEEE Trans. Automat. Contr.*, Vol. AC-25, pp. 817-818, Aug. 1980.

- [Pos. 1] I. Postlethwaite, J. M. Edmunds, and A. G. MacFarlane, "Principal gains and principal phases in the analysis of linear multivariable feedback systems," *IEEE Trans. Automat. Contr.*, Vol. AC-26, pp. 32-46, Feb. 1981.
- [Saf. 1] M. G. Safonov, "Stability and Robustness of Multivariable Feedback Systems," M.I.T. Press, Cambridge, MA, 1980.
- [San. 1] N. R. Sandell, Jr., "Robust stability of systems with applications to singular perturbations," *Automatica*, Vol. 15, pp. 467-470, 1979.
- [Zam. 1] G. Zames, "Nonlinear operators for systems analysis," Res. Lab. Electron., M.I.T., Cambridge, MA, Tech. Rep. 370, Aug. 1980 (especially pp. 34-37; the derivation is based on the small gain theorem).
- [Zam. 2] G. Zames, "Feedback and optimal sensitivity: Model reference transformations, multiplicative seminorms, and approximate inverses," *IEEE Trans. Automat. Contr.*, Vol. AC-26, pp. 301-320, April 1981.

### Figure Captions

- Fig. 1. The figure shows the system  $\tilde{S}(P, \Delta P, C)$  with inputs  $u_1, u_2, u_3$  and,  $y_1, y_2, y_3$ . If the dotted part of the diagram is removed, we are left with  ${}^1S(P, C)$  whose inputs are  $u_1, u_2$  and outputs  $y_1, y_2$ .
- Fig. 2.  ${}^1S(\Delta P, Q_{u_2})$ : obtained from Fig. 1. The "gain seen by  $\Delta P$ " going through  ${}^1S(P, C)$  is  $-Q_{u_2}$ . To obtain  ${}^1S(M, H_{u_2})$  replace  $\Delta P$  by  $M$  and  $Q_{u_2}$  by  $H_{u_2}$ .
- Fig. 3. The figure shows the multiplicatively perturbed system  $\hat{S}(P, M, C)$  with inputs  $u_1, u_2, u_3$  and outputs  $y_1, y_2$  and  $y_3$ .



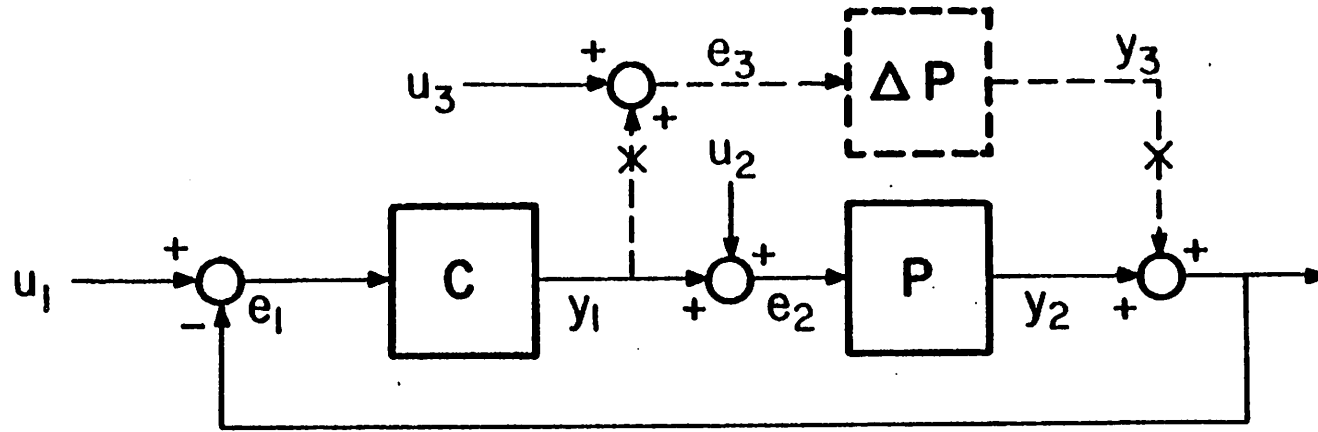


Fig. 1

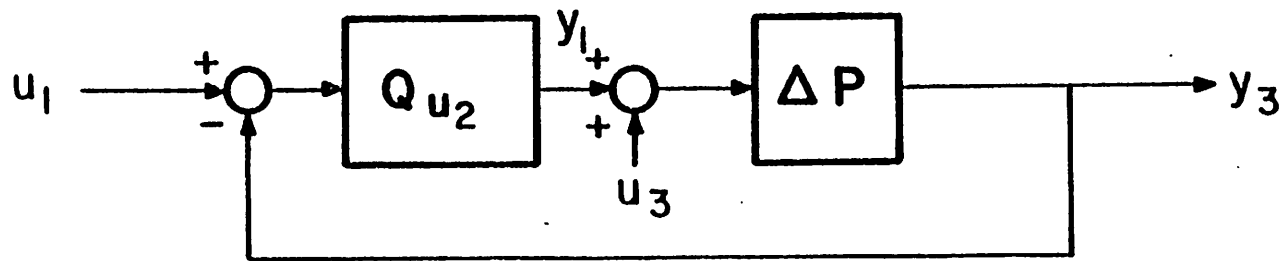


Fig. 2

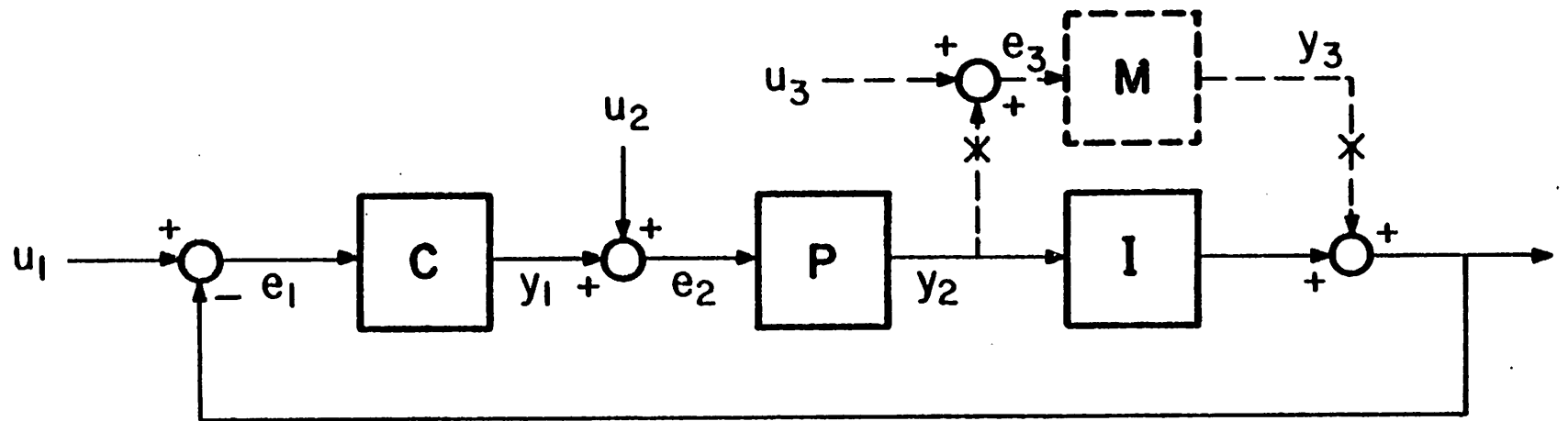


Fig. 3