# Design of Negative Capacitance Field-Effect Transistor



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# **Design of Negative Capacitance Field-Effect Transistor**

by Ming-Yen Kao

## **Research Project**

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#### I. Introduction

Gordon Moore proposed Moore's law in 1975, and after that, the number of components per integrated circuit doubles every 12-24 months [1]. The fast downscaling of the transistor benefits the whole world in many ways, including communication electronics (like iPhone and Internet router), personal computer... However, the growth of exponential is difficult to maintain at the same growth rate forever. For example, the exponential growth of the bacteria would be limited by the resources finally. There is no exception for semiconductor industry. For a long time, metal-oxide-semiconductor fieldeffect transistor (MOSFET) benefits from scaling and increase of clock frequency. By scaling effective oxide thickness (EOT), the same charge density can be achieved at lower gate voltage. By reducing the gate length, the same current density could be achieved at lower drain voltage. Several great inventions, including H-K metal gate, immersion lithography, and FinFET, keep the scaling continuing on. Scaling of EOT and increase of clock frequency together increase the performance per area. Recently, scaling of EOT is slowed down because of physical limitation, including direct tunneling, oxide breakdown, and mobility degradation, and the increase of clock frequency is hindered by the difficulty of heat dissipation.

Negative Capacitance Field-Effect Transistor (NCFET) is a promising near future solution to the slowed down of EOT scaling. The scaling can be pushed a few more nodes if NCFET is properly designed. The theory of NCFET will be explained in the next section.

# II. Theory of Negative Capacitance Field-Effect Transistor

Before explaining NCFET, let's start from how negative capacitance (NC) comes from. Fig. 1 (a) shows the energy of energy versus charge in ferroelectric (FE) and dielectric (DE) without external electric field [2]. The minimum energy of FE is not at zero charge, leading to the ferroelectricity (remanent polarization when there is no electric field). In contrast, the minimum energy of DE is at zero charge, and the shape is parabolic. If the FE is stacked on top of the DE, the total energy is plotted as the red curve in Fig. 1 (a). The equilibrium point is the small black ball at zero electric field, where the individual energy is not minimum for the FE, but the total energy of the system is minimum. Voltage is the derivative of energy with respect to the charge, and voltage versus charge plot is shown in Fig. 1 (b). Capacitance is the derivative of voltage with respect to charge, which is the slope in Fig. (b). The slope of the black curve in Fig. 1 (b) is negative, showing the negative capacitance within the red box.

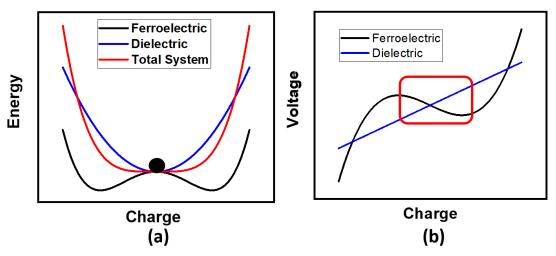


Fig. 1 (a) Energy versus charge plot of FE, DE, and the total system of FE and DE. (b) Voltage versus charge plot of FE and DE.

Fig. 2 shows the cartoon graph of a MOSFET. The capacitance from gate to channel can be simplified into two components, insulator capacitance  $(C_{ins})$  and semiconductor capacitance  $(C_s)$ . The total capacitance can be expressed as Eq. (1)

$$C_{total} = \frac{C_s C_{ins}}{C_s + C_{ins}}$$
 (1)

If the value of  $C_{ins}$  is negative,  $C_{total}$  could be larger than  $C_s$ . It is even better than EOT -> 0 scenario. Another way to demonstrate the benefits of NCFET is by looking into the equation of subthreshold slope (SS) [3]:

$$SS = \frac{\partial V_g}{\partial (log_{10}I)} = \frac{\partial V_g}{\partial \varphi_s} \frac{\partial \varphi_s}{\partial (log_{10}I)}$$
(2)

The SS of the device is limited by two parts. The first part is  $\frac{\partial V_g}{\partial \varphi_s}$ , which is determined by the capacitance divider as shown in Fig. 2.

$$\frac{\partial V_g}{\partial \varphi_{\rm S}} = 1 + \frac{c_{\rm S}}{c_{\rm ins}} (3)$$

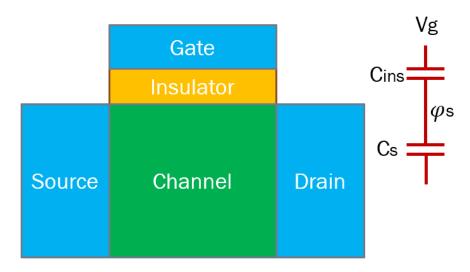


Fig. 2. Cartoon graph of a MOSFET.

From Eq. (3),  $\frac{\partial v_g}{\partial \varphi_s}$  is possible to be smaller than one if  $C_{ins}$  is negative. The second part is  $\frac{\partial \varphi_s}{\partial (log_{10}I)}$ , which is limited by the current control mechanism of the transistor and the temperature. For the conventional MOSFET, the minimum value of  $\frac{\partial \varphi_s}{\partial (log_{10}I)}$  is about 60 mV/dec at room temperature, and that is why 60mV/dec is called Boltzmann Tyranny. NC can help to overthrow the Boltzmann Tyranny by providing gain to  $\frac{\partial v_g}{\partial \varphi_s}$ .

# III. Optimization of NCFET by Matching Dielectric and Ferroelectric Nonuniformly Along the Channel

A new design to overcome the nonuniformity of capacitance matching along the channel of a negative capacitance field-effect transistor (NCFET) is presented in this section. By introducing nonuniform oxidation, the thickness of SiO<sub>2</sub> at the edge regions of the channel can be increased while maintaining the thickness of SiO<sub>2</sub> at the center region of the channel. As a result, the capacitance along the channel becomes more uniform, and better capacitance matching between the dielectric (DE) and ferroelectric (FE) can be achieved. The Sentaurus TCAD results show improvement of matching in the center region and a significant boost of on-current (20% improvement).

#### A. Motivation

The fundamental limit imposed by the Boltzmann distribution (60 mV/decade) which hinders scaling of CMOS technology is referred to as the Boltzmann Tyranny [3]-[5]. Tunneling field-effect transistors (TFETs), nanoelectromechanical (NEM) switches [6], and negative capacitance field-effect transistors (NCFETs) are promising ways to overcome the Boltzmann Tyranny [7]. However, the NEM switch is subject to reliability and scalability issues [5], while the TFET suffers from low on-current and other non-ideal effects [8]-[9]. NCFETs, on the other hand, can achieve an improved subthreshold slope (SS) while maintaining high on-current (compared to TFETs), and furthermore are fabricated through CMOS-compatible processes [10]. Many NCFETs have been made experimentally [10]-[15]. Nevertheless, many of them demonstrate a subthreshold swing (SS) of only 50-60 mV/decade without hysteresis and without an internal metal gate. An internal

metal gate breaks the ferroelectric into multiple domains, causing hysteresis [16], and introduces other non-ideal effects like charge trapping which lead to problems during device operation. Despite this, NCFETs without an internal metal gate usually demonstrate worse performance than NCFETs with an internal metal gate [17] because of bad capacitance matching in the center region of the channel. A possible solution to this problem will be proposed in this study. Many papers have discussed the nonuniformity in electric field, nonuniformity in FE [18]-[21], and methods of improving the degree of capacitance matching [22]-[23].

In this study, the difficulty of further reducing the SS below 60 mV/decade of hysteresis-free NCFETs without an internal metal gate will be demonstrated in B (i) and B (ii). The fringing field from the source and drain plays an important role in the capacitance matching in the subthreshold region, but the fringing field also limits the capacitance matching at the center of the channel. A possible solution will be proposed in B (iii) to further improve the performance of NCFETs.

#### **B.** Device Characterization and Discussion

#### i. Baseline UTB-SOI Device Structure

The baseline structure and design parameters are shown in Fig. 3, but without the FE layer. The electric field at  $V_g = 0V$  and  $V_d = 0.7V$  is plotted in Fig. 4 (a). The electric field is stronger at the edges of the gate oxide, as indicated by the red circles. The stronger electric field at the edges is caused by both the inner-fringing field (passing through the Si channel) and outer-fringing field (passing through the Si<sub>3</sub>N<sub>4</sub> spacer) [24], leading to nonuniform capacitance along the channel as shown in Fig. 4 (b) and imposing the

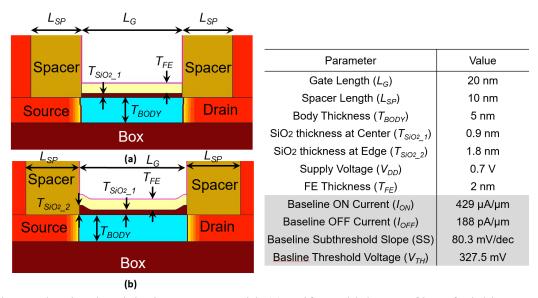


Fig. 3. The simulated device structure with (a) uniform thickness of interfacial layer and with (b) nonuniform thickness of interfacial layer. Table on the right-hand side lists important device parameters. The shaded parameters are for the baseline device. The baseline device has the same structure as (a), but without an FE layer.

limitation on capacitance matching. Note that the capacitance shown in Fig. 4 (b) is from the structures with an FE layer, and the vertical electric field in the  $SiO_2$  (perpendicular to the FE) right below the FE layer is used to extract the nonuniform  $C_{mos}$  by using Eq. (4) from the TCAD results.

$$C_{mos} = \frac{dQ}{dV} = \frac{d(E_{\perp FE} \times \epsilon_{SiO2})}{dV(x)}$$
 (4)

where  $E_{\perp FE}$  is the electric field perpendicular to and right below the FE,  $\epsilon_{SiO2}$  is the permittivity of SiO<sub>2</sub>, and dV(x) is the step size of electric potential at the interface of SiO<sub>2</sub> and FE as a function of position along the channel. To avoid hysteresis, the data in Fig. 4 (b) is extracted at  $V_g = 0.7V$  and  $V_d = 0.7V$ , where the minimum absolute value of FE capacitance is not smaller than any value of the curve.

# ii. Uniform interfacial layer NCFET

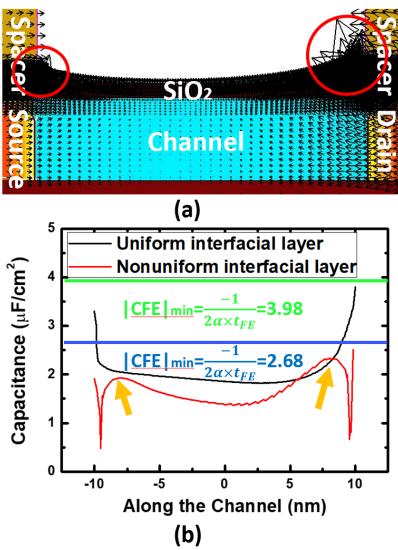


Fig. 4. (a) The electric field at  $V_g = 0V$  and  $V_d = 0.7V$ . The red circles highlight the higher electric field at the edges of the channel. (b) Capacitance versus position along the channel at  $V_g = 0.7V$ . The green line and the black curve are the traditional capacitance matching design, and the blue line and the red curve are our proposed capacitance matching design.

In traditional NCFET design, a uniform thickness of the interfacial layer along the channel is assumed, and the corresponding ferroelectric capacitance ( $C_{FE}$ ) is matched to its limit with  $C_{mos}$ . In HfO<sub>2</sub>-based NCFETs, the remanent polarization ( $P_r$ ) is more sensitive to ferroelectric doping concentration than coercive field ( $E_c$ ) [25]. Therefore,  $E_c$  is fixed to 2MV/cm, and  $P_r$  is decreased

until  $C_{FE}$  touches the largest  $C_{mos}$  along the channel. The smallest  $P_r$  reached in the TCAD simulation is  $11.5 \,\mu\text{C/cm}^2$ , which is equivalent to -3.98  $\mu\text{F/cm}^2$  (dielectric constant = 16 is included from [26]) by using Eq. (5). The physics models used in TCAD include the Ginzburg-Landau model for ferroelectric materials, Fermi Statistics, velocity saturation, Philips unified mobility model, Shockley-Read-Hall process, and quantum potential [27]. The Ginzburg-Landau model is shown in Eq. (6).

$$C_{FE} = \frac{1}{2\alpha t_{FE}} + \frac{16\varepsilon_0}{t_{FE}} (5)$$

$$E = 2\alpha P + 4\beta P^3 + 6\gamma P^5 - 2g\Delta P + \rho \frac{dP}{dt} (6)$$

In Eqs. (5) and (6), |CFE| is the estimation of the minimum absolute value of FE capacitance;  $t_{FE}$  is the thickness of the FE;  $\varepsilon_0$  is vacuum permittivity;  $\alpha$ ,  $\beta$ ,  $\gamma$  are the parameters for the FE; g is the strength of the polarization gradient (domain coupling) which is set to be 8E-5 cm<sup>3</sup>/F in this study (on the same order as [27]); and  $\rho$  is the viscosity that represents the finite time required for the polarization to switch.  $\alpha$ ,  $\beta$ , and  $\gamma$  are related to  $P_r$  and  $E_c$  by  $\alpha = -\frac{3\sqrt{3}}{4} \times \frac{E_c}{P_r}$  and  $\beta = \frac{3\sqrt{3}}{8} \times \frac{E_c}{P_r^3}$ , and  $\gamma = 0$  [29]. K. Chatterjee et. al. reports that the intrinsic delay of a doped hafnium oxide-based ferroelectric is negligible in digital circuits [30], so  $\rho$  is set to be zero in this study.

# iii. Proposed nonuniform interfacial layer NCFET

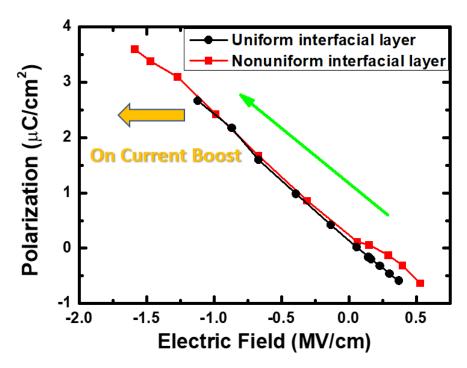


Fig. 5. Polarization and electric field of the FE right above the top of barrier (TOB). The transition from bottom right (deamplification) to upper left (amplification) represents the change of the state of the FE from Vg = 0V to Vg = 0.7V.

An NCFET with a nonuniform interfacial layer is proposed as shown in Fig. 3 (b). By tuning the thermal gradient or by introducing mask oxidation techniques during processing, a thicker  $SiO_2$  can be grown at the edge regions of the channel without affecting the thickness of the  $SiO_2$  at the center region of the channel. The  $P_r$  of the FE is now reduced to  $10.1~\mu\text{C/cm}^2$ , which is equivalent to  $-2.68~\mu\text{F/cm}^2$  (using dielectric constant = 16 from [24]) by using Eq. (5). The improvement from the perspective of polarization and electric field in the FE is shown in Fig. 5. Note that the metal gate work functions of the two cases are shifted to match the off-current of the baseline, so at  $V_g = 0.7$  V (bottom right), there is not much difference in the bias points. At  $V_g = 0.7$ 

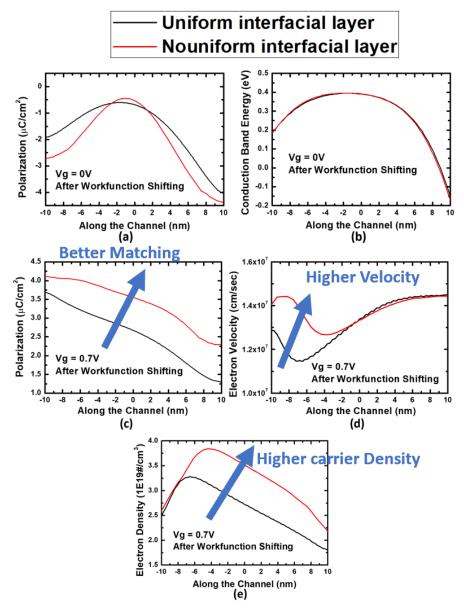


Fig. 6. (a) Polarization and (b) conduction band energy versus position along the channel in the off-state. (c) Polarization, (d) electron velocity, and (e) electron density versus position along the channel in the on-state. The black curves represent an NCFET with uniform thickness of the interfacial layer. The red curves represent an NCFET with a thicker interfacial layer at the edges of the channel.

V, the bias point of the proposed NCFET demonstrates an improvement via a left-shift as indicated by the yellow arrow (more voltage amplification).

Fig. 6 (a) plots polarization versus position along the channel at  $V_g =$ 0V (off-state). The profile of FE polarization changes because the capacitance matching condition along the channel changes. Fig. 6 (b) shows conduction band energy versus position along the channel. The barrier heights are the same because the metal gate work function is shifted to match the off-current of the baseline. Note that the position of the TOB is at the middle of the channel, which is why the matching between the FE and DE in the center of the channel is also critical. Fig. 6 (c) shows the FE polarization which is higher in the NCFET with nonuniform interfacial layer (more voltage amplification) compared to the normal NCFET. Fig. 6 (d) and 6 (e) show the carrier density and carrier velocity at a depth where the carrier concentration is highest (which is not close to the surface due to the quantum confinement effect). At the source side, a higher current can be supported by higher electron velocity, which can be seen in Fig. 6 (d). On the other hand, carriers at the drain side have already reached velocity saturation, so the carrier concentration increases more significantly near the drain side, as shown in Fig. 6 (e). It is evident from Fig. 7 that the proposed NCFET has better ON current (20% improvement) and better SS (minimal SS is 33 mV/decade). In Fig. 7, there are two bias points which show a surge in current. The first is around 0.265V and the second is around 0.615V. Better capacitance matching happens at these two gate voltages. Areas with better capacitance matching are indicated by two gold arrows in Fig. 4 (b). Better capacitance matching over the two areas accounts for the surge in current of the blue curve in Fig. 7 at 0.265V (which corresponds to the left hump of the red curve in Fig. 4(b)) and at 0.615V

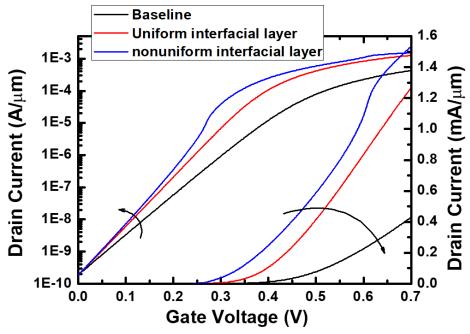


Fig. 7. Drain current versus gate voltage of baseline UTBSOI, traditional uniform-interfacial-layer NCFET, and proposed nonuniform-interfacial-layer NCFET.

(which corresponds to the right hump of the red curve in Fig. 4(b)). Note that these three cases are confirmed to be hysteresis-free by running the transient forward and reverse sweeping test. In comparison, the traditional design has no surge of current because the "good matching" parts are not in the channel and therefore contribute little current to the channel.

# **C.** Section Summery

The difficulty of capacitance matching due to the nonuniformity of capacitance along the channel is pointed out first in Fig. 4. To overcome this difficulty, a new design scheme utilizing a thicker interfacial SiO2 at the edges of the channel is proposed. The results show that the performance of the NCFET can be significantly boosted with this scheme. Therefore, the nonuniform capacitance caused by fringing fields should be taken into consideration when NCFETs are designed.

# IV. Variation Caused by Spatial Distribution of Dielectric and Ferroelectric Grains in a Negative Capacitance Field-Effect Transistor

A new scheme to consider the dielectric (DE) phases inside polycrystalline ferroelectric (FE) materials will be proposed in this section. The scheme is used to extract material parameters from experimental Polarization-Electric Field (P-E) measurements from the literature. A Sentaurus TCAD structure is constructed with the extracted parameters, and the simulated P-E curve is in good agreement with experimental data. Furthermore, variation of the device performance in a negative capacitance field-effect transistor (NCFET) due to the spatial distribution of DE and FE phases is studied using Sentaurus TCAD. It is found that the resultant variations of ON and OFF currents can be up to 14.44% and 30.23%, respectively, thus showing the impact of inhomogeneous crystalline phases of the FE material on device performance.

#### A. Motivation

As CMOS technology is aggressively scaled, power consumption becomes the most critical issue. To mitigate this issue, high mobility channel materials and three-dimensional transistors have been explored [31], [32]. However, the Boltzmann distribution poses a fundamental limit for lowering the energy dissipation in conventional electronics, and this limit is often referred to as the Boltzmann Tyranny [3]-[5], [9]. Negative capacitance FETs (NCFETs) are promising devices to overcome the subthreshold swing limit (60 mV/decade) imposed by the Boltzmann Tyranny and to achieve high Ion [3], [11]-[13], [33]-[39]. Although NCFETs experimentally exhibit sub-60

mV/decade performance [10]-[15], the non-uniformity effects of phases inside the ferroelectric (FE) haven't been investigated. To properly design NCFETs, the non-uniformity effects of phases should be carefully addressed. For example, X-ray diffraction (XRD) experimental data of HfO<sub>2</sub>-based ferroelectric thin films [10] shows that, other than the ferroelectric orthorhombic phase, there also exist cubic and monoclinic phases, which are dielectric (DE). Lun Xu et al. also reported the existence of monoclinic phases in ferroelectric HfO<sub>2</sub> films [40]. In previous work [41]-[42], the ferroelectric layer is assumed to be homogeneous so that its properties can be solely predicted by the Landau equation in the simulation, which fails to consider the essential physics in real devices.

In this section, we investigate how the locations of the dielectric grains affect the behavior of NCFETs using Sentaurus TCAD after extracting the material parameters from experimental data. The percentage of the dielectric phase in the ferroelectric is assumed to be the same under the same process conditions, but the position of the dielectric grains vary, which further affects the behavior of NCFETs. This random dielectric distribution imposes an additional variation on top of the variation from fabrication, and this variation should be properly understood to evaluate and minimize the impact on NCFET performance.

#### **B.** Device Characterization

# i. Ferroelectric parameters extraction

Symbol	Quantity	Unit
DE %	33.3	%
FE %	66.7	%
α	-5.810E+10	cm/F
β	3.286E+19	$cm^5/(F \cdot C^2)$
γ	2.165E+28	$cm^9/(F \cdot C^4)$
$P_0$	0.307	$\mu$ C/cm <sup>2</sup>
$E_0$	0.185	MV/cm
ε <sub>r</sub>	16.38	unit less

Table I. Hysteresis loop fitting results

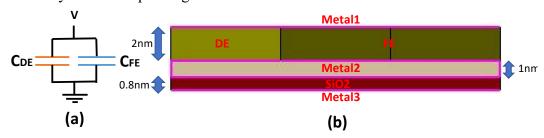


Fig. 8. (a) The circuit model used in Eq. (7) to (13) (b) MFMIM structure for TCAD

To set up the NCFET simulation using Sentaurus TCAD, the measured hysteresis loop of a 10nm ferroelectric from [14] is used to extract the ferroelectric parameters. As mentioned earlier, there are dielectric grains in the ferroelectric thin film, so only using the Landau equation to fit the hysteresis loop is insufficient. The expression for the Polarization-Electric Field (P-E) relation should include a dielectric component in addition to the Landau equation in the simulation. The model therefore should consist of a negative capacitor and a positive capacitor in parallel (Fig. 8 (a)). Note that the dielectric response is incorporated into the model to extend the sixth-order-polynomial approximation of Landau theory which has limited fitting capability in positive capacitance regions (the first and the third quadrants). This value of the dielectric constant is assumed to be the same as in DE grains.

Therefore, the polarization for a ferroelectric-dielectric-mixed thin film can be expressed as the following equations:

$$E_{FE} = 2\alpha \times P_{LD} + 4\beta \times P_{LD}^{3} + 6\gamma \times P_{LD}^{5}$$
(7)  

$$P_{FE} = P_{LD} + E_{FE} \times \varepsilon_{r} \times \varepsilon_{o}$$
(8)  

$$E_{mix} = E_{FE}$$
(9)  

$$P_{mix} = Area_{DE} \times E_{mix} \times \varepsilon_{r} \times \varepsilon_{o} + Area_{FE} \times P_{FE}$$
(10)

where  $E_{FE}$  is the electric field across negative capacitance;  $\alpha$ ,  $\beta$ , and  $\gamma$  are Landau coefficients;  $P_{LD}$  is the polarization of ferroelectric given by Landau Equation;  $P_{FE}$  is the polarization of FE part with built-in dielectric constant;  $\varepsilon_r$  is the dielectric constant;  $\varepsilon_o$  is the permittivity of vacuum;  $E_{mix}$  is the electric field across the thin film;  $P_{mix}$  is the total polarization with units of  $C/cm^2$ ;  $Area_{DE}$  and  $Area_{FE}$  are the area percentages of DE and FE in total area.

Two assumptions have to be made in order to properly extract the material parameters. It is assumed that the ferroelectric layer consists of 66.7% ferroelectric grains and 33.3% dielectric grains: a reasonable assumption because it has been reported [40] that the percentage of the monoclinic phase inside HfO2-based ferroelectric layers can range from 10% to 50%, depending on the processing conditions; and furthermore, there are cubic, tetragonal, and orthorhombic-dielectric phases possibly coexisting in the ferroelectric thin film. Another assumption is that the dielectric constants of all the grains are the same. Based on these assumptions, we can rewrite the polarization equation from (7), (8), (9) and (10):

$$E_{mix} = 2\alpha \times (P_{LD} - P_0) + 4\beta \times (P_{LD} - P_0)^3 + 6\gamma \times (P_{LD} - P_0)^5 + E_0$$

$$P_{FE} = P_{LD} + E_{mix} \times \varepsilon_r \times \varepsilon_o$$

$$P_{mix} = 33.3\% \times E_{mix} \times \varepsilon_r \times \varepsilon_o + 66.7\% \times P_{FE}$$
(13)

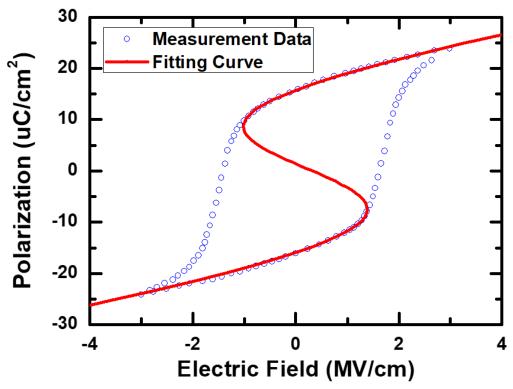


Fig. 9. Polarization-Electric Field loop of the proposed model (red line) and measured data (blue dots).

where  $P_0$  and  $E_0$  are the offset polarization and electric field due to the leakage in the thin film [43]. The extracted parameters are listed in Table 1, and the fitting results are shown in Fig. 9.

#### ii. Sentaurus TCAD MFMIM structure verification

To study the impact of the dielectric positions on NCFETs, the ferroelectric model (Landau equation) adopted in Sentaurus TCAD should be carefully calibrated. The physical models used in TCAD include the Ginzburg-Landau model for ferroelectric materials, mobility degradation due to carrier-carrier scattering, coulombic scattering, interface scattering, velocity saturation, and the Shockley-Read-Hall process [27]. The structure

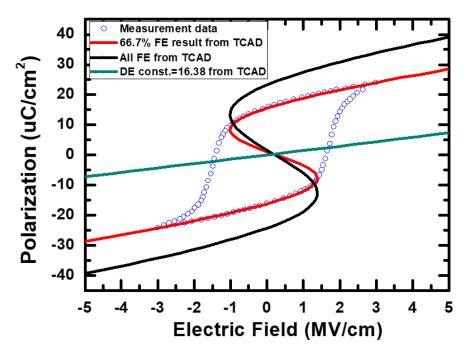


Fig. 10. Polarization versus electrical field plot.

for TCAD simulation is shown in Fig. 8 (b). In Fig. 8 (b), there is a FE-DE mixed layer which consists of 2/3 FE and 1/3 DE sandwiched by metal 1 and metal 2. The SiO<sub>2</sub> layer stabilizes the ferroelectric in the negative capacitance region. The  $\alpha$ ,  $\beta$ , and  $\gamma$  values of ferroelectric here are the same as the values in TABLE 1. Note that it is assumed no leakage in the FE layer in TCAD, so  $P_o$  and  $E_o$  is not used here. In Fig. 10, the curve generated by TCAD is shifted by  $P_o$  and  $E_o$  in the y-direction and x- direction respectively to align with experimental data.

To mimic the P-E loop from experimental measurement, the voltage of metal 1 is swept with metal 2 floating and metal 3 grounded. P-E loop is measured by sandwiching FE layer by two metal electrodes, so an internal metal is added in the TCAD simulation. The P-E curve can be extracted by plotting the charge density in metal 1 ( $P_{mix}$ ) versus the potential difference of metal 1 and

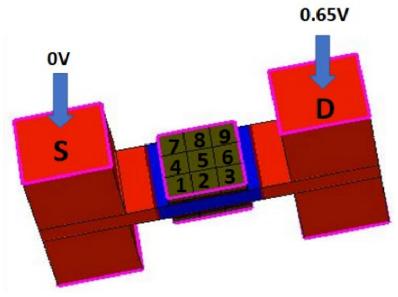


Fig. 11. NCFET structure in Sentaurus TCAD simulation. The red regions are source and drain. The blue region is 5nm-thick channel sandwiched by the gate stack, which consists of 0.8nm SiO<sub>2</sub>, 2nm segmented FE-DE mixed layer, and metal contact. The n-type source and drain doping are 2E20(#/cm<sup>3</sup>), and the p-type channel doping is 1E17(#/cm<sup>3</sup>). Gate work function is 4.6eV.

metal 2 over thickness of FE film ( $E_{mix}$ ). The P-E curve result from TCAD is shown in Fig. 10. Note that, in Fig. 10, the actual remnant polarization is higher, because the ferroelectric accounts for only 66.7% of the whole FE-DE mixed layer. At zero electric field, the DE portion contributes zero polarization so that the FE part needs 1.5 times of average remnant polarization to build up 1 time of average remnant polarization, showing that mixed FE-DE phases in the thin film would degrade the ferroelectricity.

#### iii. Sentaurus NCFET DE-FE mixed simulation

The ferroelectric layers of the n-channel double-gate NCFET with a gate length of 18 nm and a channel thickness of 5 nm are segmented into 3-by-3 matrix elements in the Sentaurus TCAD simulation as shown in Fig. 4. It is known that the grain size of HZO is in the same order as HZO thin film

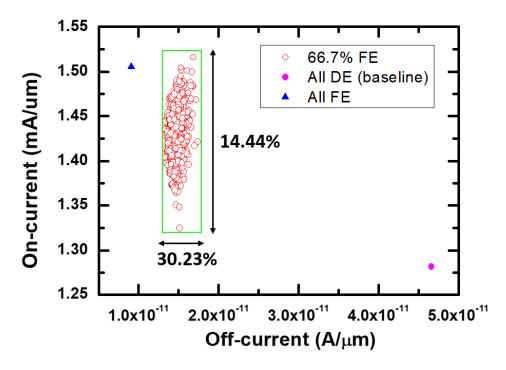


Fig. 12. Scatter plot of the random simulation results. The red circles are random simulation results, the pink dot represents that all the segments are DE (baseline), and the blue triangle represents that all the segments are FE.

thickness from the experiment [44], so grain size of 6nm by 6nm by 2nm is a reasonable assumption. Each element is either ferroelectric or dielectric with the same material parameters obtained from section B (ii). The gate stack is the same as the previous MFMIM structure shown in Fig. 9 except the intermediate metal layer is removed. Note that without intermediate metal layer, the spatial distribution of FE and DE matters, and that is why the same characterized FE film can bring out different characteristic in this part.

As mentioned in the previous section, 66.7% of the FE-DE layers are ferroelectric which means two thirds of segments/grains are ferroelectric. Note that the double-gate NCFET has top and bottom gate stacks with independent distributions of the DE and FE grains but 66.7% of each stack

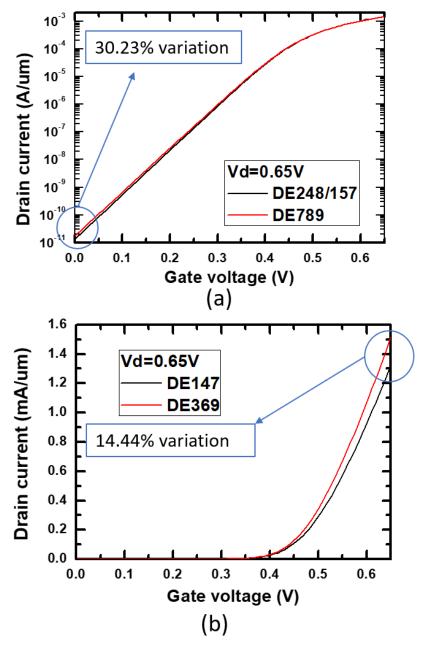


Fig. 13. Drain current versus gate voltage characteristics for (a)  $I_{\rm off}$  extreme cases with variation of 30.23% and (b)  $I_{\rm on}$  extreme cases with variation of 14.44%.

consists of FE grains. To examine how the distribution of the DE and FE grains would affect the current of the NCFET, a random simulation is carried out. There are 1128 possible combinations by considering two symmetric

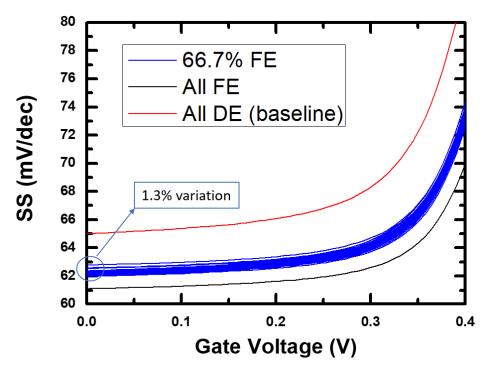


Fig. 14. The subthreshold slope versus gate voltage plot of the random DE and FE distribution simulation.

planes, and 1128 cases are simulated. For each case,  $V_{DD}$  is fixed at 0.65V.  $I_{off}$  and  $I_{on}$  are defined at  $V_{GS} = 0$  and  $V_{DD}$  at  $V_{DS} = V_{DD}$ , respectively. The results are shown in the scatter plot in Fig. 12.

In Fig. 12, the green box encloses the boundary of variation due to the different locations of FE and DE grains. Note that the all FE case means there is no dielectric, but  $\alpha$ ,  $\beta$ , and  $\gamma$  are modified to fit on the experimental P-E loop. The highest and lowest of  $I_{off}$  and  $I_{on}$  are shown in Fig. 13 (a) and (b), respectively. The highest  $I_{off}$  (red curve in Fig. 13(a)) happens when the DE grains on both gates are at the positions of 7, 8, and 9, whereas the lowest  $I_{off}$  (black curve) happens when the DE grains are at 2, 4, and 8 on the top gate and at 1, 5, and 7 on the bottom gate. In Fig. 13 (b), the highest  $I_{on}$  is obtained when the DE grains are at 3, 6, and 9 on both the top and bottom gates, which

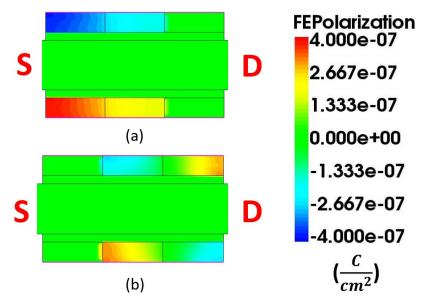


Fig. 15. Ferroelectric polarization 2-D plot at  $V_{GS} = V_{DD}$ . (a) Source-side ferroelectric (location of DE at number 3, 6, and 9 in Fig. 4) with the highest Ion, and (b) drain-side ferroelectric (location of DE at number 1, 4, and 7 in Fig. 4) with the lowest Ion.

means all FE grains are on the source side. On the other hand, the lowest Ion appears when DE grains are at 1, 4, and 7 on both top and bottom gates, which means that all FE grains are on the drain side. The subthreshold slope (SS) versus gate voltage is plotted in Fig. 14. The SS variation due to random spatial distribution of DE and FE is approximately 1.3%. Significant SS improvement from the baseline can be seen after adding FE.

#### C. Discussion

In the previous section, the random simulation shows the influence of the DE and FE distributions on the variation in drain current and SS. The physics of the simulation results will be analyzed in detail as follows.

#### i. On current extreme cases

As mentioned in the previous section, the highest I<sub>on</sub> happens when the ferroelectric grains are at the source side, and the lowest I<sub>on</sub> happens when

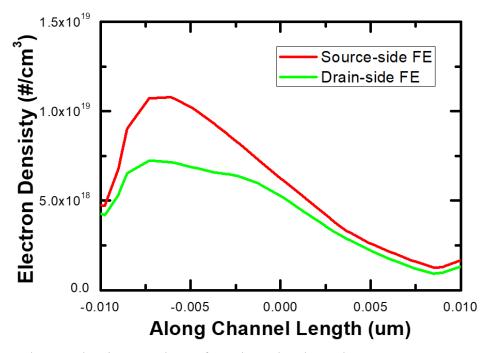


Fig. 16 Electron density near the surface along the channel.

ferroelectric grains are at the drain side. In Fig. 15 (a) and 15 (b), the polarization directions of the top and bottom FE are opposite, since the electric fields point in opposite directions for the top and bottom gate stacks. To make sure that the FE is in negative capacitance region, the absolute values of the FE polarization shown in Fig. 15 (range from -0.4  $\mu$  C/cm² to 0.4  $\mu$ C/cm²) should be within the range of negative capacitance region (range from -10  $\mu$ C/cm² to 10  $\mu$ C/cm²) as determined by the black curve in Fig. 10. Therefore, in the simulation the FE is always in the negative capacitance region. For the top gate in Fig. 15 (a), negative P<sub>FE</sub> (polarization pointing from gate to channel) means that the electric field points from channel to gate in the negative capacitance region. The electric field pointing from channel to gate means that the voltage at the interface of the FE and SiO<sub>2</sub> is higher than the applied gate voltage. The voltage amplification on bottom gate in Fig. 15 (a) can also be

#### **Lowest off-current Highest off-current** Leakage Path DE FΕ DE DE DE S D FE FE FE FE DE FΕ FE FE FE DE

Fig. 17. Layout of DE and FE for (a) the lowest  $I_{\rm off}$  case (the upper triangles and the lower triangles refer to top gate and bottom gate, respectively) and (b) the highest  $I_{\rm off}$  case.

(b)

(a)

explained in the same way. However, in Fig. 15 (b) the sign of the polarization changes from source to drain because of the fringing field from the drain to gate [41], which reduces the voltage on drain side.

To compare the ON current between these two cases, the inversion electron density along the channel length near the surface is plotted in Fig. 16. When the FE is at the source side (the carrier injection point of a MOSFET), the source inversion electron density increases due to voltage amplification. In contrast, when the FE is at the drain side, voltage de-amplification occurs because the positive drain voltage induces negative gate charges, which reduces the surface potential and inversion electron density. Therefore, the ON current of source-side FE is highest whereas that of the drain-side FE is lowest.

#### ii. Off current extreme cases

Fig. 10 (a) shows the lowest I<sub>off</sub> case where the upper triangles refer to top gate, and the lower triangles refer to bottom gate. The highest leakage path is

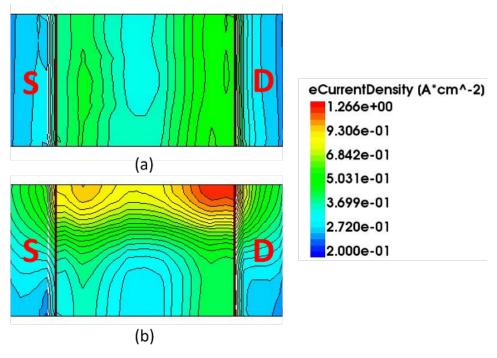


Fig. 18. Electron current density in the middle of the channel for (a) the lowest Ioff case (least leakage one and (b) the highest Ioff case (most leakage one).

at the center of channel due to degraded gate control. The layout of DE and FE grains in Fig. 17 (a) can control the leakage best among all the cases because the FE grains cover the entire drain side so that the fringing field from the drain can help the ferroelectric suppress the leakage current [41], and also evenly cover over the rest of the channel (see Fig. 18 (a)). In contrast, in Fig. 17 (b) all the DE grains are in a line along the channel, causing a leakage path along the DE region, which can be clearly seen in Fig. 18 (b).

### iii. Method of estimating the variation

The proposed method can be used to estimate the additional variation caused by the random spatial distribution of DE and FE grains. First of all, the percentage of DE and FE grains present should be quantified by XRD or other measurements. Following that, the parameters can be extracted by the method

described in section B (i). After getting all the parameters required for the simulation, one can get the extrema of I<sub>on</sub> by putting the FE grains all near the source and the drain, respectively. By putting the DE grains in a line along the channel, the highest I<sub>off</sub> can be obtained. By putting FE grains in a line on drain side and distributing the rest of FE grains evenly but complementary on two sides, the lowest I<sub>off</sub> can be obtained. Due to the inevitability of the existence of dielectric phases inside the ferroelectric [40], this additional variation should be taken into consideration. By using this method, device designers can estimate the window of variation by just running four cases of TCAD simulations.

## iv. The Capacitance matching when FE and DE are mixed

As shown in Fig. 10, the actual ferroelectricity is higher than the effective measured ferroelectricity. As a result, the capacitance matching is actually not good when the matching is made between an effective negative capacitance and a positive capacitance. By considering the FE-DE mixed model after extracting both DE and FE parameters, device designers can design their NCFETs better.

# **D. Section Summery**

A dielectric-ferroelectric mixed model is proposed to extract dielectric and ferroelectric material parameters by fitting the experimental hysteresis loop from literature. Based on these parameters, Sentaurus TCAD is properly calibrated using a MFMIM capacitor. After that, the impact of spatial distribution of dielectric and ferroelectric on NCFET performance is analyzed via Sentaurus TCAD, showing that the ON and OFF current variations can be

up to 14.44% and 30.23% respectively. This dielectric-ferroelectric mixed model is good to evaluate the variations in NCFET design.

## V. Conclusion

NCFET is a promising technology to extend the Moore's Law by reducing the EOT without reducing the physical gate oxide thickness or interfacial SiO<sub>2</sub> thickness. Therefore, the gate control can be enhanced without sacrificing gate leakage current or mobility. Two topics are brought up in this research project, including "optimization of NCFET by matching FE and DE nonuniformly along the channel" and "variation caused by spatial distribution of dielectric and ferroelectric grains in a negative capacitance field-effect transistor." The former studies how the NCFET device designer can improve the capacitance matching further, and the latter discusses how the circuit designer and the device designer can evaluate the variation of the NCFET device caused by DE grains in the FE film. If both the capacitance matching and the variation of NCFET are carefully considered, better NCFET can be made in the future.

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